Silicon Carbide (SiC) **Cascode JFET** - EliteSiC, **Power N-Channel, TOLL,** 750 V, 58 mohm

UJ4C075060L8S

Description

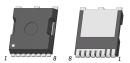
The UJ4C075060L8S is a 750 V, 58 m Ω G4 SiC FET. It is based on a unique 'cascode' circuit configuration, in which a normally-on SiC JFET is co-packaged with a Si MOSFET to produce a normally-off SiC FET device. The device's standard gate-drive characteristics allows use of off-the-shelf gate drivers hence requiring minimal re-design when replacing Si IGBTs, Si superjunction devices or SiC MOSFETs. Available in the space-saving H-PDSO-F8 package which enables automated assembly, this device exhibits ultra-low gate charge and exceptional reverse recovery characteristics, making it ideal for switching inductive loads and any application requiring standard gate drive.

Features

- On-resistance $R_{DS(on)}$: 58 m Ω (Typ)
- Operating Temperature: 175 °C (Max)
- Excellent Reverse Recovery: $Q_{rr} = 66 \text{ nC}$
- Low Body Diode V_{FSD}: 1.31 V
- Low Gate Charge: $Q_G = 37.8 \text{ nC}$
- Threshold Voltage V_{G(th)}: 4.8 V (Typ) Allowing 0 to 15 V Drive
- Low Intrinsic Capacitance
- ESD Protected, HBM Class 2
- H-PDSO-F8 Package for Faster Switching, Clean Gate Waveforms
- This Device is Pb-Free, Halogen Free and is RoHS Compliant

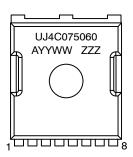
Typical Applications

- Line Rectification and Active-bridge Rectification Circuits in AC-DC Front-ends
- EV Charging
- PV Inverters
- Switch Mode Power Supplies
- Power Factor Correction Modules
- Motor Drives
- Induction Heating



H-PDSO-F8 **CASE 740AA**

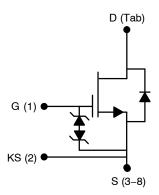
MARKING DIAGRAM



UJ4C075060 = Specific Device Code = Assembly Location

YY = Year = Work Week WW ZZZ = Lot ID

PIN CONNECTIONS



ORDERING INFORMATION

See detailed ordering and shipping information on page 9 of this data sheet.

MAXIMUM RATINGS

Parameter	Symbol	Test Conditions	Value	Unit
Drain-source Voltage	V _{DS}		750	V
Gate-source Voltage	V_{GS}	DC	-20 to +20	V
		AC (f > 1 Hz)	-25 to +25	
Continuous Drain Current (Note 1)	I _D	T _C = 25 °C	27.8	Α
		T _C = 100 °C	20.6	
Pulsed Drain Current (Note 2)	I _{DM}	T _C = 25 °C	82	Α
Single Pulsed Avalanche Energy (Note 3)	E _{AS}	L = 15 mH, I _{AS} = 1.8 A	24.3	mJ
SiC FET dv/dt Ruggedness	dv/dt	V _{DS} ≤ 500 V	200	V/ns
Power Dissipation	P _{tot}	T _C = 25 °C	155	W
Maximum Junction Temperature	$T_{J,max}$		175	°C
Operating and Storage Temperature	T_J , T_{STG}		-55 to 175	°C
Reflow Soldering Temperature	T _{solder}	Reflow MSL 1	260	°C

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

1. Limited by T_{J,max}
2. Pulse width t_p limited by T_{J,max}
3. Starting T_J = 25 °C

THERMAL CHARACTERISTICS

Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
Thermal Resistance, Junction-to-Case	$R_{\theta JC}$		-	0.75	0.97	°C/W

ELECTRICAL CHARACTERISTICS (T_J = +25 $^{\circ}$ C unless otherwise specified)

Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
TYPICAL PERFORMANCE - STATIC	-			-	-	-
Drain-source Breakdown Voltage	BV _{DS}	V _{GS} = 0 V, I _D = 1 mA	750	-	_	V
Total Drain Leakage Current	I _{DSS}	V _{DS} = 750 V, V _{GS} = 0 V, T _J = 25 °C	_	0.7	40	μΑ
		V _{DS} = 750 V, V _{GS} = 0 V, T _J = 175 °C	_	15	_	
Total Gate Leakage Current	I _{GSS}	V _{DS} = 0 V, T _J = 25 °C, V _{GS} = -20 V / +20 V	-	4.7	20	μΑ
Drain-source On-resistance	R _{DS(on)}	V _{GS} = 12 V, I _D = 20 A, T _J = 25 °C	_	58	74	mΩ
		V _{GS} = 12 V, I _D = 20 A, T _J = 125 °C	-	106	-	
		V _{GS} = 12 V, I _D = 20 A, T _J = 175 °C	-	147	_	
Gate Threshold Voltage	$V_{G(th)}$	V _{DS} = 5 V, I _D = 10 mA	4	4.8	6	V
Gate Resistance	R_{G}	f = 1 MHz, open drain	-	4.5	_	Ω
TYPICAL PERFORMANCE - REVERSE DIOD	Ε					
Diode Continuous Forward Current (Note 1)	I _S	T _C = 25 °C	_	-	27.8	Α
Diode Pulse Current (Note 2)	I _{S,pulse}	T _C = 25 °C	_	-	82	Α
Forward Voltage	V _{FSD}	V_{GS} = 0 V, I_S = 10 A, T_J = 25 °C	_	1.31	1.75	V
		V _{GS} = 0 V, I _S = 10 A, T _J = 175 °C	_	1.8	-	
Reverse Recovery Charge	Q _{rr}	V _{DS} = 400 V, I _S = 20 A, V _{GS} = 0 V,	_	66	-	nC
Reverse Recovery Time	t _{rr}	R_G = 33 Ω , di/dt = 1400 A/ μ s, $T_{,l}$ = 25 $^{\circ}$ C	_	16	_	ns
Reverse Recovery Charge	Q _{rr}	V _{DS} = 400 V, I _S = 20 A, V _{GS} = 0 V,	_	77	_	nC
Reverse Recovery Time	t _{rr}	$R_G = 33 \Omega$, di/dt = 1400 A/μs,		19	_	ns
•	۲rr	T _J = 150 °C		13		113
TYPICAL PERFORMANCE - DYNAMIC Input Capacitance	<u> </u>	V _{DS} = 400 V, V _{GS} = 0 V,		1420	_	pF
Output Capacitance	Ciss	f = 100 kHz	<u>-</u> -	41	_	ρı
Reverse Transfer Capacitance	Coss	-		2.7	_	
· ·	C _{rss}	V 0V/to 400 V/V 0V			_	~F
Effective Output Capacitance, Energy Related	C _{oss(er)}	V _{DS} = 0 V to 400 V, V _{GS} = 0 V		50	_	pF
Effective Output Capacitance, Time Related	C _{oss(tr)}	V _{DS} = 0 V to 400 V, V _{GS} = 0 V		94	_	pF
C _{OSS} Stored Energy Total Gate Charge	E _{oss}	$V_{DS} = 400 \text{ V}, V_{GS} = 0 \text{ V}$			_	μJ nC
Gate-drain Charge	Q _G	$V_{DS} = 400 \text{ V}, I_D = 20 \text{ A}, V_{GS} = 0 \text{ V to } 15 \text{ V}$		37.8 8	_	110
Gate-source Charge	Q _{GD}	-		11.8		
Turn-on Delay Time	Q _{GS}	V _{DS} = 400 V, I _D = 20 A,		10	_	ns
Rise Time	t _{d(on)}	Gate Driver = 0 V to +15 V,		27	_	113
Turn-off Delay Time		Turn-on $R_{G,EXT}$ = 1 Ω , Turn-off $R_{G,EXT}$ = 33 Ω		96	_	
Fall Time	t _{d(off)}	Inductive Load,		11		
Turn-on Energy	E _{ON}	FWD: same device with V_{GS} = 0 V, R_G = 33 Ω , T_J = 25 °C		165	_	μJ
Turn-off Energy		(Note 4)		42	_	μυ
Total Switching Energy	E _{OFF}	 		207	_	
Turn-on Delay Time		V _{DS} = 400 V, I _D = 20 A,		10	_	ns
Rise Time	t _{d(on)}	Gate Driver = 0 V to +15 V,		30	_	110
Turn-off Delay Time		Turn-on $R_{G,EXT}$ = 1 Ω , Turn-off $R_{G,EXT}$ = 33 Ω		100	_	
Fall Time	t _{d(off)}	Inductive Load,		12		
Turn-on Energy	E _{ON}	FWD: same device with V_{GS} = 0 V, R_G = 33 Ω , T_J = 150 °C		181	_	μJ
Turn-off Energy	E _{OFF}	(Note 4)		47		μο
Total Switching Energy		1		228		
Total Owitching Ellergy	E _{TOTAL}			220		L

