onsemi

MOSFET – N-Channel, POWERTRENCH[®]

150 V, 79 A, 16 m Ω

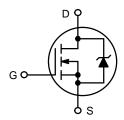
FDB2532-F085

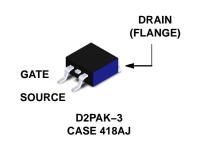
Features

- $R_{DS(ON)} = 14 \text{ m}\Omega$ (Typ.), $V_{GS} = 10 \text{ V}$, $I_D = 33 \text{ A}$
- Q_g (tot) = 82 nC (Typ.), V_{GS} = 10 V
- Low Miller Charge
- Low Q_{RR} Body Diode
- UIS Capability (Single Pulse and Repetitive Pulse)
- AEC–Q101 Qualified and PPAP Capable
- These Devices are Pb-Free and are RoHS Compliant

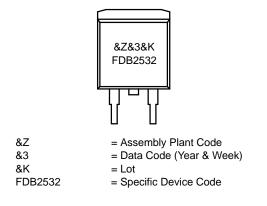
Applications

- DC/DC converters and Off-Line UPS
- Distributed Power Architectures and VRMs
- Primary Switch for 24 V and 48 V Systems
- High Voltage Synchronous Rectifier
- Direct Injection / Diesel Injection Systems
- 42 V Automotive Load Control
- Electronic Valve Train Systems
- Synchronous Rectification





MARKING DIAGRAM



ORDERING INFORMATION

See detailed ordering and shipping information on page 2 of this data sheet.

MOSFET MAXIMUM RATINGS (T_C = 25° C, Unless otherwise noted)

Symbol		Parameter	Value	Unit
V _{DSS}	Drain to Source Voltage		150	V
V _{GS}	Gate to Source Voltage		±20	V
I _D	Drain Current	– Continuous (T _C = 25°C, V _{GS} = 10 V)	79	А
		– Continuous (T _C = 100°C, V _{GS} = 10 V)	56	1
		– Continuous (T _{amb} = 25°C, V _{GS} = 10 V, R _{θJA} = 43°C/W)	8	A
Ι _D	Drain Current	– Pulsed	Figure 4	А
E _{AS}	Single Pulse Avalanche	Energy (Note 1)	400	mJ
PD	Power Dissipation	$(T_{C} = 25^{\circ}C)$	310	W
		– Derate Above 25°C	2.07	W/°C
T _J , T _{STG}	Operating and Storage	Temperature	-55 to +175	°C

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected. 1. Starting $T_J = 25^{\circ}C$, L = 0.5 mH, $I_{AS} = 40$ A

THERMAL CHARACTERISTICS

Symbol	Parameter	Value	Unit
$R_{ extsf{ heta}JC}$	Thermal Resistance Junction to Case	0.48	°C/W
R_{\thetaJA}	Thermal Resistance Junction to Ambient, 1in ² Copper Pad Area	43	°C/W

PACKAGE MARKING AND ORDERING INFORMATION

Device Marking	Device	Package	Reel Size	Tape Width	Quantity
FDB2532	FDB2532-F085	TO-263 (D ² -PAK-3)	330 mm	24 mm	800 Units

ELECTRICAL CHARACTERISTICS (T_C = 25°C unless otherwise noted)

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit	
OFF CHARACT	OFF CHARACTERISTICS						
B _{VDSS}	Drain to Source Breakdown Voltage	$I_D = 250 \ \mu A, \ V_{GS} = 0 \ V$	150			V	
I _{DSS}	Zero Gate Voltage Drain Current	$V_{DS} = 120 \text{ V}, V_{GS} = 0 \text{ V}$			1	μA	
		$V_{DS} = 120 \text{ V}, V_{GS} = 0 \text{ V}, T_{C} = 150^{\circ}\text{C}$			250		
I _{GSS}	Gate to Source Leakage Current	$V_{GS} = \pm 20 V$			±100	nA	

