

# Silicon Carbide (SiC) Cascode JFET – EliteSiC, Power N-Channel, TOLL, 750 V, 33 mohm UJ4C075033L8S

## Description

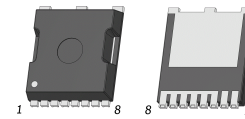
The UJ4C075033L8S is a 750 V, 33 mΩ G4 SiC FET. It is based on a unique ‘cascode’ circuit configuration, in which a normally-on SiC JFET is co-packaged with a Si MOSFET to produce a normally-off SiC FET device. The device’s standard gate-drive characteristics allows use of off-the-shelf gate drivers hence requiring minimal re-design when replacing Si IGBTs, Si superjunction devices or SiC MOSFETs. Available in the space-saving H-PDSO-F8 package which enables automated assembly, this device exhibits ultra-low gate charge and exceptional reverse recovery characteristics, making it ideal for switching inductive loads and any application requiring standard gate drive.

## Features

- On-resistance  $R_{DS(on)}$ : 33 mΩ (Typ)
- Operating Temperature: 175 °C (Max)
- Excellent Reverse Recovery:  $Q_{rr} = 89$  nC
- Low Body Diode  $V_{FSD}$ : 1.26 V
- Low Gate Charge:  $Q_G = 37.8$  nC
- Threshold Voltage  $V_{G(th)}$ : 4.8 V (Typ) Allowing 0 to 15 V Drive
- Low Intrinsic Capacitance
- ESD Protected: HBM Class 2 and CDM Class C3
- H-PDSO-F8 Package for Faster Switching, Clean Gate Waveforms
- This Device is Pb-Free, Halogen Free and is RoHS Compliant

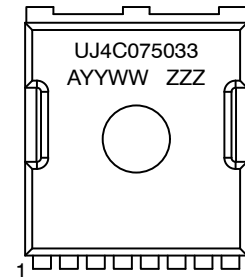
## Typical Applications

- Line Rectification and Active-bridge Rectification Circuits in AC–DC Front-ends
- EV Charging
- PV Inverters
- Switch Mode Power Supplies
- Power Factor Correction Modules
- Motor Drives
- Induction Heating



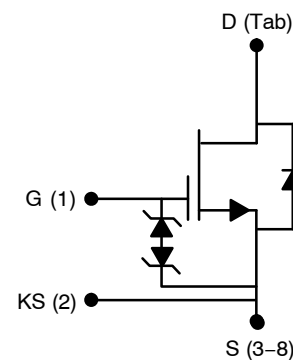
H-PDSO-F8  
CASE 740AA

## MARKING DIAGRAM



UJ4C075033 = Specific Device Code  
A = Assembly Location  
YY = Year  
WW = Work Week  
ZZZ = Lot ID

## PIN CONNECTIONS



## ORDERING INFORMATION

See detailed ordering and shipping information on page 9 of this data sheet.

## MAXIMUM RATINGS

Parameter	Symbol	Test Conditions	Value	Unit
Drain-source Voltage	$V_{DS}$		750	V
Gate-source Voltage	$V_{GS}$	DC	-20 to +20	V
		AC ( $f > 1$ Hz)	-25 to +25	
Continuous Drain Current (Note 1)	$I_D$	$T_C = 25\text{ }^{\circ}\text{C}$	44	A
		$T_C = 100\text{ }^{\circ}\text{C}$	33	
Pulsed Drain Current (Note 2)	$I_{DM}$	$T_C = 25\text{ }^{\circ}\text{C}$	132	A
Single Pulsed Avalanche Energy (Note 3)	$E_{AS}$	$L = 15\text{ mH}$ , $I_{AS} = 2.4\text{ A}$	43	mJ
SiC FET $dv/dt$ Ruggedness	$dv/dt$	$V_{DS} \leq 500\text{ V}$	200	V/ns
Power Dissipation	$P_{tot}$	$T_C = 25\text{ }^{\circ}\text{C}$	205	W
Maximum Junction Temperature	$T_{J,max}$		175	$^{\circ}\text{C}$
Operating and Storage Temperature	$T_J$ , $T_{STG}$		-55 to 175	$^{\circ}\text{C}$
Reflow Soldering Temperature	$T_{solder}$	Reflow MSL 1	260	$^{\circ}\text{C}$

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

1. Limited by  $T_{J,max}$
2. Pulse width  $t_p$  limited by  $T_{J,max}$
3. Starting  $T_J = 25\text{ }^{\circ}\text{C}$

## THERMAL CHARACTERISTICS

Parameter	Symbol	Test Conditions	Min	Typ	Max	Unit
Thermal Resistance, Junction-to-Case	$R_{\theta JC}$		-	0.56	0.73	$^{\circ}\text{C/W}$

# UJ4C075033L8S

## ELECTRICAL CHARACTERISTICS (T<sub>J</sub> = +25 °C unless otherwise specified)

Parameter	Symbol	Test Conditions	Min	Typ	Max	Unit
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### TYPICAL PERFORMANCE – STATIC

Drain-source Breakdown Voltage	BV <sub>DS</sub>	V <sub>GS</sub> = 0 V, I <sub>D</sub> = 1 mA	750	–	–	V
Total Drain Leakage Current	I <sub>DSS</sub>	V <sub>DS</sub> = 750 V, V <sub>GS</sub> = 0 V, T <sub>J</sub> = 25 °C	–	2	20	μA
		V <sub>DS</sub> = 750 V, V <sub>GS</sub> = 0 V, T <sub>J</sub> = 175 °C	–	20	–	
Total Gate Leakage Current	I <sub>GSS</sub>	V <sub>DS</sub> = 0 V, T <sub>J</sub> = 25 °C, V <sub>GS</sub> = –20 V / +20 V	–	6	±20	μA
Drain-source On-resistance	R <sub>DS(on)</sub>	V <sub>GS</sub> = 12 V, I <sub>D</sub> = 30 A, T <sub>J</sub> = 25 °C	–	33	41	mΩ
		V <sub>GS</sub> = 12 V, I <sub>D</sub> = 30 A, T <sub>J</sub> = 125 °C	–	57	–	
		V <sub>GS</sub> = 12 V, I <sub>D</sub> = 30 A, T <sub>J</sub> = 175 °C	–	75	–	
Gate Threshold Voltage	V <sub>G(th)</sub>	V <sub>DS</sub> = 5 V, I <sub>D</sub> = 10 mA	4	4.8	6	V
Gate Resistance	R <sub>G</sub>	f = 1 MHz, open drain	–	4.5	–	Ω