ELECTRICAL CHARACTERISTICS ($T_J = +25$ °C unless otherwise specified) (continued)

Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
TYPICAL PERFORMANCE - DYNAMIC	•				•	•
Turn-on Delay Time	t _{d(on)}	V _{DS} = 400 V, I _D = 20 A,	_	12	_	ns
Rise Time	t _r	Gate Driver = 0 V to +15 V, Turn-on $R_{G,EXT}$ = 1 Ω ,	_	31	-	1
Turn-off Delay Time	t _{d(off)}	Turn-off $R_{G,EXT} = 5 \Omega$ Inductive Load,	_	42	_	
Fall Time	t _f	FWD: same device with	_	9	_	1
Turn-on Energy Including R _S Energy	E _{ON}	V_{GS} = 0 V, and R_{G} = 5 Ω , RC snubber: R_{S} = 10 Ω ,	_	183	-	μJ
Turn-off Energy Including R _S Energy	E _{OFF}	and $C_S = 100 \text{ pF}$, $T_J = 25 ^{\circ}\text{C}$	_	22	_	1
Total Switching Energy	E _{TOTAL}	(Notes 5, 6)	-	205	-	1
Snubber R _S Energy During Turn-on	E _{RS_ON}		_	0.95	_	1
Snubber R _S Energy During Turn-off	E _{RS_OFF}		_	1.41	_	1
Turn-on Delay Time	t _{d(on)}	V _{DS} = 400 V, I _D = 20 A,	-	12	-	ns
Rise Time	t _r	Gate Driver = 0 V to +15 V, Turn-on $R_{G,EXT}$ = 1 Ω ,	_	34	_	1
Turn-off Delay Time	t _{d(off)}	Turn-off $R_{G,EXT} = 5 \Omega$	_	47	-	1
Fall Time	t _f	Inductive Load, FWD: same device with V _{GS} = 0 V,	_	10	_	1
Turn-on Energy Including R _S Energy	E _{ON}	and R_G = 5 Ω , RC snubber: R_S = 10 Ω ,	_	207	_	μJ
Turn-off Energy Including R _S Energy	E _{OFF}	and $C_S = 100 \text{ pF}$, $T_J = 150 ^{\circ}\text{C}$	_	25	-	1
Total Switching Energy	E _{TOTAL}	(Notes 5, 6)	-	232	-	1
Snubber R _S Energy During Turn-on	E _{RS_ON}	1	_	0.91	-	1
Snubber R _S Energy During Turn-off	E _{RS_OFF}	1	_	1.42	_	1

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

4. Measured with the half-bridge mode switching test circuit in Figure 23.

^{5.} Measured with the chopper mode switching test circuit in Figure 24.

^{6.} In this table, the switching energies (turn-on energy, turn-off energy and total energy) presented include the device RC snubber energy losses.

TYPICAL PERFORMANCE DIAGRAMS

ID, Drain Current (A)

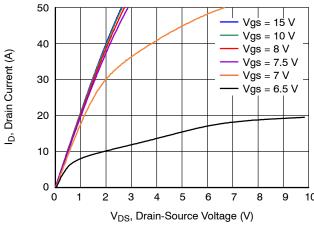


Figure 1. Typical Output Characteristics at $T_J = -55$ °C, $t_p < 250~\mu s$

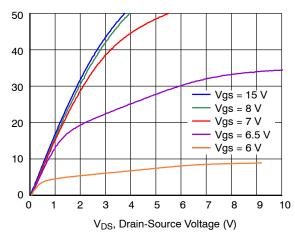


Figure 2. Typical Output Characteristics at T_J = 25 °C, t_p < 250 μs

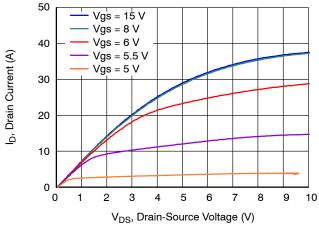


Figure 3. Typical Output Characteristics at T $_J$ = 175 $^{\circ}C,\,t_p<250~\mu s$

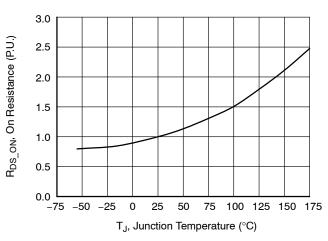


Figure 4. Normalized On-Resistance vs. Temperature at V_{GS} = 12 V and I_{D} = 20 A

