



# Dual Channel Temperature Sensor and Over Temperature Alarm

## ADT7483A

### FEATURES

- 1 local and 2 remote temperature sensors
- 0.25°C resolution/1°C accuracy on remote channels
- 1°C resolution/1°C accuracy on local channel
- Extended, switchable temperature measurement range  
0°C to 127°C (default) or -64°C to +191°C
- 2-wire SMBus serial interface with SMBus alert support
- Programmable over/under temperature limits
- Offset registers for system calibration
- Up to 2 overtemperature fail-safe THERM outputs
- Small 16-lead QSOP package
- 240 µA operating current, 5 µA standby current

### APPLICATIONS

- Desktop and notebook computers
- Industrial controllers
- Smart batteries
- Automotive
- Embedded systems
- Burn-in applications
- Instrumentation

### GENERAL DESCRIPTION

The ADT7483A<sup>1</sup> is a three-channel digital thermometer and under/over temperature alarm, intended for use in PCs and thermal management systems. It can measure the temperature in two remote locations, for example, the remote thermal diode in a CPU or GPU, or a discrete diode connected transistor. It can also measure its own ambient temperature. The temperature of the remote thermal diode and ambient temperature can be accurately measured to  $\pm 1^\circ\text{C}$ . The temperature measurement range defaults to 0°C to 127°C, compatible with ADM1032, but can be switched to a wider measurement range, from -64°C to +191°C.

The ADT7483A communicates over a 2-wire serial interface compatible with system management bus (SMBus) standards. The SMBus address is set by the ADD0 and ADD1 pins. As many as nine different SMBus addresses are possible.

An ALERT output signals when the on-chip or remote temperature is outside the programmed limits. The THERM output is a comparator output that allows, for example, on/off control of a cooling fan. The ALERT output can be reconfigured as a second THERM output, if required.

### FUNCTIONAL BLOCK DIAGRAM

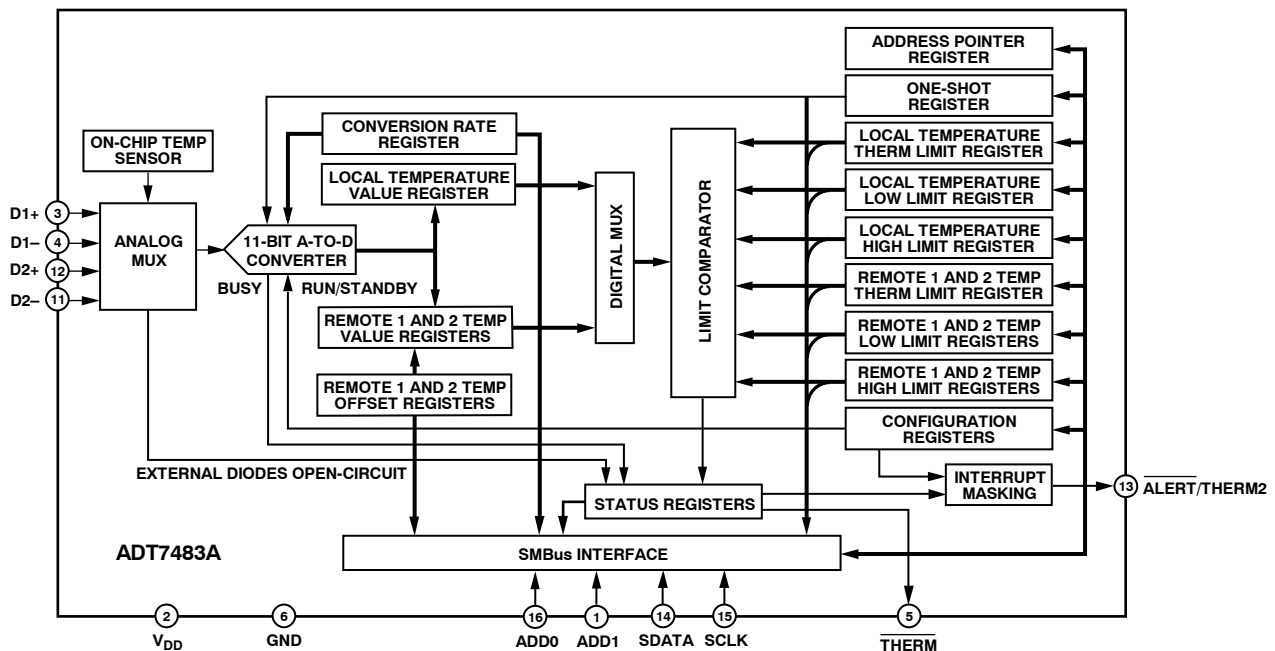


Figure 1.

<sup>1</sup> Protected by U.S. Patents 5,195,827, 5,867,012, 5,982,221, 6,097,239, 6,133,753, 6,169,442, other patents pending.

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## REVISION HISTORY

02/08—Rev 1: Conversion to ON Semiconductor

07/05—Revision 0: Initial Version

## SPECIFICATIONS

$T_A = -40^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$ ,  $V_{DD} = 3\text{ V}$  to  $3.6\text{ V}$ , unless otherwise noted.

**Table 1.**

Parameter	Min	Typ	Max	Unit	Test Conditions
<b>POWER SUPPLY</b>					
Supply Voltage, $V_{DD}$	3.0	3.30	3.6	V	
Average Operating Supply Current, $I_{DD}$		240	350	$\mu\text{A}$	0.0625 conversions/sec rate <sup>1</sup>
		5	30	$\mu\text{A}$	Standby mode
Undervoltage Lockout Threshold		2.55		V	$V_{DD}$ input, disables ADC, rising edge
Power-On-Reset Threshold	1		2.5	V	
<b>TEMPERATURE-TO-DIGITAL CONVERTER<sup>2</sup></b>					
Local Sensor Accuracy			$\pm 1$	$^{\circ}\text{C}$	$0^{\circ}\text{C} \leq T_A \leq 70^{\circ}\text{C}$
			$\pm 1.5$	$^{\circ}\text{C}$	$0^{\circ}\text{C} \leq T_A \leq 85^{\circ}\text{C}$
			$\pm 2.5$	$^{\circ}\text{C}$	$-40 \leq T_A \leq 100^{\circ}\text{C}$
Resolution		1		$^{\circ}\text{C}$	
Remote Diode Sensor Accuracy			$\pm 1$	$^{\circ}\text{C}$	$0^{\circ}\text{C} \leq T_A \leq 70^{\circ}\text{C}$ , $-55^{\circ}\text{C} \leq T_D^3 \leq 150^{\circ}\text{C}$
			$\pm 1.5$	$^{\circ}\text{C}$	$0^{\circ}\text{C} \leq T_A \leq 85^{\circ}\text{C}$ , $-55^{\circ}\text{C} \leq T_D^3 \leq 150^{\circ}\text{C}$
			$\pm 2.5$	$^{\circ}\text{C}$	$-40 \leq T_A \leq 100^{\circ}\text{C}$ , $-55^{\circ}\text{C} \leq T_D^3 \leq +150^{\circ}\text{C}$
Resolution		0.25		$^{\circ}\text{C}$	
Remote Sensor Source Current <sup>3</sup>		233		$\mu\text{A}$	High level
		14		$\mu\text{A}$	Low level
Conversion Time		73	94	ms	From stop bit to conversion complete (all channels), one-shot mode with averaging switched on
		11	14	ms	One-shot mode with averaging off, (conversion rate = 16, 32, or 64 conversions/sec)
<b>OPEN-DRAIN DIGITAL OUTPUTS (THERM, ALERT/THERM2)</b>					
Output Low Voltage, $V_{OL}$			0.4	V	$I_{OUT} = -6.0\text{ mA}$
High Level Output Leakage Current, $I_{OH}$		0.1	1	$\mu\text{A}$	$V_{OUT} = V_{DD}$
<b>SMBus INTERFACE<sup>3,4</sup></b>					
Logic Input High Voltage, $V_{IH}$ , SCLK, SDATA	2.1			V	
Logic Input Low Voltage, $V_{IL}$ , SCLK, SDATA			0.8	V	
Hysteresis		500		mV	
SDA Output Low Voltage, $V_{OL}$			0.4	V	$I_{OUT} = -6.0\text{ mA}$
Logic Input Current, $I_{IH}$ , $I_{IL}$	-1		+1	$\mu\text{A}$	
SMBus Input Capacitance, SCLK, SDATA		5		pF	
SMBus Clock Frequency			400	kHz	
SMBus Timeout <sup>5</sup>		25	32	ms	User programmable
SCLK Falling Edge to SDATA Valid Time			1	$\mu\text{s}$	Master clocking in data

<sup>1</sup> See Table 11 for information on other conversion rates.