ON CHARACTERISTICS

V _{GS(TH)}	Gate to Source Threshold Voltage	$V_{GS} = V_{DS}, I_D = 250 \ \mu A$	2.0		4.0	V
R _{DS(ON)}	Drain to Source On Resistance	$I_D = 33 \text{ A}, V_{GS} = 10 \text{ V}$		0.014	0.016	Ω
		$I_D = 16 \text{ A}, V_{GS} = 6 \text{ V}$		0.016	0.024	
		I_D = 33 A, V_{GS} = 10 V, T_C = 175 $^\circ C$		0.040	0.048	

DYNAMIC CHARACTERISTICS

C _{iss}	Input Capacitance	V_{DS} = 25 V, V_{GS} = 0 V, f = 1 MHz	5870		pF
C _{oss}	Output Capacitance		615		pF
C _{rss}	Reverse Transfer Capacitance		135		pF
Q _{g(tot)}	Total Gate Charge at 10 V	$V_{GS} = 0 V \text{ to } 10 V,$ $V_{DD} = 75 V, I_D = 33 A, I_g = 1.0 \text{ mA}$	82	107	nC
Q _{g(th)}	Threshold Gate Charge	V_{GS} = 0 V to 2 V, V_{DD} = 75 V, I_{D} = 33 A, I_{g} = 1.0 mA	11	14	nC
Q _{gs}	Gate to Source Gate Charge	$V_{DD} = 75 \text{ V}, \text{ I}_{D} = 33 \text{ A}, \text{ I}_{g} = 1.0 \text{ mA}$	23		nC
Q _{gs2}	Gate Charge Threshold to Plateau]	13		nC
Q _{gd}	Gate to Drain "Miller" Charge		19		nC

RESISTIVE SWITCHING CHARACTERISTICS (V_{GS} = 10 V)

t _{ON}	Turn-On Time	V_{DD} = 75 V, I _D = 33 A, V _{GS} = 10 V, R _{GS} = 3.6 Ω		69	ns
t _{d(ON)}	Turn-On Delay Time	$v_{GS} = 10 v, R_{GS} = 3.6 \Omega$	16		ns
t _r	Rise Time		30		ns
t _{d(OFF)}	Turn-Off Delay Time		39		ns
t _f	Fall Time		17		ns
t _{OFF}	Turn-Off Time			84	ns

DRAIN-SOURCE DIODE CHARACTERISTICS

V _{SD}	Source to Drain Diode Voltage	I _{SD} = 33 A		1.25	V
		I _{SD} = 16 A		1	V
t _{rr}	Reverse Recovery Time	I_{SD} = 33 A, dI _{SD} /dt = 100 A/µs		105	ns
Q _{RR}	Reverse Recovery Charge	I_{SD} = 33 A, dI _{SD} /dt = 100 A/µs		327	nC

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions. 1. Pulse Width = 100 s