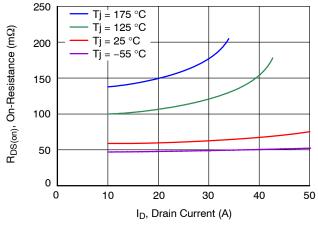


Figure 5. Typical Drain-Source On-Resistances at V_{GS} = 12 V

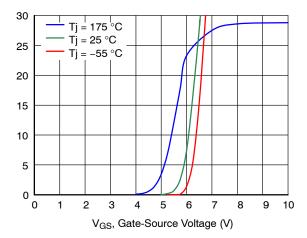


Figure 6. Typical Transfer Characteristics at $V_{DS} = 5 \text{ V}$

ID, Drain Current (A)

TYPICAL PERFORMANCE DIAGRAMS (continued)

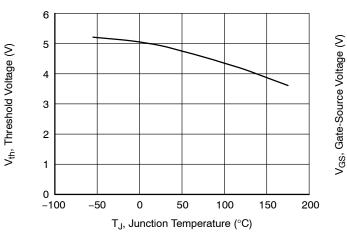


Figure 7. Threshold Voltage vs. Junction Temperature at V_{DS} = 5 V and I_{D} = 10 mA

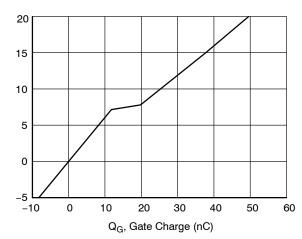


Figure 8. Typical Gate Charge at $V_{DS} = 400 \text{ V}$ and $I_D = 20 \text{ A}$

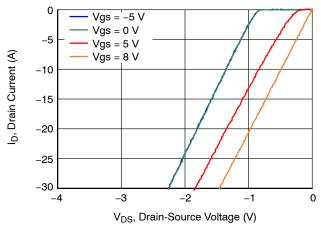


Figure 9. 3^{rd} Quadrant Characteristics at $T_J = -55$ °C

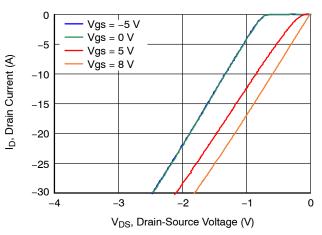


Figure 10. 3^{rd} Quadrant Characteristics at $T_J = 25$ °C

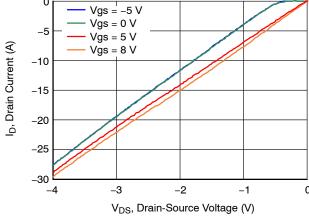


Figure 11. 3rd Quadrant Characteristics at T_J = 175 °C

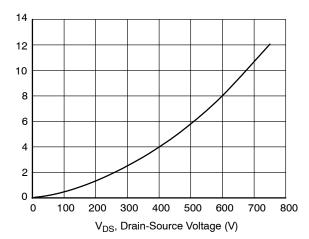


Figure 12. Typical Stored Energy in C_{OSS} at $V_{GS} = 0 \text{ V}$

Eoss (µJ)

TYPICAL PERFORMANCE DIAGRAMS (continued)

DC Drain Current (A)