<sup>2</sup> Temperature accuracy guaranteed with averaging enabled.

<sup>3</sup> Guaranteed by design, but not production tested.

<sup>4</sup> See the SMBus Timing Specifications section for more information.

<sup>5</sup> Disabled by default. Instructions to enable it are in the Serial Bus Interface section.

## SMBUS TIMING SPECIFICATIONS

Table 2.

Parameter <sup>1</sup>	Limit at T <sub>MIN</sub> , T <sub>MAX</sub>	Unit	Description
f <sub>SCLK</sub>	400	kHz max	
t <sub>LOW</sub>	4.7	μs min	Clock low period, between 10% points
t <sub>HIGH</sub>	4	μs min	Clock high period, between 90% points
t <sub>R</sub>	1	μs max	Clock/data rise time
t <sub>F</sub>	300	ns max	Clock/data fall time
t <sub>SU;STA</sub>	4.7	μs min	Start condition setup time
t <sub>HD;STA</sub> <sup>2</sup>	4	μs min	Start condition hold time
t <sub>SU;DAT</sub> <sup>3</sup>	250	ns min	Data setup time
t <sub>SU;STO</sub> <sup>4</sup>	4	μs min	Stop condition setup time
t <sub>BUF</sub>	4.7	μs min	Bus free time between stop and start conditions

<sup>1</sup> Guaranteed by design, but not production tested.

<sup>2</sup> Time from 10% of SDATA to 90% of SCLK.

<sup>3</sup> Time for 10% or 90% of SDATA to 10% of SCLK.

<sup>4</sup> Time for 90% of SCLK to 10% of SDATA.

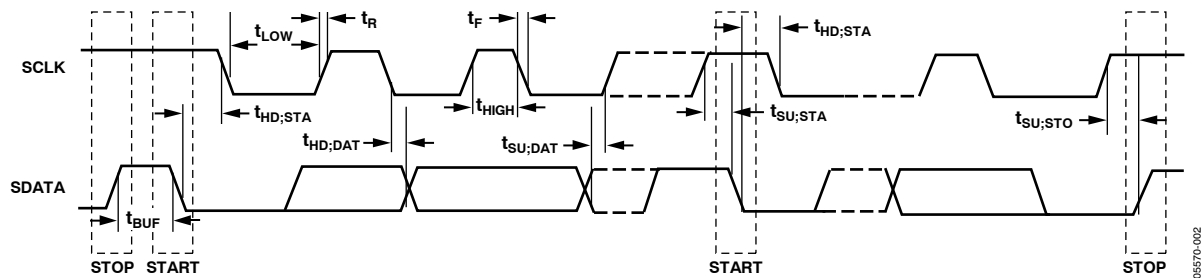


Figure 2. Serial Bus Timing

05570-002

## ABSOLUTE MAXIMUM RATINGS

Table 3.

Parameter	Rating
Positive Supply Voltage ( $V_{DD}$ ) to GND	-0.3 V, +3.6 V
D+	-0.3 V to $V_{DD} + 0.3$ V
D- to GND	-0.3 V to +0.6 V
SCLK, SDATA, ALERT, THERM	-0.3 V to +3.6 V
Input Current, SDATA, THERM	-1 mA, +50 mA
Input Current, D-	$\pm 1$ mA
ESD Rating, All Pins (Human Body Model)	1,500 V
Maximum Junction Temperature ( $T_{J\text{ MAX}}$ )	150°C
Storage Temperature Range	-65°C to +150°C
IR Reflow Peak Temperature	220°C
IR Reflow Peak Temperature Pb-Free	260°C
Lead Temperature (Soldering 10 sec)	300°C

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### THERMAL CHARACTERISTICS

16-lead QSOP package:

$$\theta_{JA} = 150^{\circ}\text{C}/\text{W}$$

$$\theta_{JC} = 38.8^{\circ}\text{C}/\text{W}$$

### ESD CAUTION

ESD (electrostatic discharge) sensitive device. Electrostatic charges as high as 4000V readily accumulate on the human body and test equipment and can discharge without detection. Although this product features proprietary ESD protection circuitry, permanent damage may occur on devices subjected to high energy electrostatic discharges. Therefore, proper ESD precautions are recommended to avoid performance degradation or loss of functionality.



## PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

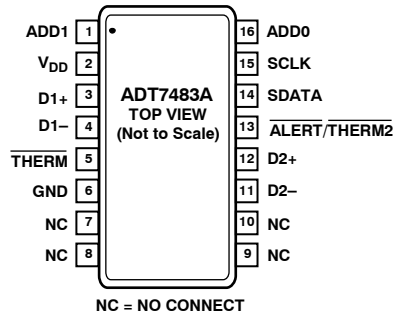


Figure 3.16-Lead QSOP Pin Configuration

Table 4. Pin Function Descriptions

Pin No.	Mnemonic	Description
1	ADD1	Address 1 Pin. Tri-state input to set the SMBus address.
2	V <sub>DD</sub>	Positive Supply, 3 V to 3.6 V.
3	D1+	Positive Connection. Connects to the first remote temperature sensor.
4	D1-	Negative Connection. Connects to the first remote temperature sensor.
5	THERM	Open-Drain Output. Turns a fan on/off, or throttles a CPU clock in the event of an overtemperature condition.
6	GND	Supply Ground Connection.
7	NC	No Connect.
8	NC	No Connect.
9	NC	No Connect.
10	NC	No Connect.
11	D2-	Negative Connection. Connects to the second remote temperature sensor.
12	D2+	Positive Connection. Connects to the second remote temperature sensor.
13	ALERT/THERM2	Open-Drain Logic Output. Used as interrupt or SMBus alert. This may also be configured as a second THERM output. Requires a pull-up resistor.
14	SDATA	Logic Input/Output, SMBus Serial Data. Open-drain output. Requires a pull-up resistor.
15	SCLK	Logic Input, SMBus Serial Clock. Requires a pull-up resistor.
16	ADD0	Address 0 Pin. Tri-state input to set the SMBus address.