### TYPICAL PERFORMANCE – REVERSE DIODE

Diode Continuous Forward Current (Note 1)	I <sub>S</sub>	T <sub>C</sub> = 25 °C	–	–	44	A
Diode Pulse Current (Note 2)	I <sub>S,pulse</sub>	T <sub>C</sub> = 25 °C	–	–	132	A
Forward Voltage	V <sub>FSD</sub>	V <sub>GS</sub> = 0 V, I <sub>S</sub> = 15 A, T <sub>J</sub> = 25 °C	–	1.26	1.42	V
		V <sub>GS</sub> = 0 V, I <sub>S</sub> = 15 A, T <sub>J</sub> = 175 °C	–	1.59	–	
Reverse Recovery Charge	Q <sub>rr</sub>	V <sub>DS</sub> = 400 V, I <sub>S</sub> = 30 A, V <sub>GS</sub> = 0 V, R <sub>G,EXT</sub> = 50 Ω, di/dt = 1400 A/μs, T <sub>J</sub> = 25 °C	–	89	–	nC
Reverse Recovery Time	t <sub>rr</sub>		–	20.8	–	ns
Reverse Recovery Charge	Q <sub>rr</sub>	V <sub>DS</sub> = 400 V, I <sub>S</sub> = 30 A, V <sub>GS</sub> = 0 V, R <sub>G,EXT</sub> = 50 Ω, di/dt = 1400 A/μs, T <sub>J</sub> = 150 °C	–	97	–	nC
Reverse Recovery Time	t <sub>rr</sub>		–	21.6	–	ns

### TYPICAL PERFORMANCE – DYNAMIC

Input Capacitance	C <sub>iss</sub>	V <sub>DS</sub> = 400 V, V <sub>GS</sub> = 0 V, f = 100 kHz	–	1400	–	pF
Output Capacitance	C <sub>oss</sub>		–	68	–	
Reverse Transfer Capacitance	C <sub>rss</sub>		–	2.5	–	
Effective Output Capacitance, Energy Related	C <sub>oss(er)</sub>	V <sub>DS</sub> = 0 V to 400 V, V <sub>GS</sub> = 0 V	–	83	–	pF
Effective Output Capacitance, Time Related	C <sub>oss(tr)</sub>	V <sub>DS</sub> = 0 V to 400 V, V <sub>GS</sub> = 0 V	–	162	–	pF
C <sub>oss</sub> Stored Energy	E <sub>oss</sub>	V <sub>DS</sub> = 400 V, V <sub>GS</sub> = 0 V	–	6.6	–	μJ
Total gate Charge	Q <sub>G</sub>	V <sub>DS</sub> = 400 V, I <sub>D</sub> = 30 A, V <sub>GS</sub> = 0 V to 15 V	–	37.8	–	nC
Gate-drain Charge	Q <sub>GD</sub>		–	8	–	
Gate-source Charge	Q <sub>GS</sub>		–	11.8	–	
Turn-on Delay Time	t <sub>d(on)</sub>	V <sub>DS</sub> = 400 V, I <sub>D</sub> = 30 A, Gate Driver = 0 V to +15 V, Turn-on R <sub>G,EXT</sub> = 1 Ω, Turn-off R <sub>G,EXT</sub> = 50 Ω Inductive Load, FWD: same device with V <sub>GS</sub> = 0 V, R <sub>G</sub> = 50 Ω, T <sub>J</sub> = 25 °C (Note 4)	–	10.5	–	ns
Rise Time	t <sub>r</sub>		–	21	–	
Turn-off Delay Time	t <sub>d(off)</sub>		–	121	–	
Fall Time	t <sub>f</sub>		–	12	–	
Turn-on Energy	E <sub>ON</sub>		–	176	–	μJ
Turn-off Energy	E <sub>OFF</sub>		–	79	–	
Total Switching Energy	E <sub>TOTAL</sub>		–	255	–	
Turn-on Delay Time	t <sub>d(on)</sub>	V <sub>DS</sub> = 400 V, I <sub>D</sub> = 30 A, Gate Driver = 0 V to +15 V, Turn-on R <sub>G,EXT</sub> = 1 Ω, Turn-off R <sub>G,EXT</sub> = 50 Ω Inductive Load, FWD: same device with V <sub>GS</sub> = 0 V, R <sub>G</sub> = 50 Ω, T <sub>J</sub> = 150 °C (Note 4)	–	10.5	–	ns
Rise Time	t <sub>r</sub>		–	22	–	
Turn-off Delay Time	t <sub>d(off)</sub>		–	128	–	
Fall Time	t <sub>f</sub>		–	14	–	
Turn-on Energy	E <sub>ON</sub>		–	189	–	μJ
Turn-off Energy	E <sub>OFF</sub>		–	86	–	
Total Switching Energy	E <sub>TOTAL</sub>		–	275	–	

