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Dual Output Driver

Description

The CS3706 integrated circuit provides an interface between low-level TTL inputs and high-power switching devices such as power MOSFETs. A typical application is single-ended PWM control to push-pull power control conversion.

The primary function of this device is to convert a bipolar single-ended low current digital input to a pair of totem pole outputs which can source or sink up to 1.5A. An internal flip-flop, driven by double-pulse suppression logic, can be enabled to provide single-ended to push-pull conversion. With the flip-flop disabled, the outputs work in parallel for 3.0A capability.

Protection functions are also included for pulse-by-pulse current limiting, automatic deadband control and thermal shutdown.

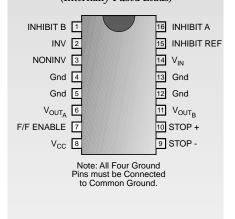
Features

- Dual 1.5A Totem Pole Outputs
- 40nsec Rise and Fall into 1000pF
- Parallel or Push-Pull Operation
- Single-Ended to Push-Pull Conversion
- High-Speed Power MOSFET Compatible
- Low Cross-Conduction Current Spike
- Analog Latched Shutdown
- Internal Deadband Inhibit Circuit
- Low Quiescent Current
- 5V to 40V Operation
- Thermal Shutdown Protection

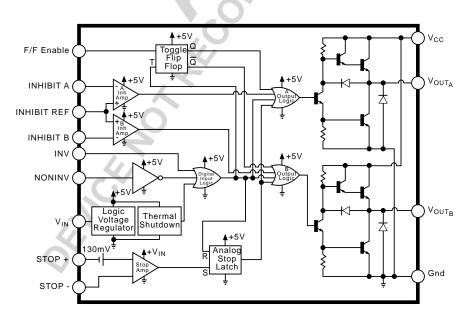
Package Options

16 Lead PDIP

(Internally Fused Leads)



Block Diagram





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Absolute Maximum Ratings

Logic Supply Voltage (V _{IN})	40.0V
Output Supply Voltage (V _{CC})	40.0V
Output Current (each output, source, or sink)	
Steady State	±500mA
Peak Transient for Less Than 100μs	
Capacitive Discharge Energy	20.0µJ
Digital Inputs (INV, NONINV)	5.5V
Analog Inputs (STOP +, STOP -)	V _{IN}
Inhibit Inputs (INHIBIT A, INHIBIT B, INHIBIT REF)	
Operating Temperature Range	0 to 70°C
Storage Temperature Range	
Lead Temperature Soldering	
Wave Solder (through hole styles only)	10 sec. max, 260°C peak

Notes: All voltages are with respect to the four ground pins which must be connected together. All currents are positive into, negative out of the specified terminal.

Electrical Characteristics: These specifications apply over the operating temperature range of the IC. $(V_{\rm IN}=V_{\rm CC}=20{\rm V}, {\rm Pins}~4, 5, 12~\&13=0{\rm V}; {\rm unless}~{\rm otherwise}~{\rm stated.})$					
PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _{IN} Supply Current	$\begin{split} V_{IN} &= 40V, V_{CC} = 20V, INV = 0V, \\ Unused pins = open. \end{split}$		8	12	mA
V _{CC} Supply Current	V_{IN} = 20V, V_{CC} = 40V, Outputs low		3	5	mA
V _{CC} Leakage Current	$V_{IN} = 0V$, $V_{CC} = 40V$		0.05	0.10	mA
Digital Input Low Level				0.8	V
Digital Input High Level		2.2			V
Digital Input Current	$V_I = 0V$		-0.6	-1.0	mA
Digital Input Leakage	$V_I = 5V$		0.05	0.10	mA
Output High Sat., V _C -V _{OUT}	$I_{OUT} = -50 \text{mA}$			2.0	V
Output High Sat., V_{C} - V_{OUT}	at High Sat., V_C - V_{OUT} $I_{OUT} = -500 \text{mA}$			2.5	V
Output Low Sat., V _{OUT}	$I_{OUT} = 50 \text{mA}$			0.4	V
Output Low Sat., V _{OUT}	$I_{OUT} = 500 \text{mA}$			2.5	V
Inhibit Threshold	$V_{REF} = 0.5V$	0.4		0.6	V
Inhibit Threshold	$V_{REF} = 3.5V$	3.3		3.7	V
Inhibit Input Current	$V_{REF} = 0V$		-10	-20	μΑ