# TYPICAL PERFORMANCE CHARACTERISTICS

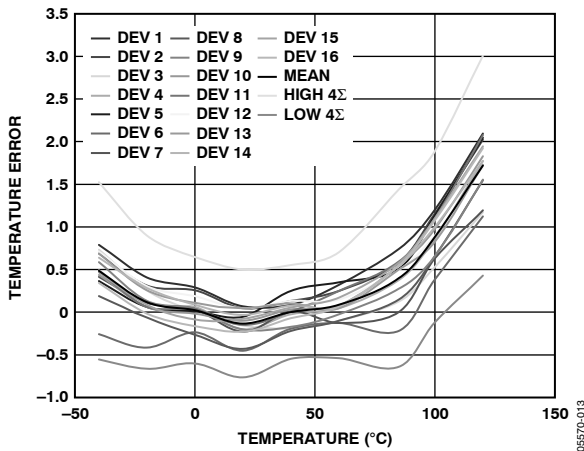


Figure 4. Local Temperature Error vs. Temperature

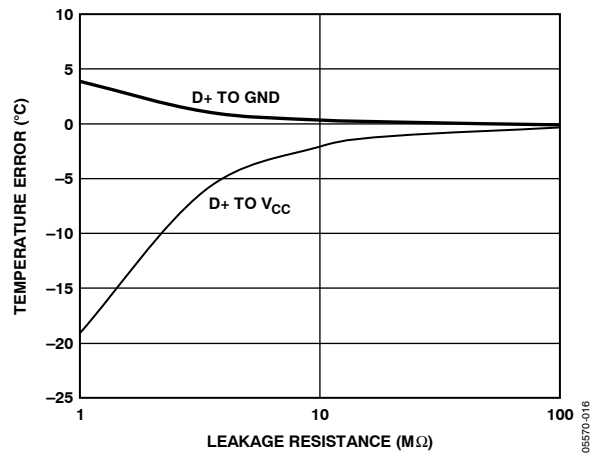


Figure 7. Temperature Error vs. D+/D- Leakage Resistance

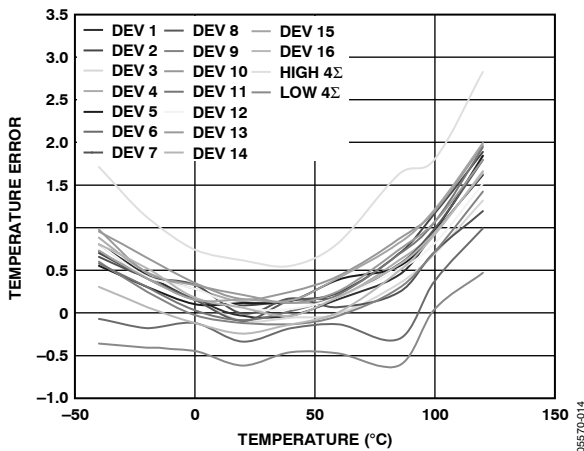


Figure 5. Remote 1 Temperature Error vs. Temperature

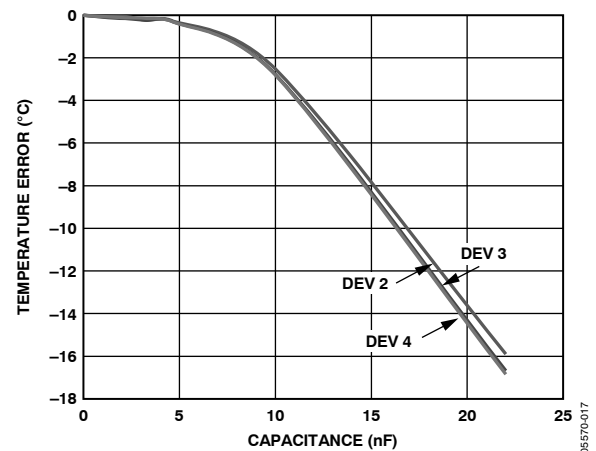


Figure 8. Temperature Error vs. D+/D- Capacitance

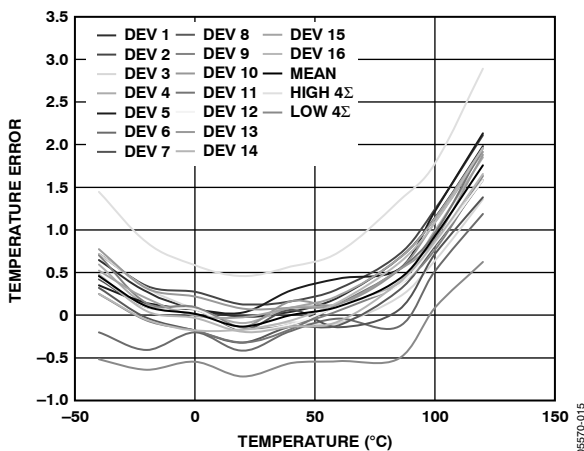


Figure 6. Remote 2 Temperature Error vs. Temperature

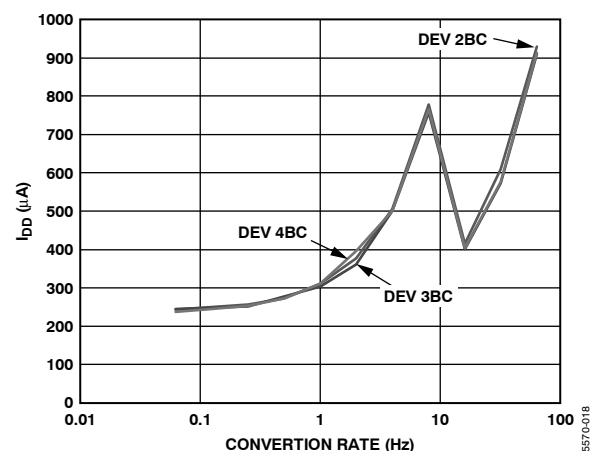


Figure 9. Operating Supply Current vs. Conversion Rate

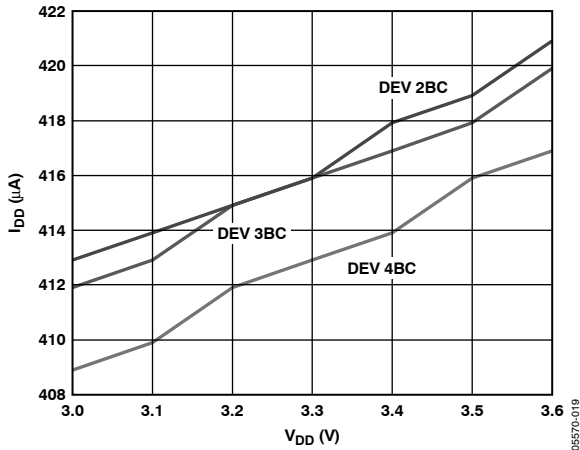


Figure 10. Operating Supply Current vs. Voltage

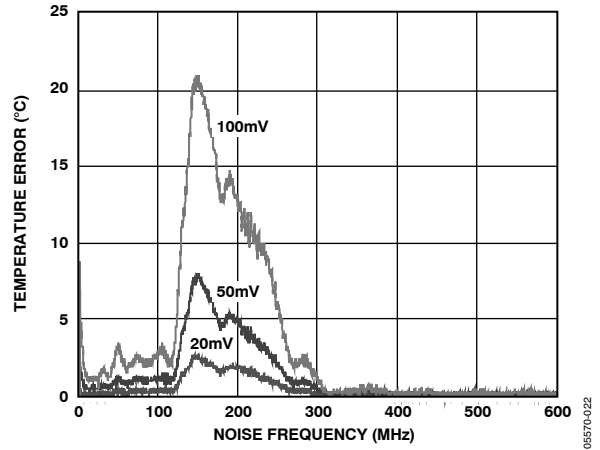


Figure 13. Temperature Error vs. Common-Mode Noise Frequency



Figure 11. Standby Supply Current vs. Voltage

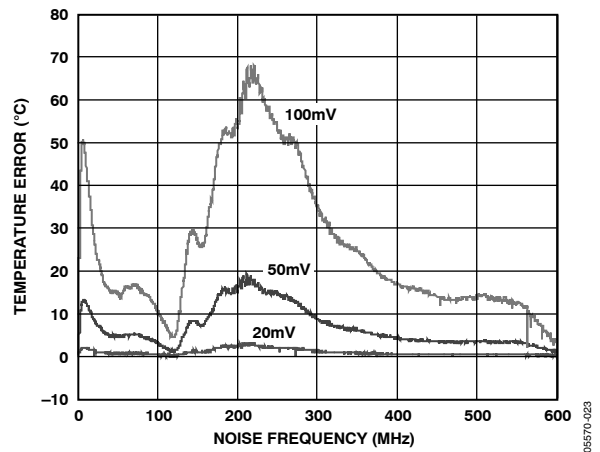


Figure 14. Temperature Error vs. Differential Mode Noise Frequency

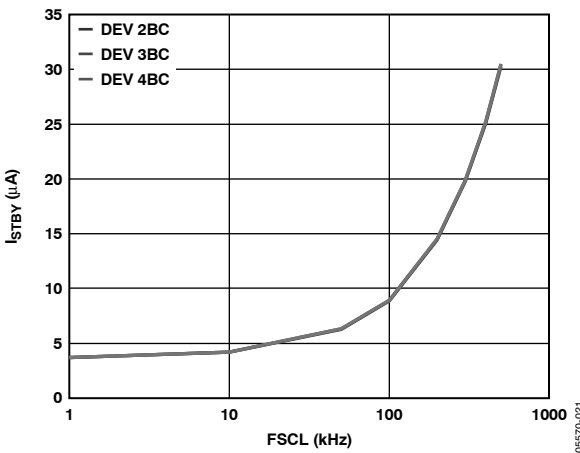


Figure 12. Standby Supply Current vs. SCLK Frequency

## THEORY OF OPERATION

The ADT7483A is a local and 2x remote temperature sensor and over/under temperature alarm. When the ADT7483A is operating normally, the on-board ADC operates in a free-running mode. The analog input multiplexer alternately selects either the on-chip temperature sensor or one of the remote temperature sensors to measure its local temperature. The ADC digitizes these signals, and the results are stored in the local, Remote 1, and Remote 2 temperature value registers.