# UJ4C075033L8S

## ELECTRICAL CHARACTERISTICS ( $T_J = +25\text{ }^{\circ}\text{C}$ unless otherwise specified) (continued)

Parameter	Symbol	Test Conditions	Min	Typ	Max	Unit
<b>TYPICAL PERFORMANCE – DYNAMIC</b>						
Turn-on Delay Time	$t_{d(on)}$	$V_{DS} = 400\text{ V}$ , $I_D = 30\text{ A}$ , Gate Driver = $0\text{ V}$ to $+15\text{ V}$ , Turn-on $R_{G,EXT} = 1\text{ }\Omega$ , Turn-off $R_{G,EXT} = 5\text{ }\Omega$ Inductive Load, FWD: same device with $V_{GS} = 0\text{ V}$ , and $R_G = 5\text{ }\Omega$ , RC snubber: $R_S = 10\text{ }\Omega$ , and $C_S = 100\text{ pF}$ , $T_J = 25\text{ }^{\circ}\text{C}$ (Notes 5, 6)	–	11.5	–	ns
Rise Time	$t_r$		–	22	–	
Turn-off Delay Time	$t_{d(off)}$		–	33	–	
Fall Time	$t_f$		–	6	–	
Turn-on Energy Including $R_S$ Energy	$E_{ON}$		–	179	–	$\mu\text{J}$
Turn-off Energy Including $R_S$ Energy	$E_{OFF}$		–	18	–	
Total Switching Energy	$E_{TOTAL}$		–	197	–	
Snubber $R_S$ Energy During Turn-on	$E_{RS\_ON}$		–	0.95	–	
Snubber $R_S$ Energy During Turn-off	$E_{RS\_OFF}$		–	1.43	–	
Turn-on Delay Time	$t_{d(on)}$	$V_{DS} = 400\text{ V}$ , $I_D = 30\text{ A}$ , Gate Driver = $0\text{ V}$ to $+15\text{ V}$ , Turn-on $R_{G,EXT} = 1\text{ }\Omega$ , Turn-off $R_{G,EXT} = 5\text{ }\Omega$ Inductive Load, FWD: same device with $V_{GS} = 0\text{ V}$ , and $R_G = 5\text{ }\Omega$ , RC snubber: $R_S = 10\text{ }\Omega$ , and $C_S = 100\text{ pF}$ , $T_J = 150\text{ }^{\circ}\text{C}$ (Notes 5, 6)	–	11	–	ns
Rise Time	$t_r$		–	22	–	
Turn-off Delay Time	$t_{d(off)}$		–	33.4	–	
Fall Time	$t_f$		–	7	–	
Turn-on Energy Including $R_S$ Energy	$E_{ON}$		–	185	–	$\mu\text{J}$
Turn-off Energy Including $R_S$ Energy	$E_{OFF}$		–	20	–	
Total Switching Energy	$E_{TOTAL}$		–	205	–	
Snubber $R_S$ Energy During Turn-on	$E_{RS\_ON}$		–	1	–	
Snubber $R_S$ Energy During Turn-off	$E_{RS\_OFF}$		–	1.41	–	

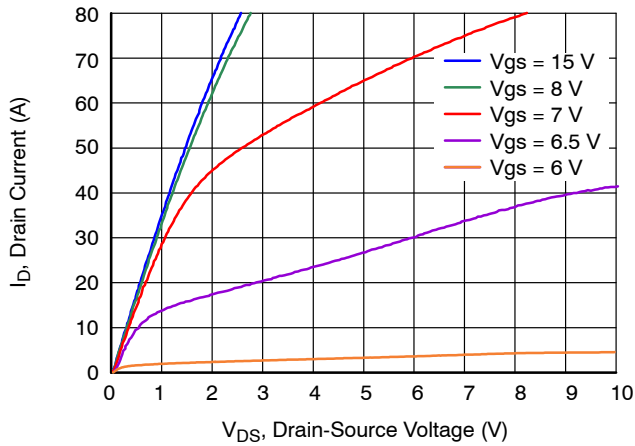
Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

4. Measured with the switching test circuit in Figure 23.

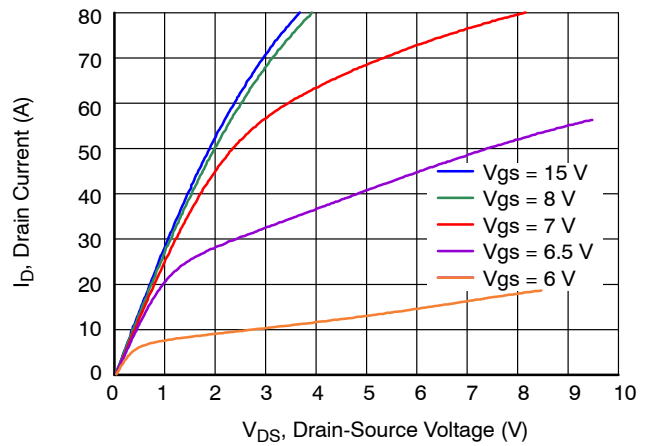
5. Measured with the switching test circuit in Figure 24.

6. In this table, the switching energies (turn-on energy, turn-off energy and total energy) presented include the device RC snubber energy losses.

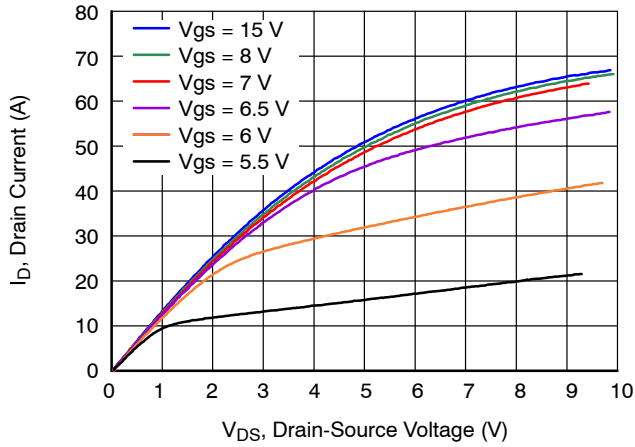
TYPICAL PERFORMANCE DIAGRAMS



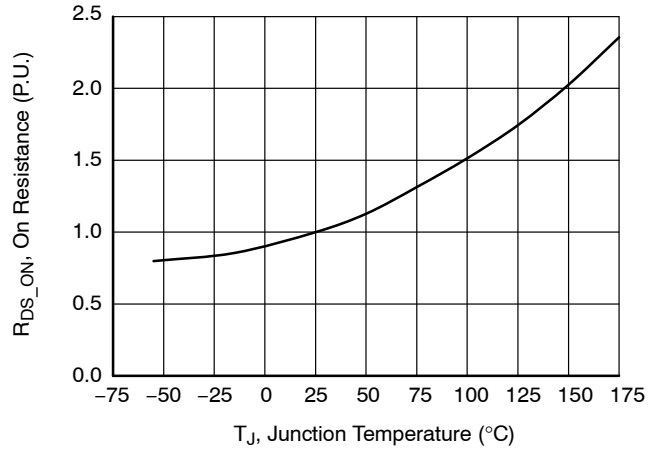
**Figure 1. Typical Output Characteristics**  
at  $T_J = -55\text{ }^{\circ}\text{C}$ ,  $t_p < 250\text{ }\mu\text{s}$