TYPICAL CHARACTERISTICS

 $T_{C} = 25^{\circ}C$ UNLESS OTHERWISE NOTED

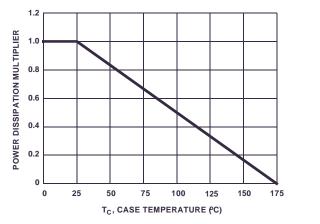


Figure 1. Normalized Power Dissipation vs. Ambient Temperature

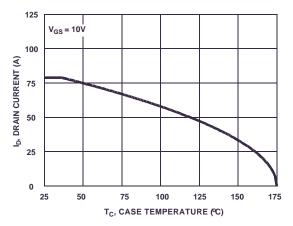


Figure 2. Maximum Continuous Drain Current vs Case Temperature

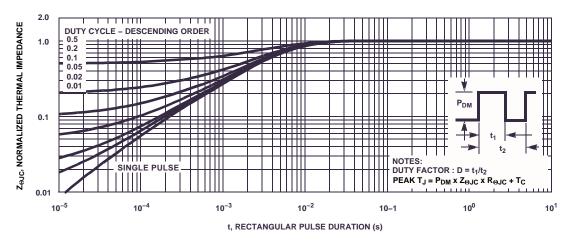


Figure 3. Normalized Maximum Transient Thermal Impedance

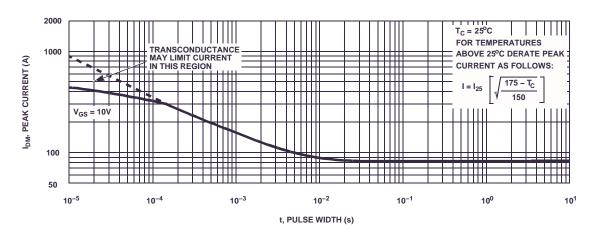


Figure 4. Peak Current Capability

TYPICAL CHARACTERISTICS (CONTINUED)

 $T_C = 25^{\circ}C$ UNLESS OTHERWISE NOTED

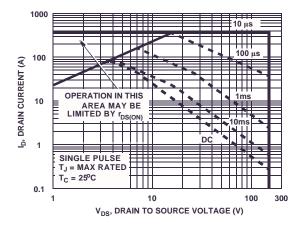


Figure 5. Forward Bias Safe Operating Area

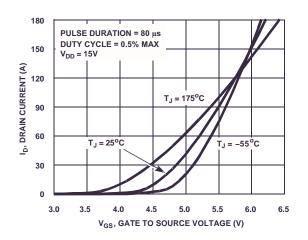


Figure 7. Transfer Characteristics

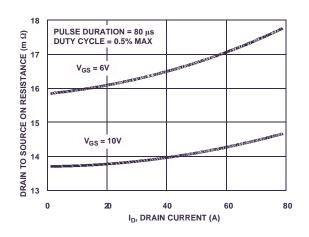


Figure 9. Drain to Source On Resistance vs Drain Current

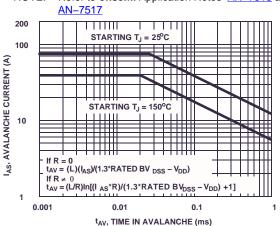


Figure 6. Unclamped Inductive Switching Capability

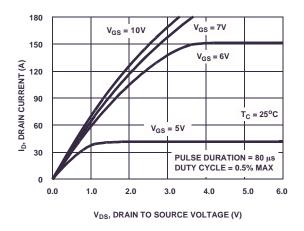
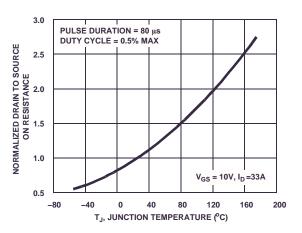