The local and remote measurement results are compared with the corresponding high, low, and THERM temperature limits stored in on-chip registers. Out-of-limit comparisons generate flags that are stored in the status register. A result that exceeds the high temperature limit, the low temperature limit, or a remote diode open circuit causes the ALERT output to assert low. Likewise, exceeding THERM temperature limits causes the THERM output to assert low. The ALERT output can be reprogrammed as a second THERM output.

The limit registers can be programmed, and the device controlled and configured, via the serial SMBus. The contents of any register can also be read back via the SMBus.

Control and configuration functions consist of:

- Switching the device between normal operation and standby mode.
- Selecting the temperature measurement scale.
- Masking or enabling the ALERT output.
- Switching Pin 13 between ALERT and THERM2.
- Selecting the conversion rate.

## TEMPERATURE MEASUREMENT METHOD

A simple method of measuring temperature is to exploit the negative temperature coefficient of a diode, measuring the base-

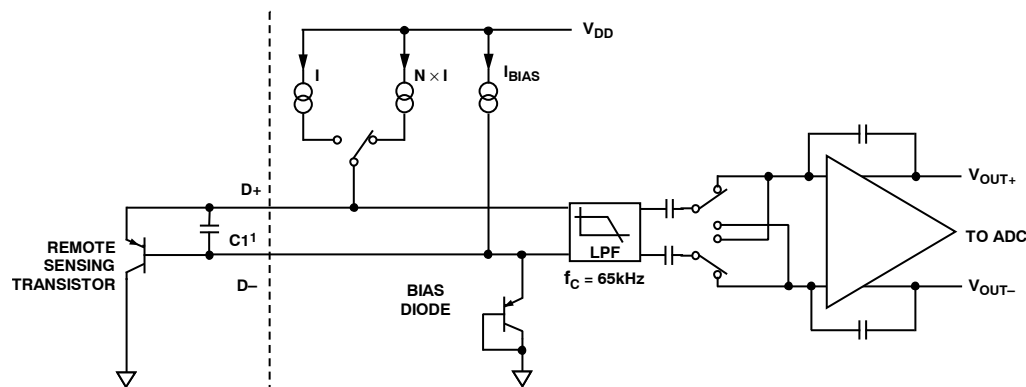
emitter voltage ( $V_{BE}$ ) of a transistor, operated at constant current. Unfortunately, this technique requires calibration to null the effect of the absolute value of  $V_{BE}$ , which varies from device to device. The technique used in the ADT7483A is to measure the change in  $V_{BE}$  when the device is operated at two different currents.

Figure 15 shows the input signal conditioning used to measure the output of a remote temperature sensor. This figure shows the remote sensor as a substrate transistor, but it could equally be a discrete transistor. If a discrete transistor is used, the collector is not grounded and should be linked to the base. To prevent ground noise interfering with the measurement, the more negative terminal of the sensor is not referenced to ground but is biased above ground by an internal diode at the D- input. C1 can be optionally added as a noise filter (recommended maximum value 1,000 pF).

To measure  $\Delta V_{BE}$ , the operating current through the sensor is switched among two related currents,  $I$  and  $N \times I$ . The currents through the temperature diode are switched between  $I$  and  $N \times I$ , giving  $\Delta V_{BE}$ . The temperature is then calculated using the  $\Delta V_{BE}$  measurement.

The resulting  $\Delta V_{BE}$  waveforms pass through a 65 kHz low-pass filter to remove noise and then to a chopper-stabilized amplifier. This amplifies and rectifies the waveform to produce a dc voltage proportional to  $\Delta V_{BE}$ . The ADC digitizes this voltage and produces a temperature measurement. To reduce the effects of noise, digital filtering is performed by averaging the results of 16 measurement cycles for low conversion rates. At rates of 16, 32, and 64 conversions/second, no digital averaging takes place.

Signal conditioning and measurement of the local temperature sensor is performed in the same manner.



<sup>1</sup>CAPACITOR C1 IS OPTIONAL. IT IS ONLY NECESSARY IN NOISY ENVIRONMENTS. C1 = 1000pF MAX.

Figure 15. Input Signal Conditioning

## TEMPERATURE MEASUREMENT RESULTS

The results of the local and remote temperature measurements are stored in the local and remote temperature value registers and are compared with limits programmed into the local and remote high and low limit registers.

The local temperature measurement is an 8-bit measurement with 1°C resolution. The remote temperature measurements are 10-bit measurements, with eight MSBs stored in one register and two LSBs stored in another register. Table 5 lists the temperature measurement registers.

**Table 5. Register Address for the Temperature Values**

Temperature Channel	MSB Register Address	LSB Register Address
Local	0x00	N/A
Remote 1	0x01	0x10 (2 MSBs)
Remote 2	0x30	0x33 (2 MSBs)

By setting Bit 3 of the Configuration 1 Register to 1, the Remote 2 temperature values can be read from the following register addresses:

Remote 2, MSBs = 0x01

Remote 2, LSBs = 0x10

The above is true only when Bit 3 of the Configuration 1 register is set. To read the Remote 1 temperatures, this bit needs to be switched back to 0.

Only the two MSBs in the remote temperature low byte are used. This gives the remote temperature measurement a resolution of 0.25°C. Table 6 shows the data format for the remote temperature low byte.

**Table 6. Extended Temperature Resolution (Remote Temperature Low Byte)**

Extended Resolution	Remote Temperature Low Byte
0.00°C	0 000 0000
0.25°C	0 100 0000
0.50°C	1 000 0000
0.75°C	1 100 0000

When reading the full remote temperature value, both the high and low byte, the two registers should be read LSB first and then the MSB. This is because reading the LSB will cause the MSB to be locked until it is read, guaranteeing that the two values read are a result of the same temperature measurement.

## TEMPERATURE MEASUREMENT RANGE

The temperature measurement range for both local and remote measurements is, by default, 0°C to 127°C. However, the ADT7483A can be operated using an extended temperature range from -64°C to +191°C. This means, the ADT7483A can measure the full temperature range of a remote thermal diode, from -55°C to +150°C. The user can switch between these two temperature ranges by setting or clearing Bit 2 in the Configuration 1 register. A valid result is available in the next measurement cycle after changing the temperature range.

In extended temperature mode, the upper and lower temperatures that can be measured by the ADT7483A are limited by the remote diode selection. The temperature registers themselves can have values from -64°C to +191°C. However, most temperature sensing diodes have a maximum temperature range of -55°C to +150°C.

Note that although both local and remote temperature measurements can be made while the part is in extended temperature mode, the ADT7483A should not be exposed to temperatures greater than those specified in the Absolute Maximum Ratings section. Further, the device is only guaranteed to operate as specified at ambient temperatures from -40°C to +120°C.

## TEMPERATURE DATA FORMAT

The ADT7483A has two temperature data formats. When the temperature measurement range is from 0°C to 127°C (default), the temperature data format for both local and remote temperature results is binary. When the measurement range is in extended mode, an offset binary data format is used for both local and remote results. Temperature values in the offset binary data format are offset by +64. Examples of temperatures in both data formats are shown in Table 7.