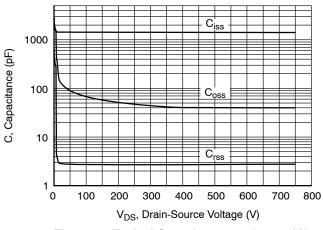


Figure 13. Typical Capacitances at f = 100 kHz and V_{GS} = 0 V

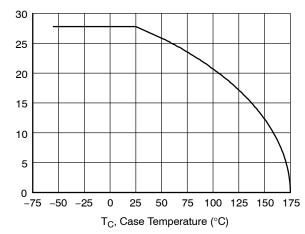


Figure 14. DC Drain Current Derating

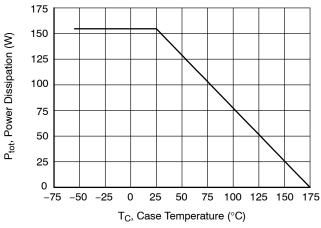


Figure 15. Total Power Dissipation

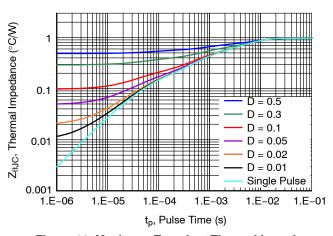


Figure 16. Maximum Transient Thermal Impedance

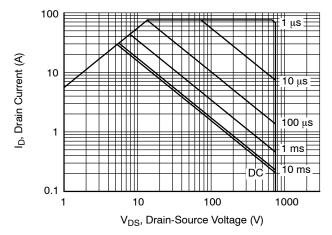


Figure 17. Safe Operation Area at T_C = 25 °C, D = 0, Parameter t_p

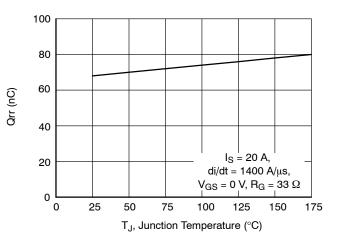


Figure 18. Reverse Recovery Charge Qrr vs. Junction Temperature at V_{DS} = 400 V

TYPICAL PERFORMANCE DIAGRAMS (continued)

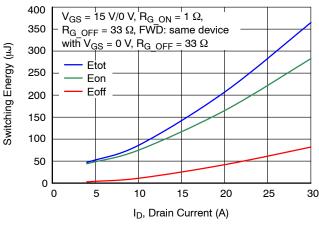


Figure 19. Clamped Inductive Switching Energy vs. Drain Current at V_{DS} = 400 V and T_J = 25 °C

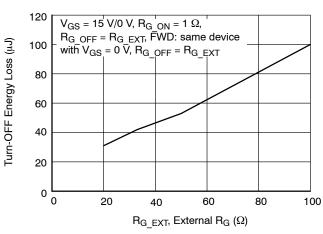


Figure 20. Clamped Inductive Switching Turn-off Energy vs. R_{G_EXT_OFF}

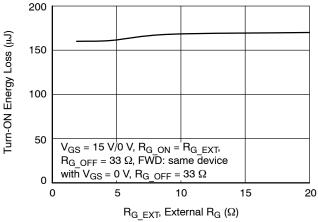


Figure 21. Clamped Inductive Switching Turn-on Energy vs. R_{G, EXT_ON}

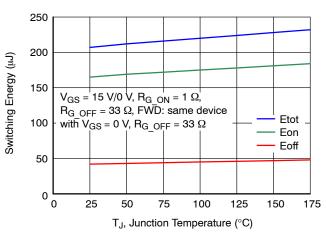


Figure 22. Clamped Inductive Switching Energy vs. Junction Temperature at V_{DS} = 400 V and I_{D} = 20 A

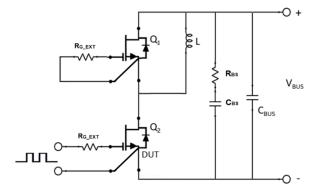


Figure 23. Schematic of the Half-bridge Mode Switching Test Circuit. Note, a Bus RC Snubber (R_{BS} = 2.5 Ω , C_{BS} = 100 nF) is Used to Reduce the Power Loop High Frequency Oscillations

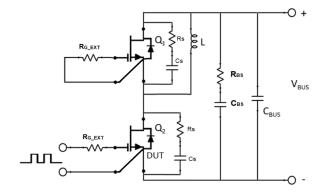


Figure 24. Schematic of the Half-bridge Mode Switching Test Circuit with Device RC Snubbers (Rs = 10 Ω , C_S = 100 pF) and a Bus RC Snubber (R_{BS} = 2.5 Ω , C_{BS} = 100 nF)

APPLICATIONS INFORMATION

SiC FETs are enhancement—mode power switches formed by a high–voltage SiC depletion—mode JFET and a low–voltage silicon MOSFET connected in series. The silicon MOSFET serves as the control unit while the SiC JFET provides high voltage blocking in the off state. This combination of devices in a single package provides compatibility with standard gate drivers and offers superior performance in terms of low on–resistance ($R_{DS(on)}$), output capacitance (C_{oss}), gate charge (Q_G), and reverse recovery charge (Q_{rr}) leading to low conduction and switching losses. The SiC FETs also provide excellent reverse conduction capability eliminating the need for an external anti–parallel diode.

Like other high performance power switches, proper PCB layout design to minimize circuit parasitics is strongly recommended due to the high dv/dt and di/dt rates. An external gate resistor is recommended when the FET is

working in the diode mode in order to achieve the optimum reverse recovery performance. For more information on SiC FET operation, see www.onsemi.com.