Figure 8. Saturation Characteristics





TYPICAL CHARACTERISTICS (CONTINUED) $T_C = 25^{\circ}C$ UNLESS OTHERWISE NOTED

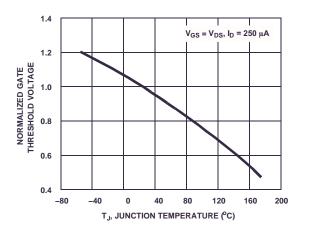


Figure 11. Normalized Gate Threshold Voltage vs. Junction Temperature

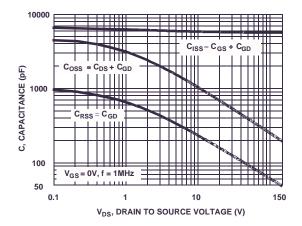


Figure 13. Capacitance vs. Drain to Source Voltage

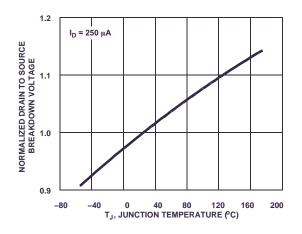


Figure 12. Normalized Drain to Source Breakdown Voltage vs Junction Temperature

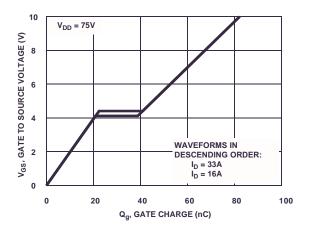


Figure 14. Gate Charge Waveforms for Constant Gate Currents

TEST CIRCUITS WAVEFORMS

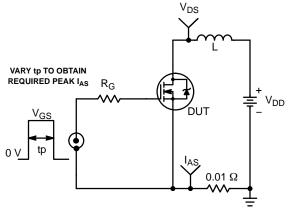


Figure 15. Unclamped Energy Test Circuit

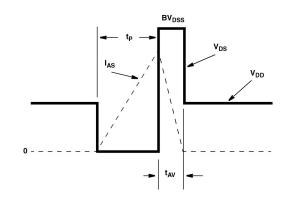


Figure 16. Unclamped Energy Waveforms

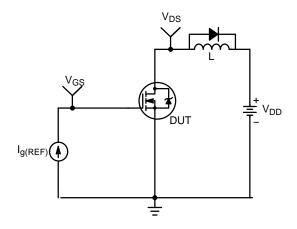


Figure 17. Gate Charge Test Circuit

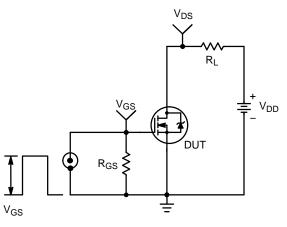


Figure 19. Switching Time Test Circuit

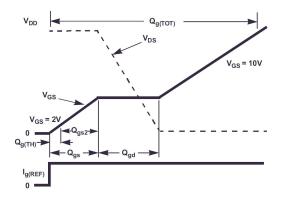
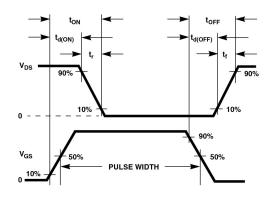
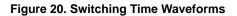


Figure 18. Gate Charge Waveforms





THERMAL RESISTANCE VS. MOUNTING PAD AREA

The maximum rated junction temperature, T_{JM} , and the thermal resistance of the heat dissipating path determines the maximum allowable device power dissipation, P_{DM} , in an application. Therefore the application's ambient temperature, T_A (°C), and thermal resistance $R_{\theta JA}$ (°C/W) must be reviewed to ensure that T_{JM} is never exceeded. Equation 1 mathematically represents the relationship and serves as the basis for establishing the rating of the part.

$$P_{DM} = \frac{(T_{JM} - T_A)}{R_{\Theta JA}}$$
 (eq. 1)

In using surface mount devices such as the TO–263 (D^2 –PAK–3) package, the environment in which it is applied will have a significant influence on the part's current and maximum power dissipation ratings. Precise determination of P_{DM} is complex and influenced by many factors:

- 1. Mounting pad area onto which the device is attached and whether there is copper on one side or both sides of the board.
- 2. The number of copper layers and the thickness of the board.
- 3. The use of external heat sinks.
- 4. The use of thermal vias.
- 5. Air flow and board orientation.
- 6. For non steady state applications, the pulse width, the duty cycle and the transient thermal response of the part, the board and the environment they are in.

onsemi provides thermal information to assist the designer's preliminary application evaluation. Figure 21 defines the $R_{\theta JA}$ for the device as a function of the top copper (component side) area. This is for a horizontally positioned FR-4 board with 10z copper after 1 000 seconds of steady state power with no air flow. This graph provides the necessary information for calculation of the steady state junction temperature or power dissipation. Pulse applications can be evaluated using the **onsemi** device Spice thermal model or manually utilizing the normalized maximum transient thermal impedance curve.