Table 7. Temperature Data Format (Local and Remote Temperature High Byte)

Temperature	Binary	Offset Binary <sup>1</sup>
-55°C	0 000 0000 <sup>2</sup>	0 000 1001
0°C	0 000 0000	0 100 0000
+1°C	0 000 0001	0 100 0001
+10°C	0 000 1010	0 100 1010
+25°C	0 001 1001	0 101 1001
+50°C	0 011 0010	0 111 0010
+75°C	0 100 1011	1 000 1011
+100°C	0 110 0100	1 010 0100
+125°C	0 111 1101	1 011 1101
+127°C	0 111 1111	1 011 1111
+150°C	0 111 1111 <sup>3</sup>	1 101 0110

<sup>1</sup> Offset binary scale temperature values are offset by +64.

<sup>2</sup> Binary scale temperature measurement returns 0 for all temperatures <0°C.

<sup>3</sup> Binary scale temperature measurement returns 127 for all temperatures >127°C.

The user may switch between measurement ranges at any time. Switching the range also switches the data format. The next temperature result following the switching is reported back to the register in the new format. However, the contents of the limit registers do not change. It is up to the user to ensure that when the data format changes, the limit registers are reprogrammed as necessary (for more information, see the Limit Registers section).

## REGISTERS

The registers in the ADT7483A are eight bits wide. These registers are used to store the results of remote and local temperature measurements, and high and low temperature limits, and to configure and control the device. A description of these registers is provided in this section.

### Address Pointer Register

The address pointer register does not have, nor does it require, an address because the first byte of every write operation is automatically written to this register. The data in this first byte always contains the address of another register on the ADT7483A, which is stored in the address pointer register. It is to this other register address that the second byte of a write operation is written, or to which a subsequent read operation is performed.

The power-on default value of the address pointer register is 0x00, so if a read operation is performed immediately after

power-on without first writing to the address pointer, the value of the local temperature will be returned, since its register address is 0x00.

### Temperature Value Registers

The ADT7483A has five registers to store the results of local and remote temperature measurements. These registers can only be written to by the ADC and can be read by the user over the SMBus.

- The local temperature value register is at Address 0x00.
- The Remote 1 temperature value high byte register is at Address 0x01, with the Remote 1 low byte register at Address 0x10.
- The Remote 2 temperature value high byte register is at Address 0x30, with the Remote 2 low byte register at Address 0x33.
- The Remote 2 temperature values can be read from Address 0x01 for the high byte and Address 0x10 for the low byte if Bit 3 of Configuration Register 1 is set to 1.
- To read the Remote 1 temperature values, Bit 3 of Configuration Register 1 should be set to 0.
- The power-on default for all five registers is 0x00.

# ADT7483A

## Configuration 1 Register

Table 8. Configuration 1 Register (Read Address = 0x03, Write Address = 0x09)

Bit	Mnemonic	Function
7	Mask	Setting this bit to 1 masks all ALERTs on the ALERT pin. Default = 0 = ALERT enabled. This applies only if Pin 13 is configured as ALERT, otherwise it has no effect.
6	Mon/STBY	Setting this bit to 1 places the ADT7483A in standby mode, that is, suspends all temperature measurements (ADC). The SMBus remains active and values can be written to, and read from, the registers. THERM and ALERT are also active in standby mode. Changes made to the limit registers in standby mode that affect the THERM or ALERT outputs will cause these signals to be updated. Default = 0 = temperature monitoring enabled.
5	AL/TH	This bit selects the function of Pin 13. Default = 0 = ALERT. Setting this bit to 1 configures Pin 13 as the THERM2 pin.
4	Reserved	Reserved for future use.
3	Remote 1/Remote 2	Setting this bit to 1 enables the user to read the Remote 2 values from the Remote 1 registers. Default = 0 = Remote 1 temperature values and limits are read from these registers. This bit is not lockable.
2	Temp Range	Setting this bit to 1 enables the extended temperature measurement range (–50°C to +150°C). Default = 0 = 0°C to +127°C.
1	Mask R1	Setting this bit to 1 masks ALERTs due to the Remote 1 temperature exceeding a programmed limit. Default = 0.
0	Mask R2	Setting this bit to 1 masks ALERTs due to the Remote 2 temperature exceeding a programmed limit. Default = 0.

## Configuration 2 Register

Table 9. Configuration 2 Register (Address = 0x24)

Bit	Mnemonic	Function
7	Lock Bit	Setting this bit to 1 locks all lockable registers to their current values. This prevents settings being tampered with until the device is powered down. Default = 0.
<6:0>	Reserved	Reserved for future use.

## Conversion Rate/Channel Selector Register

The conversion rate/channel selector register is at Address 0x04 for reads, and Address 0x0A for writes. The four LSBs of this register are used to program the conversion times from 15.5 ms (Code 0x0A) to 16 seconds (Code 0x00). To program the ADT7483A to perform continuous measurements, set the conversion rate register to 0x0B. For example, a conversion rate of 8 conversions/second means that beginning at 125 ms intervals, the device performs a conversion on the local and the remote temperature channels.

This register can be written to and read back over the SMBus. The default value of this register is 0x07, giving a rate of 8 conversions/second.

Bit 7 in this register can be used to disable averaging of the temperature measurements. The ADT7483A can be configured to take temperature measurements of either a single temperature channel or all temperature channels. Bit 5 and Bit 4 can be used to specify which temperature channel or channels are measured.

Table 10. Conversion Rate/Channel Selector Register

Bit	Mnemonic	Function																								
7	Averaging	Setting this bit to 1 disables averaging of the temperature measurements at the slower conversion rates (averaging cannot take place at the three faster rates, hence, setting this bit has no effect). Default = 0 = averaging enabled.																								
6	Reserved	Reserved for future use. Do not write to this bit.																								
<5:4>	Channel Selector	These bits are used to select the temperature measurement channels. 00 = round robin = default = all channels. 01 = local temperature. 10 = Remote 1 temperature. 11 = Remote 2 temperature.																								
<3:0>	Conversion Rates	These bits set how often the ADT7483A measures each temperature channel.																								
		<table border="1"> <thead> <tr> <th>Conversions/second</th> <th>Time (seconds)</th> </tr> </thead> <tbody> <tr> <td>0000 = 0.0625</td> <td>16</td> </tr> <tr> <td>0001 = 0.125</td> <td>8</td> </tr> <tr> <td>0010 = 0.25</td> <td>4</td> </tr> <tr> <td>0011 = 0.5</td> <td>2</td> </tr> <tr> <td>0100 = 1</td> <td>1</td> </tr> <tr> <td>0101 = 2</td> <td>500 m</td> </tr> <tr> <td>0110 = 4</td> <td>250 m</td> </tr> <tr> <td>0111 = 8 = default</td> <td>125 m</td> </tr> <tr> <td>1000 = 16</td> <td>62.5 m</td> </tr> <tr> <td>1001 = 32</td> <td>31.25 m</td> </tr> <tr> <td>1010 = continuous measurements</td> <td></td> </tr> </tbody> </table>	Conversions/second	Time (seconds)	0000 = 0.0625	16	0001 = 0.125	8	0010 = 0.25	4	0011 = 0.5	2	0100 = 1	1	0101 = 2	500 m	0110 = 4	250 m	0111 = 8 = default	125 m	1000 = 16	62.5 m	1001 = 32	31.25 m	1010 = continuous measurements	
Conversions/second	Time (seconds)																									
0000 = 0.0625	16																									
0001 = 0.125	8																									
0010 = 0.25	4																									
0011 = 0.5	2																									
0100 = 1	1																									
0101 = 2	500 m																									
0110 = 4	250 m																									
0111 = 8 = default	125 m																									
1000 = 16	62.5 m																									
1001 = 32	31.25 m																									
1010 = continuous measurements																										

### Limit Registers

The ADT7483A has three limits for each temperature channel: high, low, and THERM temperature limits for local, Remote 1, and Remote 2 temperature measurements. The remote temperature high and low limits span two registers each to contain an upper and lower byte for each limit. There is also a THERM hysteresis register. All limit registers can be written to and read back over the SMBus. See Table 8 for details of the limit registers' addresses and their power-on default values.