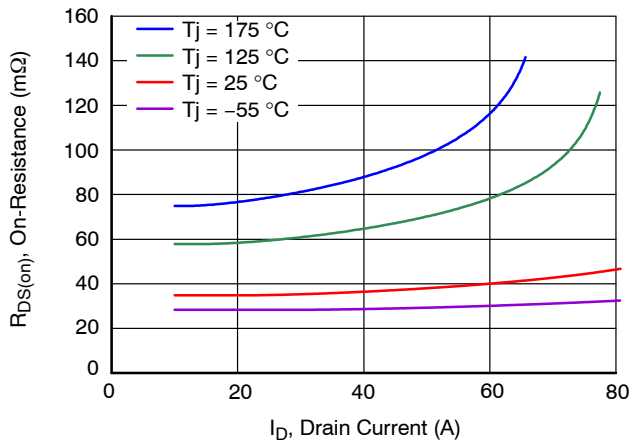
**Figure 2. Typical Output Characteristics**  
at  $T_J = 25\text{ }^{\circ}\text{C}$ ,  $t_p < 250\text{ }\mu\text{s}$



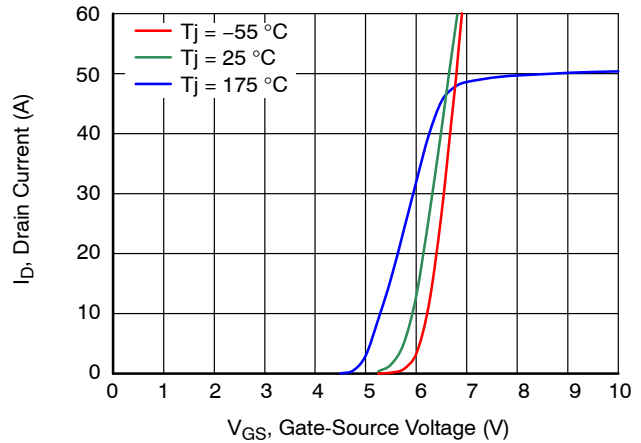
**Figure 3. Typical Output Characteristics**  
at  $T_J = 175\text{ }^{\circ}\text{C}$ ,  $t_p < 250\text{ }\mu\text{s}$



**Figure 4. Normalized On-Resistance**  
vs. Temperature at  $V_{GS} = 12\text{ V}$  and  $I_D = 30\text{ A}$



**Figure 5. Typical Drain-Source On-Resistances**  
at  $V_{GS} = 12\text{ V}$



**Figure 6. Typical Transfer Characteristics**  
at  $V_{DS} = 5\text{ V}$

TYPICAL PERFORMANCE DIAGRAMS (continued)

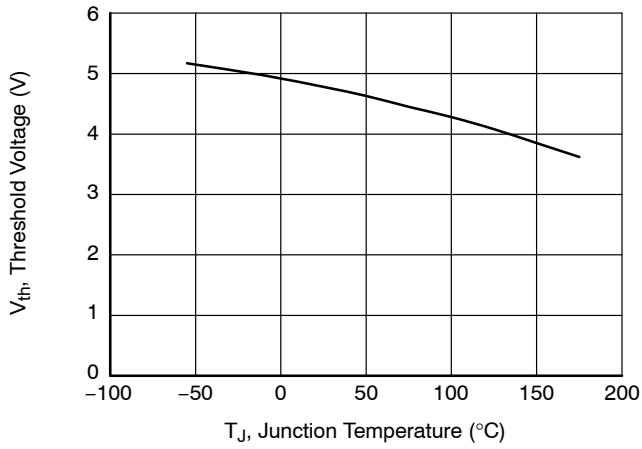


Figure 7. Threshold Voltage vs. Junction Temperature at  $V_{DS} = 5\text{ V}$  and  $I_D = 10\text{ mA}$

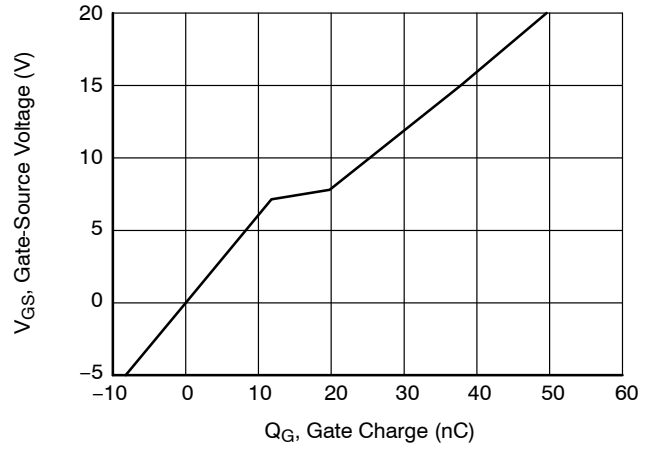


Figure 8. Typical Gate Charge at  $I_D = 30\text{ A}$

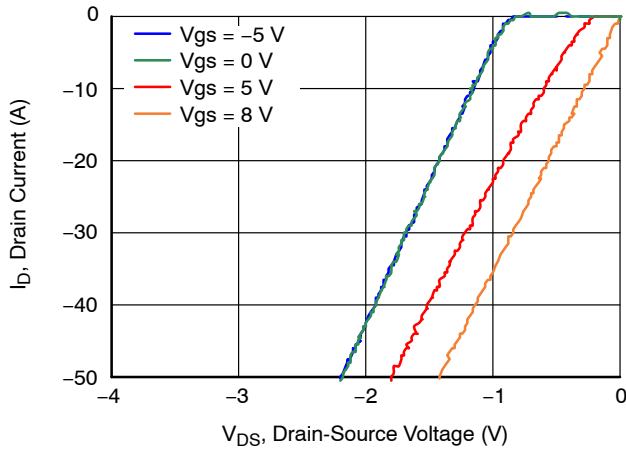


Figure 9. 3<sup>rd</sup> Quadrant Characteristics at  $T_J = -55\text{ }^{\circ}\text{C}$

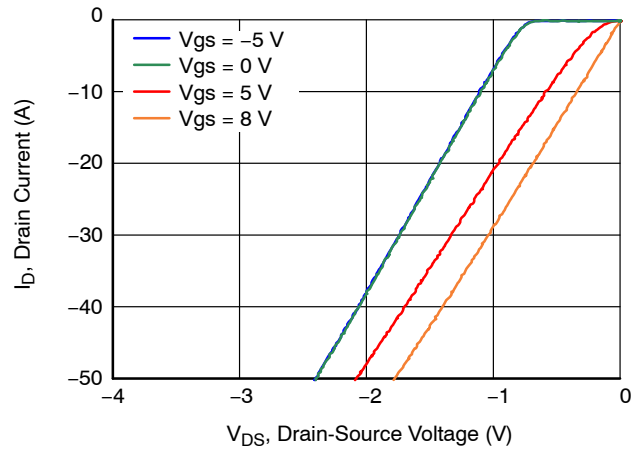


Figure 10. 3<sup>rd</sup> Quadrant Characteristics at  $T_J = 25\text{ }^{\circ}\text{C}$