Thermal resistances corresponding to other copper areas can be obtained from Figure 21 or by calculation using Equation 2 or 3. Equation 2 is used for copper area defined in inches square and equation 3 is for area in centimeter square. The area, in square inches or square centimeters is the top copper area including the gate and source pads.

$$R_{\Theta JA} = 26.51 + \frac{19.84}{(0.262 + Area)}$$
 (eq. 2)

Area in Inches Squared.

$$R_{\Theta JA} = 26.51 + \frac{128}{(1.69 + Area)}$$
 (eq. 3)

Area in Centimeters Squared.

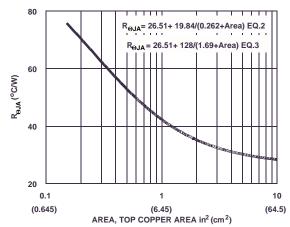


Figure 21. Thermal Resistance vs. Mounting Pad Area

PSPICE ELECTRICAL MODEL

.SUBCKT FDB2532 2 1 3 ; rev April 2002 CA 12 8 1.4e–9 CB 15 14 1.6e–9 CIN 6 8 5.61e–9

Dbody 7 5 DbodyMOD Dbreak 5 11 DbreakMOD Dplcap 10 5 DplcapMOD

Ebreak 11 7 17 18 159 Eds 14 8 5 8 1 Egs 13 8 6 8 1 Esg 6 10 6 8 1 Evthres 6 21 19 8 1 Evtemp 20 6 18 22 1

It 8 17 1

Lgate 1 9 9.56e–9 Ldrain 2 5 1.0e–9 Lsource 3 7 7.71e–9

RLgate 1 9 95.6 RLdrain 2 5 10 RLsource 3 7 77.1

Mmed 16 6 8 8 MmedMOD Mstro 16 6 8 8 MstroMOD Mweak 16 21 8 8 MweakMOD

Rbreak 17 18 RbreakMOD 1 Rdrain 50 16 RdrainMOD 9.6e–3 Rgate 9 20 1.01 RSLC1 5 51 RSLCMOD 1.0e–6 RSLC2 5 50 1.0e3 Rsource 8 7 RsourceMOD 3.0e–3 Rvthres 22 8 RvthresMOD 1 Rvtemp 18 19 RvtempMOD 1 S1a 6 12 13 8 S1AMOD S1b 13 12 13 8 S1BMOD S2a 6 15 14 13 S2AMOD S2b 13 15 14 13 S2BMOD

Vbat 22 19 DC 1

ESLC 51 50 VALUE={(V(5,51)/ABS(V(5,51)))*(PWR(V(5,51)/(1e-6*190),3))}

.MODEL DbodyMOD D (IS=6.0E-11 N=1.09 RS=2.3e-3 TRS1=3.0e-3 TRS2=1.0e-6 + CJO=3.9e-9 M=0.65 TT=4.8e-8 XTI=4.2) .MODEL DbreakMOD D (RS=0.17 TRS1=3.0e-3 TRS2=-8.9e-6) .MODEL DplcapMOD D (CJO=1.0e-9 IS=1.0e-30 N=10 M=0.6)

.MODEL MmedMOD NMOS (VTO=3.55 KP=10 IS=1e-30 N=10 TOX=1 L=1u W=1u RG=1.01) .MODEL MstroMOD NMOS (VTO=4.2 KP=145 IS=1e-30 N=10 TOX=1 L=1u W=1u) .MODEL MweakMOD NMOS (VTO=2.9 KP=0.05 IS=1e-30 N=10 TOX=1 L=1u W=1u RG=10.1 RS=0.1)

.MODEL RbreakMOD RES (TC1=1.1e-3 TC2=-9.0e-7) .MODEL RdrainMOD RES (TC1=9.0e-3 TC2=3.5e-5) .MODEL RSLCMOD RES (TC1=3.4e-3 TC2=1.5e-6) .MODEL RsourceMOD RES (TC1=4.0e-3 TC2=1.0e-6)