A snubber circuit with a small $R_{(G)}$, or gate resistor, provides better EMI suppression with higher efficiency compared to using a high $R_{(G)}$ value. There is no extra gate delay time when using the snubber circuitry, and a small $R_{(G)}$ will better control both the turn-off $V_{(DS)}$ peak spike and ringing duration, while a high $R_{(G)}$ will damp the peak spike but result in a longer delay time. In addition, the total switching loss when using a snubber circuit is less than using high $R_{(G)}$, while greatly reducing $E_{(OFF)}$ from mid-to-full load range with only a small increase in $E_{(ON)}$. Efficiency will therefore improve with higher load current. For more information on how a snubber circuit will improve overall system performance, visit the **onsemi** website at www.onsemi.com.

ORDERING INFORMATION

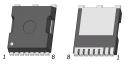
Part Number	Marking	Package	Shipping [†]
UJ4C075060L8S	UJ4C075060	H-PDSO-F8 (Pb-Free, Halogen Free)	2000 / Tape and Reel

[†]For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

REVISION HISTORY

Revision	Description of Changes	Date
D	Acquired the original Qorvo JFET Division Data Sheet and updated the main document title to comply with onsemi standards for SiC products.	1/15/2025
4	Converted the Data Sheet to onsemi format.	6/10/2025

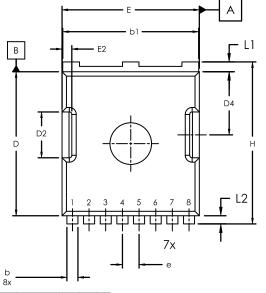


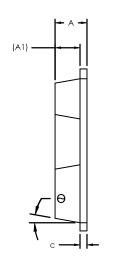


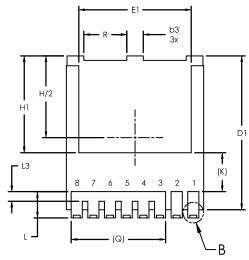
H-PDSO-F8 9.90x10.38x2.30, 1.20P

CASE 740AA ISSUE B

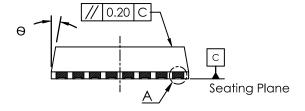
DATE 24 JUN 2025

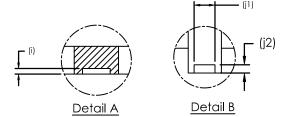






⊕ 0.25M C A B





Note:

- 1. Dimensioning and tolerancing as per ASME Y14.5 2018
- 2. Controlling dimension: millimeters
- 3. Dimensions does not include Burrs and Mold Flashes

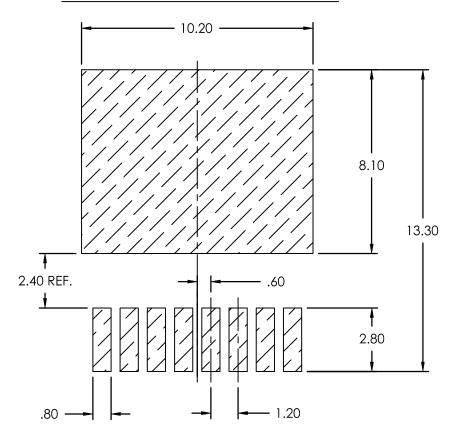
TO-LL					
SYMBOL	Value				
31MBOL	Min	Nom	Max		
Α	2.15	2.30	2.45		
A1		1.80 REF			
b	0.65	0.80	0.90		
b1	9.65	9.80	9.95		
b3	1.10	1.20	1.30		
c D	0.40	0.50	0.60		
	10.18	10.38	10.58		
D1	10.88	11.08	11.28		
D2	3.15	3.30	3.45		
D4	4.40	4.55	4.70		
Е	9.70	9.90	10.10		
E1	7.95	8.10	8.25		
E2	0.60	0.70	0.80		
Ф		1.20 BSC			
Н	11.48	11.68	11.88		
H1	6.80	6.95	7.10		
i		0.10 REF			
j1		0.46 REF			
j2		0.20 REF			
K		2.80 REF			
L	1.40	1.90	2.10		
L1	0.50	0.70	0.90		
L2	0.48	0.60	0.72		
L3 Q	0.30	0.70	0.80		
Q		6.80 REF			
R	3.00	3.10	3.20		
θ		10°			

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DESCRIPTION:	H-PDSO-F8 9.90x10.38x2.	H-PDSO-F8 9.90x10.38x2.30, 1.20P	

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DATE 24 JUN 2025

RECOMMENDED PCB LAND PATTERN



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