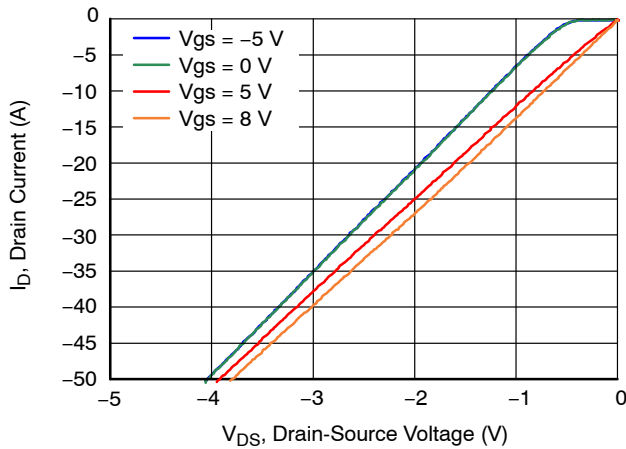


Figure 11. 3<sup>rd</sup> Quadrant Characteristics at  $T_J = 175\text{ }^{\circ}\text{C}$

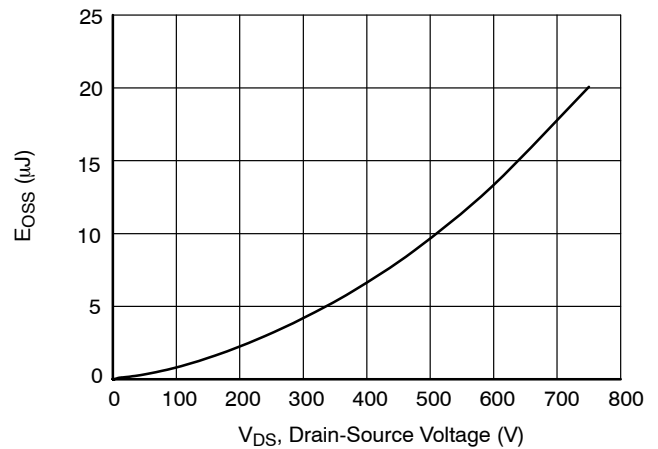


Figure 12. Typical Stored Energy in  $C_{OSS}$  at  $V_{GS} = 0\text{ V}$

TYPICAL PERFORMANCE DIAGRAMS (continued)

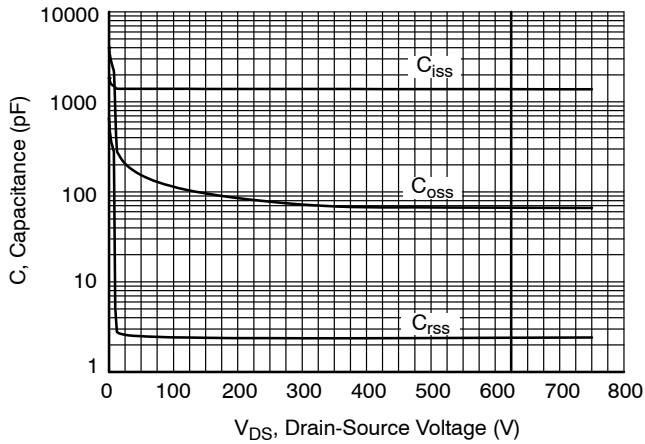


Figure 13. Typical Capacitances at  $f = 100 \text{ kHz}$  and  $V_{GS} = 0 \text{ V}$