.MODEL RvthresMOD RES (TC1=-4.1e-3 TC2=-1.4e-5)

.MODEL RvtempMOD RES (TC1=-4.0e-3 TC2=3.5e-6)

.MODEL S1AMOD VSWITCH (RON=1e-5 ROFF=0.1 VON=-6.0 VOFF=-4.0)

.MODEL S1BMOD VSWITCH (RON=1e-5 ROFF=0.1 VON=-4.0 VOFF=-6.0)

.MODEL S2AMOD VSWITCH (RON=1e-5 ROFF=0.1 VON=-1.4 VOFF=1.0)

.MODEL S2BMOD VSWITCH (RON=1e-5 ROFF=0.1 VON=1.0 VOFF=-1.4)

NOTE: For further discussion of the PSPICE model, consult A New PSPICE Sub–Circuit for the Power MOSFET Featuring Global Temperature Options; IEEE Power Electronics Specialist Conference Records, 1991, written by William J. Hepp and C. Frank Wheatley.

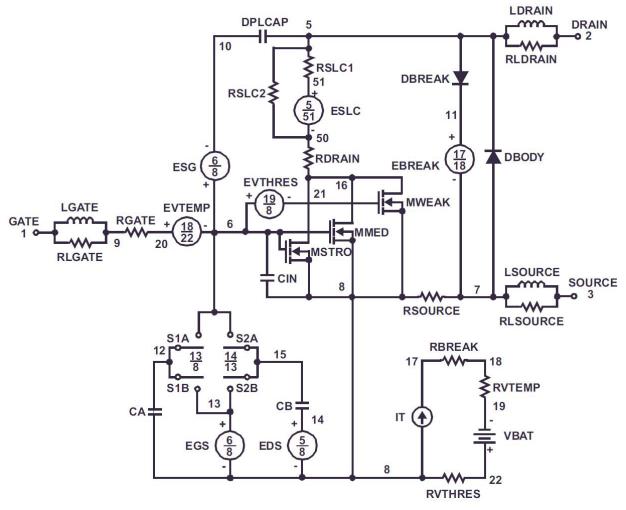


Figure 22. PSPICE Electrical Model

SABER ELECTRICAL MODEL

```
REV April 2002
ttemplate FDB2532 n2,n1,n3
electrical n2,n1,n3
var i iscl
dp..model \ dbodymod = (isl=6.0e-11,nl=1.09,rs=2.3e-3,trs1=3.0e-3,trs2=1.0e-6,cjo=3.9e-9,m=0.65,tt=4.8e-8,xti=4.2)
dp..model dbreakmod = (rs=0.17, trs1=3.0e-3, trs2=-8.9e-6)
dp..model dplcapmod = (cjo=1.0e-9,isl=10.0e-30,nl=10,m=0.6)
m..model mmedmod = (type=_n, vto=3.55, kp=10, is=1e-30, tox=1)
m..model mstrongmod = (type= n,vto=4.2,kp=145,is=1e-30,tox=1)
m..model mweakmod = (type= n,vto=2.9,kp=0.05,is=1e-30,tox=1,rs=0.1)
sw vcsp..model s1amod = (ron=1e-5, roff=0.1, von=-6.0, voff=-4.0)
sw vcsp..model s1bmod = (ron=1e-5, roff=0.1, von=-4.0, voff=-6.0)
sw_vcsp..model s2amod = (ron=1e-5,roff=0.1,von=-1.4,voff=1.0)