When Pin 13 is configured as an ALERT output, the high limit registers perform a > comparison while the low limit registers perform a ≤ comparison. For example, if the high limit register is programmed with 80°C, then measuring 81°C will result in an out-of-limit condition, setting a flag in the status register. If the low limit register is programmed with 0°C, measuring 0°C or lower will result in an out-of-limit condition.

Exceeding either the local or remote THERM limit asserts THERM low. When Pin 13 is configured as THERM2,

exceeding either the local limit or remote high limit asserts THERM2 low. A default hysteresis value of 10°C is provided that applies to both THERM channels. This hysteresis value may be reprogrammed to any value after power-up using Register Address 0x21.

It is important to remember that the data format for temperature limits is the same as the temperature measurement data format. Thus, if the temperature measurement uses the default (binary), then the temperature limits also use the binary scale. If the temperature measurement scale is switched, however, the temperature limits do not automatically switch. The user must reprogram the limit registers to the desired value in the correct data format. For example, if the remote low limit is set at 10°C and the default binary scale is being used, the limit register value should be 0000 1010b. If the scale is switched to offset binary, the value in the low temperature limit register should be reprogrammed to be 0100 1010b.

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## Status Registers

The status registers are read-only registers, at Address 0x02 (Status Register 1) and Address 0x23 (Status Register 2). They contain status information for the ADT7483A.

**Table 11. Status Register 1 Bit Assignments**

Bit	Mnemonic	Function	ALERT
7	BUSY	Bit set to 1 when ADC converting	No
6	LHIGH <sup>1</sup>	Bit set to 1 when local high temperature limit tripped	Yes
5	LLOW <sup>1</sup>	Bit set to 1 when local low temperature limit tripped	Yes
4	R1HIGH <sup>1</sup>	Bit set to 1 when remote 1 high temperature limit tripped	Yes
3	R1LOW <sup>1</sup>	Bit set to 1 when remote 1 low temperature limit tripped	Yes
2	D1 OPEN <sup>1</sup>	Bit set to 1 when remote 1 sensor open circuit	Yes
1	R1THRM1	Bit set to 1 when remote1 $\overline{\text{THERM}}$ limit tripped	No
0	LTHRM1	Bit set to 1 when local $\overline{\text{THERM}}$ limit tripped	No

<sup>1</sup> These flags stay high until the status register is read, or they are reset by POR.

**Table 12. Status Register 2 Bit Assignments**

Bit	Mnemonic	Function	ALERT
7	Res	Reserved for future use	No
6	Res	Reserved for future use	No
5	Res	Reserved for future use	No
4	R2HIGH <sup>1</sup>	Bit set to 1 when Remote 2 high temperature limit tripped	Yes
3	R2LOW <sup>1</sup>	Bit set to 1 when Remote 2 low temperature limit tripped	Yes
2	D2 OPEN <sup>1</sup>	Bit set to 1 when Remote 2 sensor open circuit	Yes
1	R2THRM1	Bit set to 1 when Remote2 $\overline{\text{THERM}}$ limit tripped	No
0	$\overline{\text{ALERT}}$	Bit set to 1 when $\overline{\text{ALERT}}$ condition exists	No

<sup>1</sup> These flags stay high until the status register is read, or they are reset by POR.

The eight flags that can generate an  $\overline{\text{ALERT}}$  are NOR'd together, so if any of them are high, the  $\overline{\text{ALERT}}$  interrupt latch is set and the  $\overline{\text{ALERT}}$  output goes low (provided they are not masked out).

Reading the Status 1 register will clear the five flags, Bit 6 to Bit 2 in Status Register 1, provided the error conditions that caused the flags to be set have gone away. Reading the Status 2 register will clear the three flags, Bit 4 to Bit 2 in Status Register 2, provided the error conditions that caused the flags to be set have gone away. A flag bit can only be reset if the corresponding value register contains an in-limit measurement or if the sensor is good.

The  $\overline{\text{ALERT}}$  interrupt latch is not reset by reading the status register. It is reset when the  $\overline{\text{ALERT}}$  output has been serviced by the master reading the device address, provided the error condition has gone away and the status register flag bits have been reset.

When Flag 1 and/or Flag 0 of Status Register 1, or Flag 1 of Status Register 2 are set, the  $\overline{\text{THERM}}$  output goes low to indicate that the temperature measurements are outside the programmed limits. The  $\overline{\text{THERM}}$  output does not need to be reset, unlike the  $\overline{\text{ALERT}}$  output. Once the measurements are within the limits, the corresponding status register bits are automatically reset and the  $\overline{\text{THERM}}$  output goes high. The user may add hysteresis by programming Register 0x21. The  $\overline{\text{THERM}}$  output will be reset only when the temperature falls below the  $\overline{\text{THERM}}$  limit minus hysteresis.

When Pin 13 is configured as  $\overline{\text{THERM2}}$ , only the high temperature limits are relevant. If Flag 6, Flag 4 of Status Register 1, or Flag 4 of Status Register 2 are set, the  $\overline{\text{THERM2}}$  output goes low to indicate that the temperature measurements are outside the programmed limits. Flag 5 and Flag 3 of Status Register 1, and Flag 3 of Status Register 2 have no effect on  $\overline{\text{THERM2}}$ . The behavior of  $\overline{\text{THERM2}}$  is otherwise the same as  $\overline{\text{THERM}}$ .

Bit 0 of Status Register 2 is set whenever the  $\overline{\text{ALERT}}$  output of the ADT7483A is asserted low. This means that the user need only read Status Register 2 to determine if the ADT7483A is responsible for the  $\overline{\text{ALERT}}$ . Bit 0 of Status Register 2 is reset when the  $\overline{\text{ALERT}}$  output is reset. If the  $\overline{\text{ALERT}}$  output is masked, then this bit is not set.

## Offset Register

Offset errors may be introduced into the remote temperature measurement by clock noise or by the thermal diode being located away from the hot spot. To achieve the specified accuracy on this channel, these offsets must be removed.

The offset values are stored as 10-bit, twos complement values. The Remote 1 offset MSBs are stored in Register 0x11, and the LSBs are stored 0x12 (low byte, left justified). The Remote 2 offset MSBs are stored in Register 0x34, and the LSBs are stored 0x35 (low byte, left justified). The Remote 2 offset can be written to, or read from, the Remote 1 offset registers if Bit 3 of the Configuration 1 register is set to 1. This bit should be set to 0 (default) to read the Remote 1 offset values.

Only the upper 2 bits of the LSB registers are used. The MSB of the MSB offset registers is the sign bit. The minimum offset that can be programmed is  $-128^{\circ}\text{C}$ , and the maximum is  $+127.75^{\circ}\text{C}$ .

The value in the offset register is added or subtracted to the measured value of the remote temperature.

The offset register powers up with a default value of 0°C and will have no effect unless the user writes a different value to it.

**Table 13. Sample Offset Register Codes**

Offset Value	0x11/0x34	0x12/0x35
-128°C	1000 0000	00 00 0000
-4°C	1111 1100	00 00 0000
-1°C	1111 1111	00 000000
-0.25°C	1111 1111	10 00 0000
0°C	0000 0000	00 00 0000
+0.25°C	0000 0000	01 00 0000
+1°C	0000 0001	00 00 0000
+4°C	0000 0100	00 00 0000
+127.75°C	0111 1111	11 00 0000

**One-Shot Register**

The one-shot register is used to initiate a conversion and comparison cycle when the ADT7483A is in standby mode, after which the device returns to standby. Writing to the one-shot register address (0x0F) causes the ADT7483A to perform a conversion and comparison on both the local and the remote temperature channels. This is not a data register as such, and it is the write operation to Address 0x0F that causes the one-shot

conversion. The data written to this address is irrelevant and is not stored.

**Consecutive ALERT Register**

The value written to this register determines how many out-of-limit measurements must occur before an ALERT is generated. The default value is that one out-of-limit measurement generates an ALERT. The maximum value that can be chosen is 4. The purpose of this register is to allow the user to perform some filtering of the output. This is particularly useful at the fastest three conversion rates, where no averaging takes place. This register is at Address 0x22.