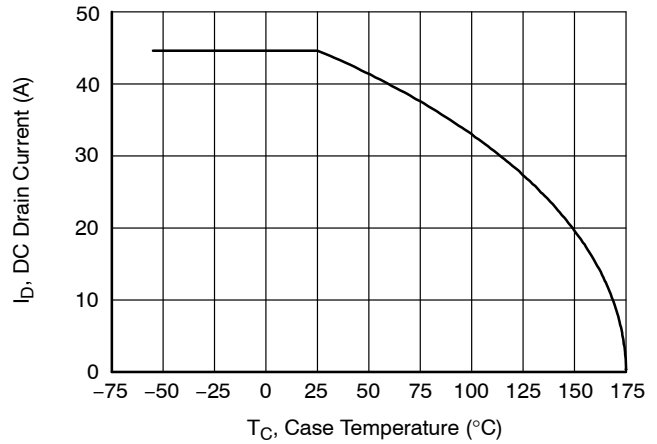


Figure 14. DC Drain Current Derating

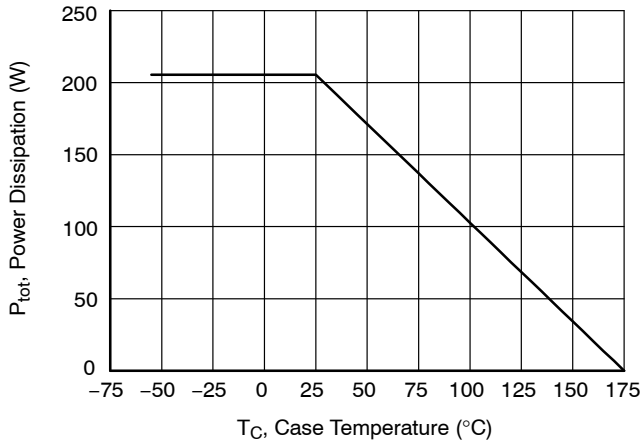


Figure 15. Total Power Dissipation

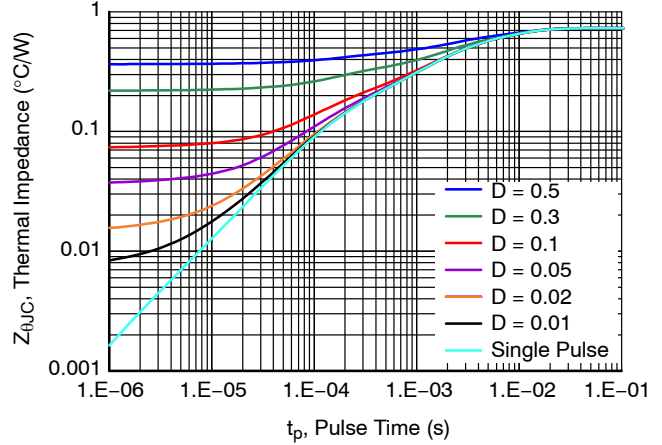


Figure 16. Maximum Transient Thermal Impedance

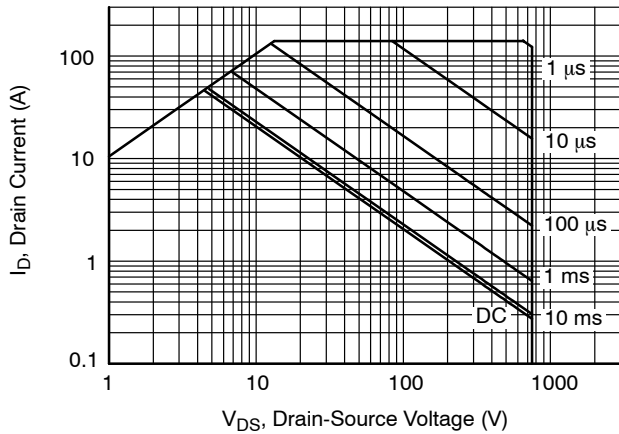


Figure 17. Safe Operation Area at  $T_C = 25 \text{ }^{\circ}\text{C}$ ,  $D = 0$ , Parameter  $t_p$

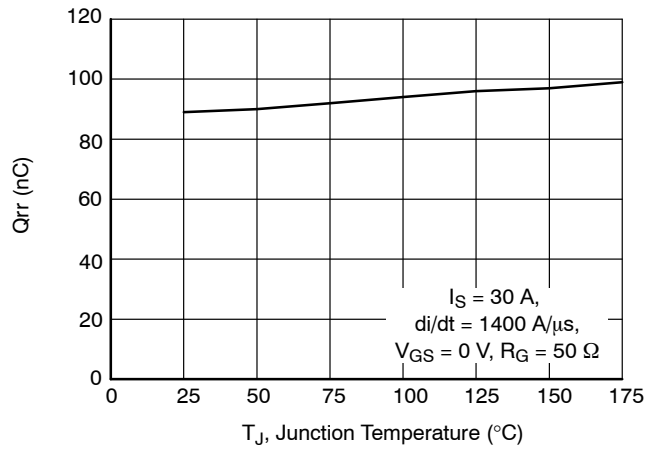


Figure 18. Reverse Recovery Charge  $Q_{rr}$  vs. Junction Temperature at  $V_{DS} = 400 \text{ V}$

TYPICAL PERFORMANCE DIAGRAMS (continued)

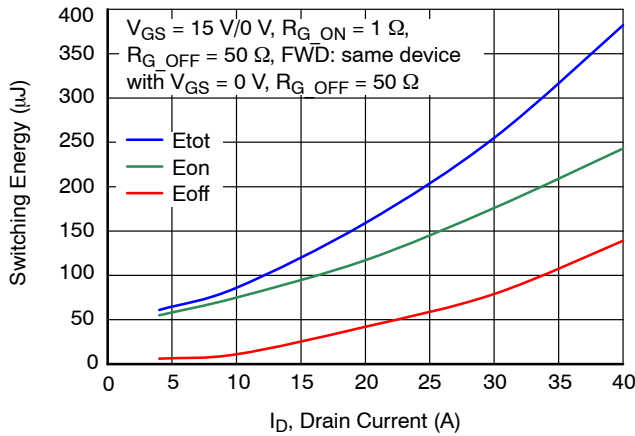


Figure 19. Clamped Inductive Switching Energy vs. Drain Current at  $V_{DS} = 400\text{ V}$  and  $T_J = 25^\circ\text{C}$

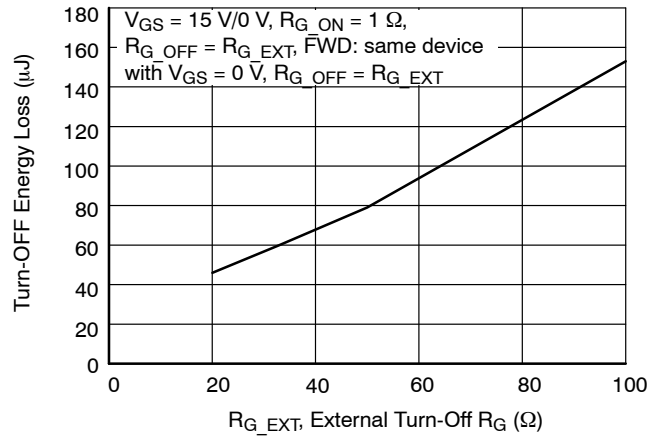


Figure 20. Turn-OFF Clamped Inductive Switching Energies vs.  $R_{G\_EXT}$  at  $V_{DS} = 400\text{ V}$ ,  $I_D = 30\text{ A}$ , and  $T_J = 25^\circ\text{C}$

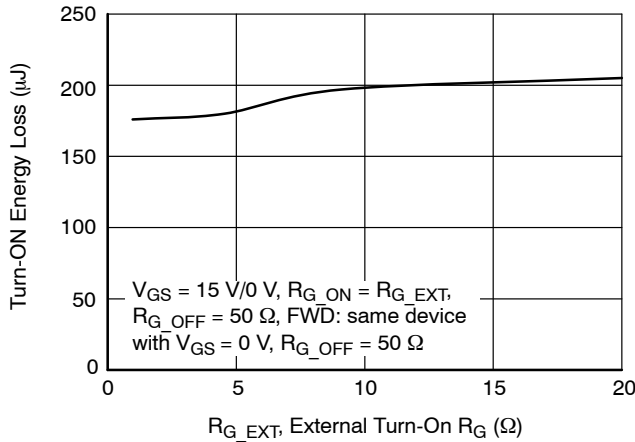


Figure 21. Turn-ON Clamped Inductive Switching Energies vs.  $R_{G\_EXT}$  at  $V_{DS} = 400\text{ V}$ ,  $I_D = 30\text{ A}$ , and  $T_J = 25^\circ\text{C}$