100

130

-10

155

125

150

-20

mV

μA °C

°C

 $V_{CM} = 0V$ to 15 V

Turn on

Turn off

 $V_I = 0V$, $V_{CM} = 15V$

Analog Threshold

Thermal Shutdown

Thermal Shutdown

Analog Input Bias Current

ns

Typical Switching Characteristics: $(V_{IN} = V_{CC} = 20V, T_A = 25^{\circ}C.$ Delays measured 50% in to 50% out.)					
PARAMETER	TEST CONDITIONS		OUTPUT C _L =		UNIT
From Inv. Input to Output: Rise Time Delay 10% to 90% Rise Fall Time Delay 90% to 10% Fall		open 110 20 80 25	1.0 130 40 90 30	2.2 140 60 110 50	nF ns ns ns
From N.I. Input to Output: Rise Time Delay 10% to 90% Rise Fall Time Delay 90% to 10% Fall		120 20 100 25	130 40 120 30	140 60 130 50	ns ns ns
V _C Cross-Conduction Current Spike Duration	Output Rise Output Fall	25 0			ns ns
Inhibit Delay	Inhibit Ref. = 1V Inhibit = 0.5 to 1.5V	250			ns

180

Stop (+) Ref. = 0 Stop (-) Input = 0 to 0.5V

Analog Shutdown Delay

		Package Pin Description
PACKAGE PIN #	PIN SYMBOL	FUNCTION
16L PDIP (Internally Fused Leads)		
1	INHIBIT B	Control pin for deadband control on Channel B.
2	INV	Inverting input for output drivers.
3	NONINV	Noninverting input for output drivers.
4	Gnd	Ground.
5	Gnd	Ground.
6	$V_{OUT(A)}$	Channel A output.
7	F/F ENABLE	Controls the phase of the two outputs. F/F ENABLE = Gnd Out of phase. F/F ENABLE = floating In phase.
8	V_{CC}	Supply voltage (5V to 40V) for output drivers.
9	STOP -	Inverting input for stop latch comparator.
10	STOP +	Noninverting input for stop latch comparator.
11	$V_{\mathrm{OUT(B)}}$	Channel B output.
12	Gnd	Ground.
13	Gnd	Ground.
14	V_{IN}	Supply voltage (5V to 40V) for IC (except output driver).
15	INHIBIT REF	Reference input for deadband control.
16	INHIBIT A	Control pin for deadband control on channel A.

Outputs

The totem-pole outputs have been designed to minimize cross-conduction current spikes while maximizing fast, high-current rise and fall times. Current limiting can be done externally either at the outputs or at the common V_{CC} pin. The output diodes included have slow recovery and should be shunted with high-speed external diodes when driving high-frequency inductive loads.

Flip/Flop

Grounding F/F Enable activates the internal flip-flop to alternate the two outputs. With pin open, the two outputs operate simultaneously and can be paralleled for higher current operation. Since the flip-flop is triggered by the digital input, an off-time of at least 200nsec. must be provided to allow the flip/flop to change states. Note that the circuit logic is configured such that the "OFF" state is defined as the outputs low.

Digital Inputs

With both an inverting and non-inverting input available, either active-high or active-low signals may be accepted. These are true TTL compatible inputs—the threshold is approximately 1.2V with no hysteresis; and external pullup resistors are not required.