**Table 14. Consecutive ALERT Register Bit**

Register Value	Number of Out-of-Limit Measurements Required
yzax 000x	1
yzax 001x	2
yzax 011x	3
yzax 111x	4

x = don't care bit.  
y = SMBus SCL timeout bit. Default = 0. See the SMBus section for more information.  
z = SMBus SDA timeout bit. Default = 0. See the SMBus section for more information.  
a = mask internal ALERTs

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Table 15. List of ADT7483A Registers

Read Address (Hex)	Write Address (Hex)	Mnemonic	Power-On Default	Comment	Lock
N/A	N/A	Address Pointer	Undefined		No
00	N/A	Local Temperature Value	0000 0000 (0x00)		No
01	N/A	Remote 1 Temperature Value High Byte	0000 0000 (0x00)	Bit 3 Conf Reg = 0	No
01	N/A	Remote 2 Temperature Value High Byte	0000 0000 (0x00)	Bit 3 Conf Reg = 1	No
02	N/A	Status Register 1	Undefined		No
03	09	Configuration Register 1	0000 0000 (0x00)		Yes
04	0A	Conversion Rate/Channel Selector	0000 0111 (0x07)		Yes
05	0B	Local Temperature High Limit	0101 0101 (0x55) (85°C)		Yes
06	0C	Local Temperature Low Limit	0000 0000 (0x00) (0°C)		Yes
07	0D	Remote 1 Temp High Limit High Byte	0101 0101 (0x55) (85°C)	Bit 3 Conf Reg = 0	Yes
07	0D	Remote 2 Temp High Limit High Byte	0101 0101 (0x55) (85°C)	Bit 3 Conf Reg = 1	Yes
08	0E	Remote 1 Temp Low Limit High Byte	0000 0000 (0x00) (0°C)	Bit 3 Conf Reg = 0	Yes
08	0E	Remote 2 Temp Low Limit High Byte	0000 0000 (0x00) (0°C)	Bit 3 Conf Reg = 1	Yes
N/A	0F <sup>1</sup>	One-Shot			N/A
10	N/A	Remote 1 Temperature Value Low Byte	0000 0000	Bit 3 Conf Reg = 0	No
10	N/A	Remote 2 Temperature Value Low Byte	0000 0000	Bit 3 Conf Reg = 1	No
11	11	Remote 1 Temperature Offset High Byte	0000 0000	Bit 3 Conf Reg = 0	Yes
11	11	Remote 2 Temperature Offset High Byte	0000 0000	Bit 3 Conf Reg = 1	Yes
12	12	Remote 1 Temperature Offset Low Byte	0000 0000	Bit 3 Conf Reg = 0	Yes
12	12	Remote 2 Temperature Offset Low Byte	0000 0000	Bit 3 Conf Reg = 1	Yes
13	13	Remote 1 Temp High Limit Low Byte	0000 0000	Bit 3 Conf Reg = 0	Yes
13	13	Remote 2 Temp High Limit Low Byte	0000 0000	Bit 3 Conf Reg = 1	Yes
14	14	Remote 1 Temp Low Limit Low Byte	0000 0000	Bit 3 Conf Reg = 0	Yes
14	14	Remote 2 Temp Low Limit Low Byte	0000 0000	Bit 3 Conf Reg = 1	Yes
19	19	Remote 1 $\overline{\text{THERM}}$ Limit	0101 0101 (0x55) (85°C)	Bit 3 Conf Reg = 0	Yes
19	19	Remote 2 $\overline{\text{THERM}}$ Limit	0101 0101 (0x55) (85°C)	Bit 3 Conf Reg = 1	Yes
20	20	Local $\overline{\text{THERM}}$ Limit	0101 0101 (0x55) (85°C)		Yes
21	21	$\overline{\text{THERM}}$ Hysteresis	0000 1010 (0x0A) (10°C)		Yes
22	22	Consecutive $\overline{\text{ALERT}}$	0000 0001 (0x01)		Yes
23	N/A	Status Register 2	0000 0000 (0x00)		No
24	24	Configuration 2 Register	0000 0000 (0x00)		Yes
30	N/A	Remote 2 Temperature Value High Byte	0000 0000 (0x00)		No
31	31	Remote 2 Temp High Limit High Byte	0101 0101 (0x55) (85°C)		Yes
32	32	Remote 2 Temp Low Limit High Byte	0000 0000 (0x00) (0°C)		Yes
33	N/A	Remote 2 Temperature Value Low Byte	0000 0000 (0x00)		No
34	34	Remote 2 Temperature Offset High Byte	0000 0000 (0x00)		Yes
35	35	Remote 2 Temperature Offset Low Byte	0000 0000 (0x00)		Yes
36	36	Remote 2 Temp High Limit Low Byte	0000 0000 (0x00) (0°C)		Yes
37	37	Remote 2 Temp Low Limit Low Byte	0000 0000 (0x00) (0°C)		Yes
39	39	Remote 2 $\overline{\text{THERM}}$ limit	0101 0101 (0x55) (85°C)		Yes
FE	N/A	Manufacturer ID	0100 0001 (0x41)		N/A
FF	N/A	Die Revision Code	1001 0100 (0x94)		N/A

<sup>1</sup> Writing to Address 0F causes the ADT7483A to perform a single measurement. It is not a data register, as such, and it does not matter what data is written to it.

## SERIAL BUS INTERFACE

Control of the ADT7483A is carried out via the serial bus. The ADT7483A is connected to the serial bus as a slave device, under the control of a master device.

The ADT7483A has an SMBus timeout feature. When this is enabled, the SMBus typically times out after 25 ms of no activity. However, this feature is not enabled by default. Bit 7 (SCL timeout bit) of the consecutive ALERT register (Address = 0x22) should be set to enable the SCL timeout. Bit 6 (SDA timeout bit) of the consecutive ALERT register (Address = 0x22) should be set to enable the SDA timeout.

The ADT7483A supports packet error checking (PEC) and its use is optional. It is triggered by supplying the extra clock for the PEC byte. The PEC byte is calculated using CRC-8. The frame check sequence (FCS) conforms to CRC-8 by the polynomial

$$C(x) = x^8 + x^2 + x^1 + 1$$

Consult the SMBus 1.1 specification for more information ([www.smbus.org](http://www.smbus.org)).

## ADDRESSING THE DEVICE

### Address Pins

In general, every SMBus device has a 7-bit device address (except for some devices that have extended, 10-bit addresses). When the master device sends a device address over the bus, the slave device with that address will respond. The ADT7483A has two address pins, ADD0 and ADD1, to allow selection of the device address, so that several ADT7483As can be used on the same bus, and/or to avoid conflict with other devices.

Although only two address pins are provided, these are three-state, and can be grounded, left unconnected, or tied to  $V_{DD}$ , so that a total of nine different addresses are possible, as shown in Table 16. It should be noted that the state of the address pins is only sampled at power-up, so changing them after power-up has no effect.

**Table 16. Device Addresses**

ADD1	ADD0	Device Address
0	0	0011 000
0	NC	0011 001
0	1	0011 010
NC	0	0101 001
NC	NC	0101 010
NC	1	0101 011
1	0	1001 100
1	NC	1001 101
1	1	1001 110

The serial bus protocol operates as follows:

1. The master initiates data transfer by establishing a start condition, defined as a high-to-low transition on the serial data line (SDATA), while the serial clock line (SCLK) remains high. This indicates that an address/data stream follows. All slave peripherals connected to the serial bus respond to the start condition and shift in the next eight bits, consisting of a 7-bit address (MSB first) plus an R/W bit, which determines the direction of the data transfer, that is, whether data will be written to, or read from, the slave device. The peripheral whose address corresponds to the transmitted address responds by pulling the data line low during the low period before the ninth clock pulse, known as the acknowledge bit. All other devices on the bus now remain idle while the selected device waits for data to be read from or written to it. If the R/W bit is a 0, the master writes to the slave device. If the R/W bit is a 1, the master reads from the slave device.
2. Data is sent over the serial bus in a sequence of nine clock pulses, eight bits of data followed by an acknowledge bit from the slave device. Transitions on the data line must occur during the low period of the clock signal and remain stable during the high period, since a low-to-high transition when the clock is high may be interpreted as a stop signal. The number of data bytes that can be transmitted over the serial bus in a single read or write operation is limited only by what the master and slave devices can handle.
3. When all data bytes have been read or written, stop conditions are established. In write mode, the master will pull the data line high during the tenth clock pulse to assert a stop condition. In read mode, the master device will override the acknowledge bit by pulling the data line high during the low period before the ninth clock pulse. This is known as no acknowledge. The master will then take the data line low during the low period before the tenth clock pulse, then high during the tenth clock pulse to assert a stop condition.