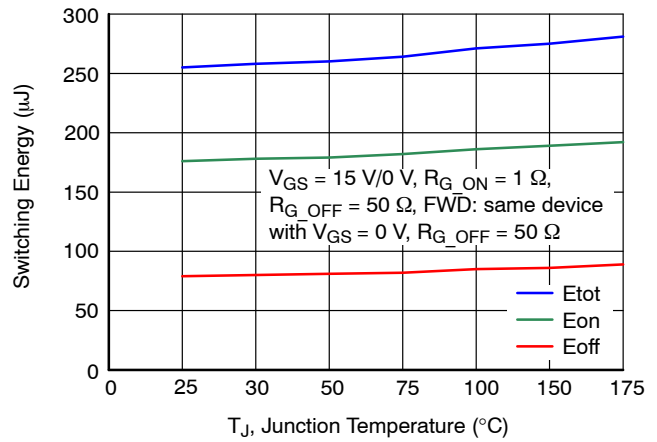


Figure 22. Clamped Inductive Switching Energy vs. Junction Temperature at  $V_{DS} = 400\text{ V}$  and  $I_D = 30\text{ A}$

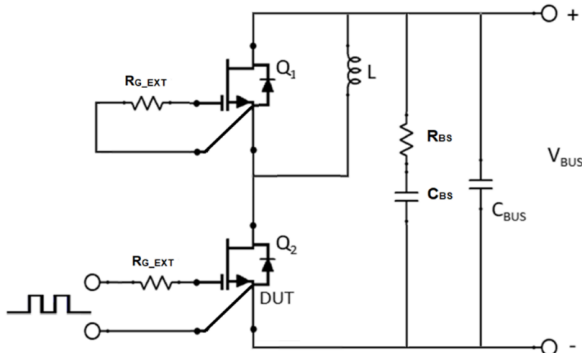


Figure 23. Schematic of the Half-bridge Mode Switching Test Circuit. Note, a Bus RC Snubber ( $R_{BS} = 2.5\ \Omega$ ,  $C_{BS} = 100\text{ nF}$ ) is Used to Reduce the Power Loop High Frequency Oscillations

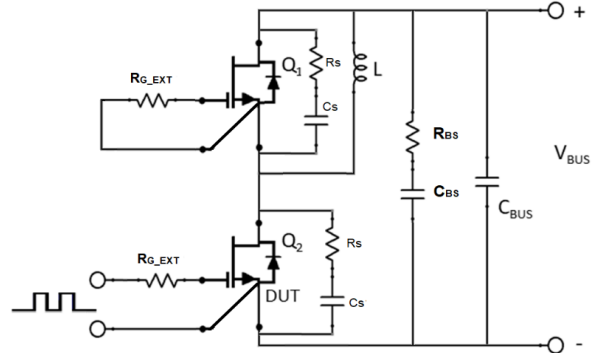


Figure 24. Schematic of the Half-bridge Mode Switching Test Circuit with Device RC Snubbers ( $R_S = 10\ \Omega$ ,  $C_S = 100\text{ pF}$ ) and a Bus RC Snubber ( $R_{BS} = 2.5\ \Omega$ ,  $C_{BS} = 100\text{ nF}$ )



## APPLICATIONS INFORMATION

SiC FETs are enhancement-mode power switches formed by a high-voltage SiC depletion-mode JFET and a low-voltage silicon MOSFET connected in series. The silicon MOSFET serves as the control unit while the SiC JFET provides high voltage blocking in the off state. This combination of devices in a single package provides compatibility with standard gate drivers and offers superior performance in terms of low on-resistance ( $R_{DS(on)}$ ), output capacitance ( $C_{oss}$ ), gate charge ( $Q_G$ ), and reverse recovery charge ( $Q_{rr}$ ) leading to low conduction and switching losses. The SiC FETs also provide excellent reverse conduction capability eliminating the need for an external anti-parallel diode.

Like other high performance power switches, proper PCB layout design to minimize circuit parasitics is strongly recommended due to the high  $dv/dt$  and  $di/dt$  rates. An external gate resistor is recommended when the FET is

working in the diode mode in order to achieve the optimum reverse recovery performance. For more information on SiC FET operation, see [www.onsemi.com](http://www.onsemi.com).

A snubber circuit with a small  $R_G$ , or gate resistor, provides better EMI suppression with higher efficiency compared to using a high  $R_G$  value. There is no extra gate delay time when using the snubber circuitry, and a small  $R_G$  will better control both the turn-off  $V_{(DS)}$  peak spike and ringing duration, while a high  $R_G$  will damp the peak spike but result in a longer delay time. In addition, the total switching loss when using a snubber circuit is less than using high  $R_G$ , while greatly reducing  $E_{(OFF)}$  from mid-to-full load range with only a small increase in  $E_{(ON)}$ . Efficiency will therefore improve with higher load current. For more information on how a snubber circuit will improve overall system performance, visit the **onsemi** website at [www.onsemi.com](http://www.onsemi.com).

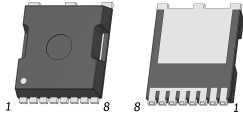
## ORDERING INFORMATION

Part Number	Marking	Package	Shipping <sup>†</sup>
UJ4C075033L8S	UJ4C075033	H-PDSO-F8 (Pb-Free, Halogen Free)	2000 / Tape and Reel

<sup>†</sup>For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, [BRD8011/D](http://www.onsemi.com).