Inhibit Circuit

Although it may have other uses, this circuit is included to eliminate the need for deadband control when driving relatively slow bipolar power transistors. A diode from each inhibit input to the opposite power switch collector will keep one output from turning on until the other has turned-off. The threshold is determined by the voltage on INHIBIT REF which can be set from 0.5 to 3.5 V. When this circuit is not used, ground INHIBIT REF and leave INHIBIT A&B open.

Analog Shutdown

This circuit is included to get a latched shutdown as close to the outputs as possible, from a time standpoint. With an internal 130mV threshold, this comparator has a common-mode range from ground to $(V_{\rm IN}$ - 3V). When not used, both inputs should be grounded. The time required for this circuit to latch is inversely proportional to the amount of overdrive but reaches a minimum of 180nsec. As with the flip-flop, an input off-time of at least 200nsec is required to reset the latch between pulses.

Supply Voltage

With an internal 5V regulator, this circuit is optimized for use with a 7 to 40V supply, however, with some slight response time degradation, it can also be driven from 5V. When V_{IN} is low, the entire circuit is disabled and no current is drawn from $V_{CC}.$ When combined with a CS384X PWM, the Driver Bias switch can be used to supply V_{IN} to the CS3706. V_{IN} switching should be fast as undefined operation of the outputs may occur with V_{IN} less than 5V.

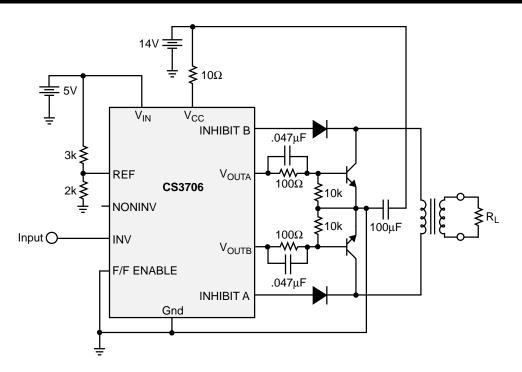
Thermal Considerations

Should the chip temperature reach approximately 155° C, a parallel, non-inverting input is activated driving both outputs to the low state.

Truth Table

INV.	N.I.	OUT
Н	Н	L
L	Н	Н
Н	L	L
L	L	L

 $\frac{OUT}{OUT} = \overline{INV} \text{ and } N.I.$ $\overline{OUT} = INV \text{ or } \overline{N.I.}$



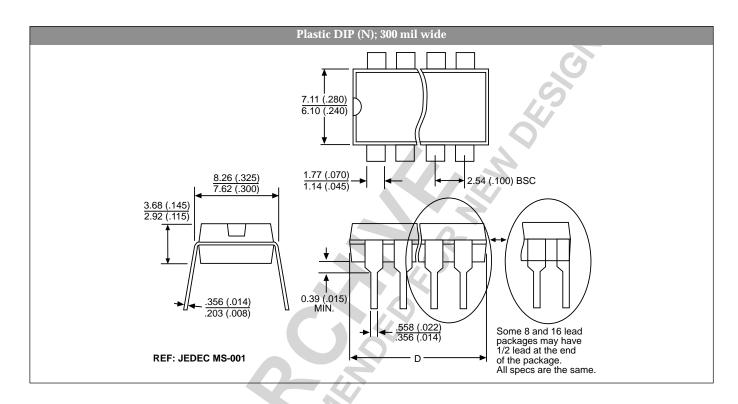
Package Specification

PACKAGE DIMENSIONS IN mm (INCHES)

	D			
Lead Count	Met	tric	Eng	glish
	Max	Min	Max	Min
16L PDIP (Internally Fused Leads)	19.69	18.67	.775	.735

PACKAGE THERMAL DATA

Thermal I	Data	16 Lead PDIP (Internally Fused Leads)	
$R_{\Theta JC}$	typ	15	°C/W
$R_{\Theta JA}$	typ	50	°C/W



Ordering Information

Part Number	Description	
CS3706GNF16	16 Lead PDIP	
	(Internally Fused Leads)	

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