sw vcsp..model s2bmod = (ron=1e-5,roff=0.1,von=1.0,voff=-1.4)
c.ca n12 n8 = 1.4e-9
c.cb n15 n14 = 1.6e-9
c.cin n6 n8 = 5.61e-9
dp.dbody n7 n5 = model = dbodymod
dp.dbreak n5 n11 = model=dbreakmod
dp.dplcap n10 n5 = model = dplcapmod
spe.ebreak n11 n7 n17 n18 = 159
spe.eds n14 n8 n5 n8 = 1
spe.egs n13 n8 n6 n8 = 1
spe.esg n6 n10 n6 n8 = 1
spe.evthres n6 n21 n19 n8 = 1
spe.evtemp n20 n6 n18 n22 = 1
i.it n8 n17 = 1
1.1gate n1 n9 = 9.56e-9
1.1 drain n2 n5 = 1.0 e-9
1.1source n3 n7 = 7.71e-9
res.rlgate n1 n9 = 95.6
res.rldrain n2 n5 = 10
res.rlsource n3 n7 = 77.1
m.mmed n16 n6 n8 n8 = model=mmedmod, l=1u, w=1u
m.mstrong n16 n6 n8 n8 = model=mstrongmod, l=1u, w=1u
m.mweak n16 n21 n8 n8 = model=mweakmod, l=1u, w=1u
res.rbreak n17 n18 = 1, tc1=1.1e-3.tc2=-9.0e-7
res.rdrain n50 n16 = 9.6e-3, tc1=9.0e-3,tc2=3.5e-5
res.rgate n9 n20 = 1.01
res.rslc1 n5 n51 = 1.0e-6, tc1=3.4e-3,tc2=1.5e-6
res.rslc2 n5 n50 = 1.0e3
res.rsource n8 n7 = 3.0e-3, tc1=4.0e-3, tc2=1.0e-6
res.rvthres n22 n8 = 1, tc1 = -4.1e - 3, tc2 = -1.4e - 5
res.rvtemp n18 n19 = 1, tc1=-4.0e-3,tc2=3.5e-6
sw_vcsp.s1a n6 n12 n13 n8 = model=s1amod
sw vcsp.s1b n13 n12 n13 n8 = model=s1bmod
sw_vcsp.s2a n6 n15 n14 n13 = model=s2amod
```

```
sw_vcsp.s2b n13 n15 n14 n13 = model=s2bmod
```

```
v.vbat n22 n19 = dc=1
equations {
i (n51->n50) +=iscl
iscl: v(n51,n50) = ((v(n5,n51)/(1e-9+abs(v(n5,n51))))*((abs(v(n5,n51)*1e6/190))** 3))
}
}
```