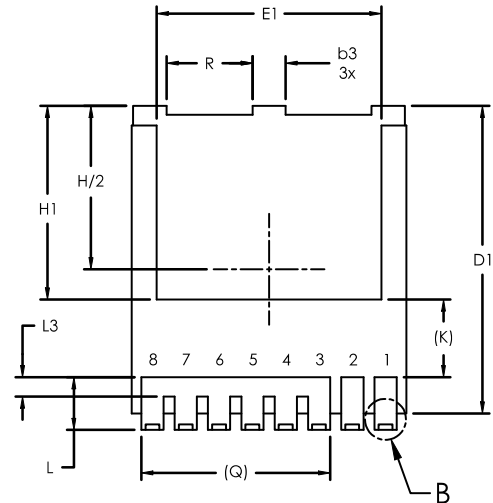
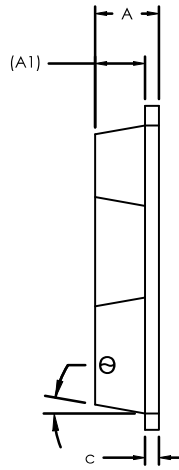
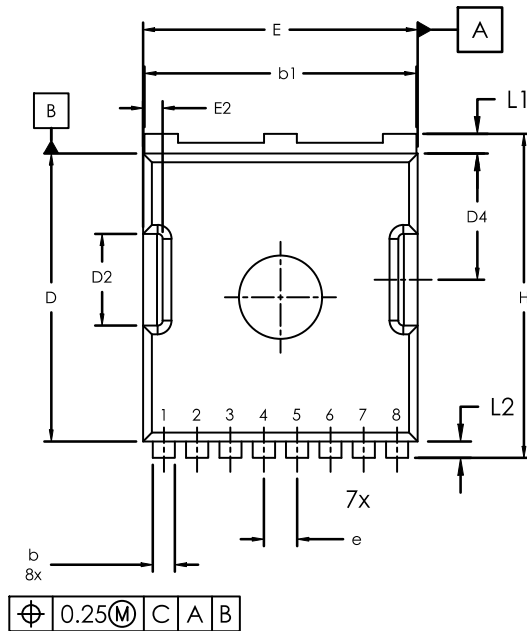
REVISION HISTORY

Revision	Description of Changes	Date
D	Acquired the original Qorvo JFET Division Data Sheet and updated the main document title to comply with <b>onsemi</b> standards for SiC products.	1/15/2025
4	Converted the Data Sheet to <b>onsemi</b> format.	5/28/2025

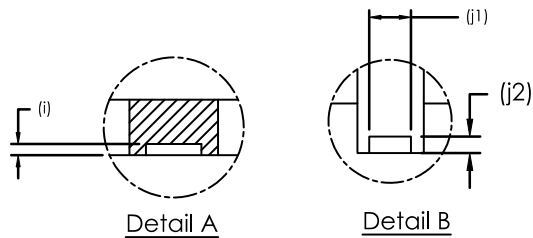
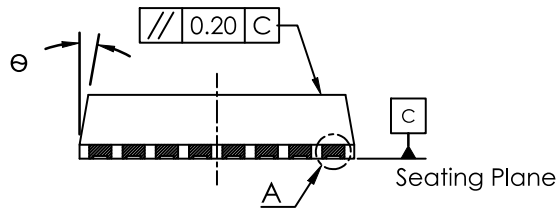


**H-PDSO-F8 9.90x10.38x2.30, 1.20P**  
**CASE 740AA**  
**ISSUE B**

DATE 24 JUN 2025



⌀ 0.25 (M) C A B



Note:

1. Dimensioning and tolerancing as per ASME Y14.5 - 2018
2. Controlling dimension : millimeters
3. Dimensions does not include Burrs and Mold Flashes

TO-LL			
SYMBOL	Value		
	Min	Nom	Max
A	2.15	2.30	2.45
A1	1.80 REF		
b	0.65	0.80	0.90
b1	9.65	9.80	9.95
b3	1.10	1.20	1.30
c	0.40	0.50	0.60
D	10.18	10.38	10.58
D1	10.88	11.08	11.28
D2	3.15	3.30	3.45
D4	4.40	4.55	4.70
E	9.70	9.90	10.10
E1	7.95	8.10	8.25
E2	0.60	0.70	0.80
e	1.20 BSC		
H	11.48	11.68	11.88
H1	6.80	6.95	7.10
i	0.10 REF		
j1	0.46 REF		
j2	0.20 REF		
K	2.80 REF		
L	1.40	1.90	2.10
L1	0.50	0.70	0.90
L2	0.48	0.60	0.72
L3	0.30	0.70	0.80
Q	6.80 REF		
R	3.00	3.10	3.20
θ	10°		

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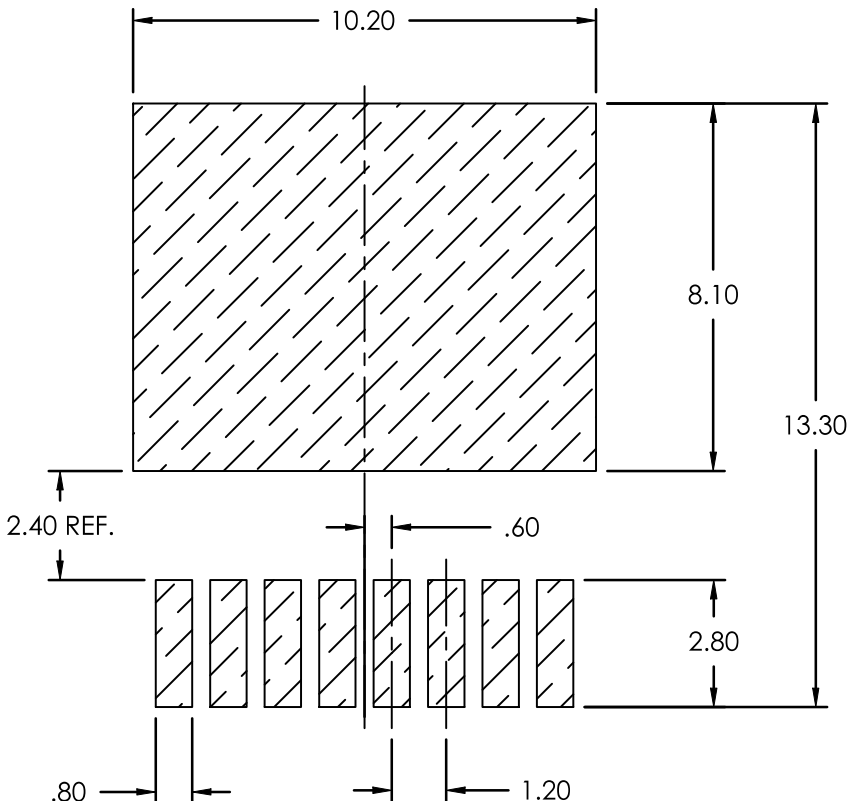
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H-PDSO-F8 9.90x10.38x2.30, 1.20P  
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RECOMMENDED PCB LAND PATTERN



NOTE: LAND PATTERN AND THROUGH HOLE DIMENSIONS SERVE ONLY AS AN INITIAL GUIDE.  
END-USER PCB DESIGN RULES AND TOLERANCES SHOULD ALWAYS PREVAIL.

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DESCRIPTION:	H-PDSO-F8 9.90x10.38x2.30, 1.20P	PAGE 2 OF 2

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