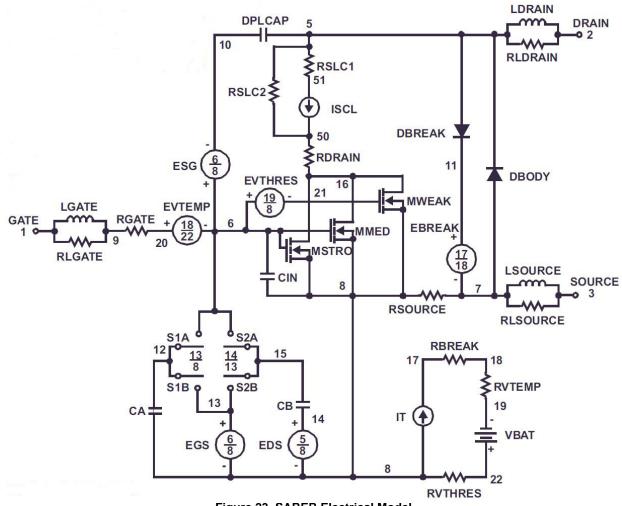


Figure 23. SABER Electrical Model

SPICE THERMAL MODEL

REV 26 February 2002

FDB2532

CTHERM1 TH 6 7.5e-3 CTHERM2 6 5 8.0e-3 CTHERM3 5 4 9.0e-3 CTHERM4 4 3 2.4e-2 CTHERM5 3 2 3.4e-2 CTHERM6 2 TL 6.5e-2

RTHERM1 TH 6 3.1e-4 RTHERM2 6 5 2.5e-3 RTHERM3 5 4 2.0e-2 RTHERM4 4 3 8.0e-2 RTHERM5 3 2 1.2e-1 RTHERM6 2 TL 1.3e-1

SABER THERMAL MODEL

SABER thermal model FDB2532 template thermal_model th tl thermal_c th, tl { ctherm.ctherm1 th 6 =7.5e-3 ctherm.ctherm2 6 5 =8.0e-3 ctherm.ctherm3 5 4 =9.0e-3 ctherm.ctherm4 4 3 =2.4e-2 ctherm.ctherm5 3 2 =3.4e-2 ctherm.ctherm6 2 tl =6.5e-2 rrtherm.rtherm1 th 6 =3.1e-4 rtherm.rtherm1 th 6 =3.1e-4 rtherm.rtherm3 5 4 =2.0e-2 rtherm.rtherm3 5 4 =2.0e-2 rtherm.rtherm4 4 3 =8.0e-2 rtherm.rtherm5 3 2 =1.2e-1 rtherm.rtherm5 2 tl =1.3e-1

}

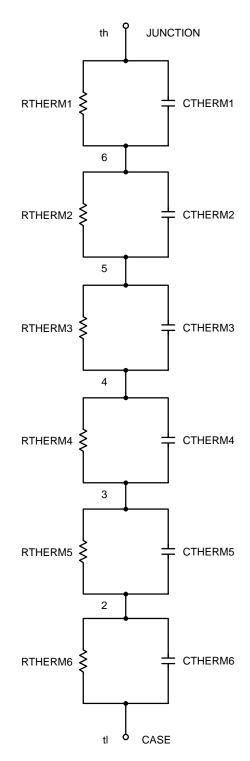
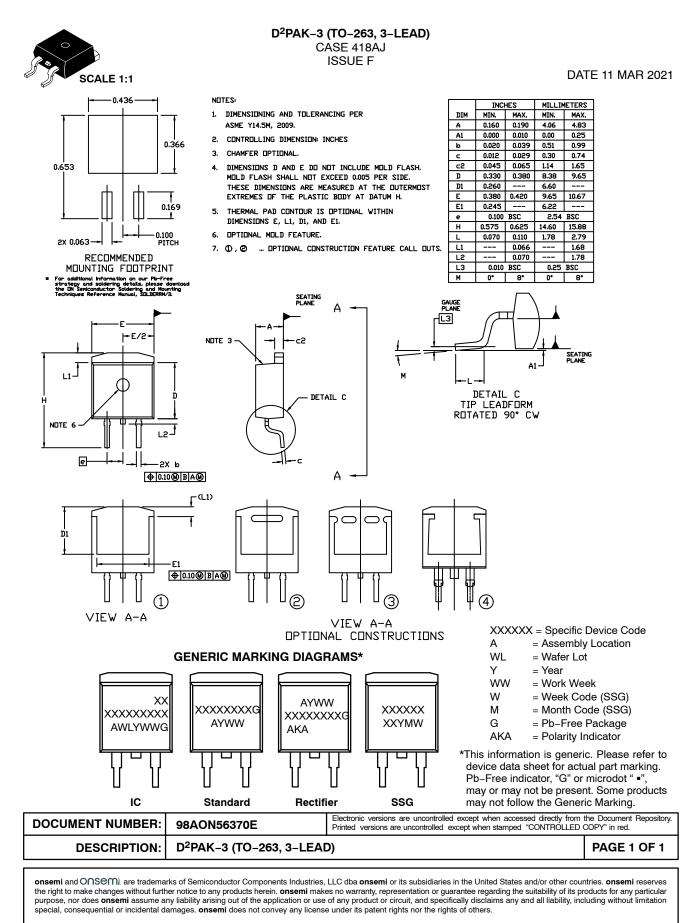


Figure 24. Thermal Model

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