Any number of bytes of data may be transferred over the serial bus in one operation, but it is not possible to mix read and write in one operation because the type of operation is determined at the beginning and cannot subsequently be changed without starting a new operation. For the ADT7483A, write operations contain either one or two bytes, while read operations contain one byte.

To write data to one of the device data registers, or to read data from it, the address pointer register must be set so that the correct data register is addressed. The first byte of a write operation always contains a valid address that is stored in the address pointer register. If data is to be written to the device, the

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write operation contains a second data byte that is written to the register selected by the address pointer register (see Figure 16).

The device address is sent over the bus followed by  $\overline{R/\overline{W}}$  set to 0. This is followed by two data bytes. The first data byte is

the address of the internal data register to be written to, which is stored in the address pointer register. The second data byte is the data to be written to the internal data register.

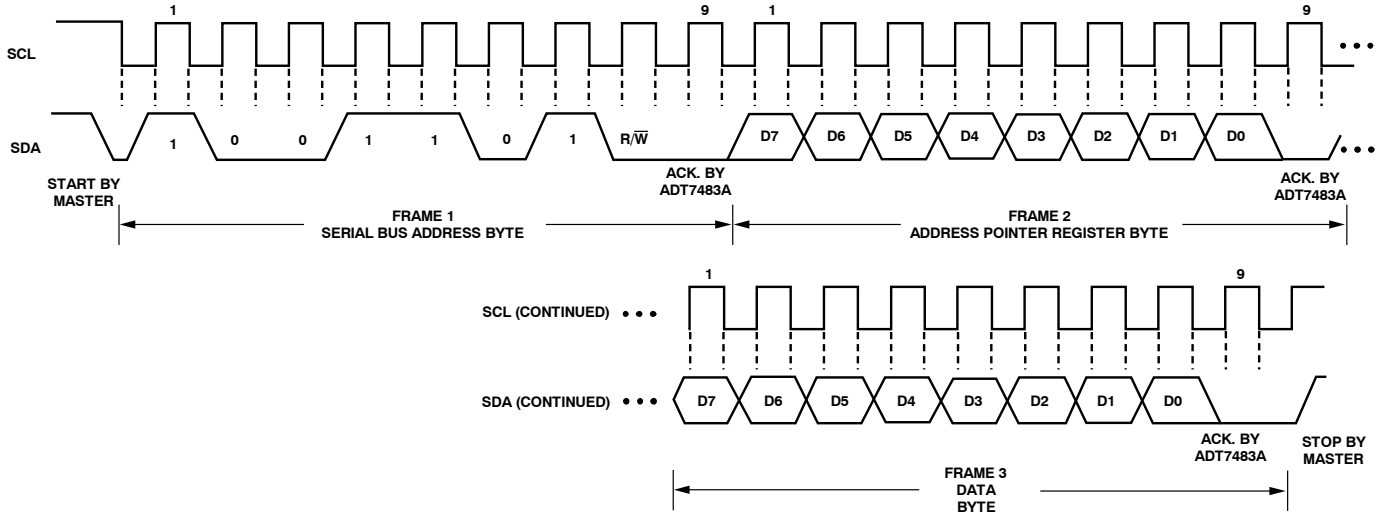


Figure 16. Writing a Register Address to the Address Pointer Register, then Writing Data to the Selected Register

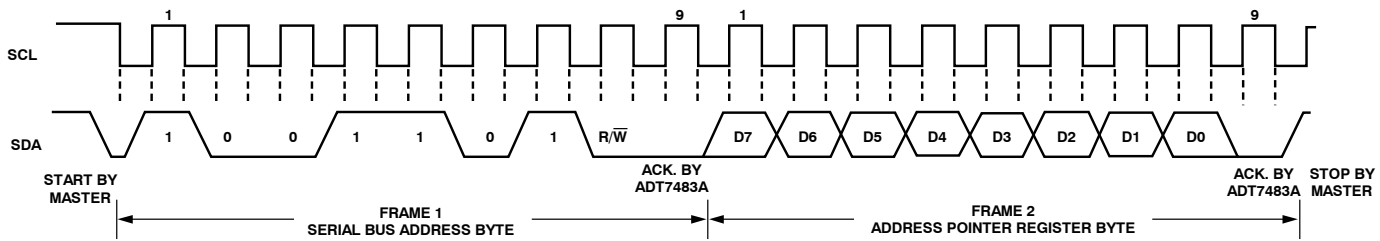


Figure 17. Writing to the Address Pointer Register Only

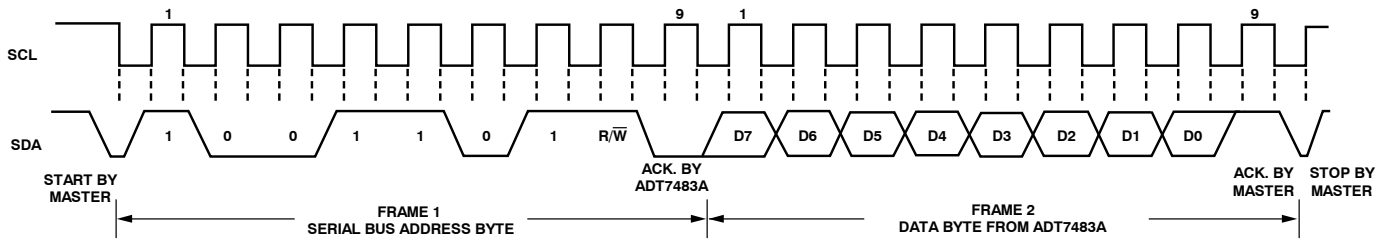


Figure 18. Reading from a Previously Selected Register

When reading data from a register there are two possibilities:

- If the address pointer register value of the ADT7483A is unknown or not the desired value, it is first necessary to set it to the correct value before data can be read from the desired data register. This is done by performing a write to the ADT7483A as before, but only the data byte containing the register read address is sent, as data is not to be written to the register (see Figure 17).

A read operation is then performed consisting of the serial bus address,  $\overline{R/\overline{W}}$  bit set to 1, followed by the data byte read from the data register (see Figure 18).

- If the address pointer register is known to be already at the desired address, data can be read from the corresponding data register without first writing to the address pointer register and the bus transaction shown in Figure 17 can be omitted.

**Notes**

- It is possible to read a data byte from a data register without first writing to the address pointer register. However, if the address pointer register is already at the correct value, it is not possible to write data to a register without writing to the address pointer register because the first data byte of a write is always written to the address pointer register.
- Remember that some of the ADT7483A registers have different addresses for read and write operations. The write address of a register must be written to the address pointer if data is to be written to that register, but it may not be possible to read data from that address. The read address of a register must be written to the address pointer before data can be read from that register.

**ALERT OUTPUT**

This is applicable when Pin 13 is configured as an ALERT output. The ALERT output goes low whenever an out-of-limit measurement is detected, or if the remote temperature sensor is open circuit. It is an open-drain output and requires a pull-up to V<sub>DD</sub>. Several ALERT outputs can be wire-ORed together, so that the common line will go low if one or more of the ALERT outputs goes low.

The ALERT output can be used as an interrupt signal to a processor, or it can be used as an SMBALERT. Slave devices on the SMBus cannot normally signal to the bus master that they want to talk, but the SMBALERT function allows them to do so.

One or more ALERT outputs can be connected to a common SMBALERT line connected to the master. When the SMBALERT line is pulled low by one of the devices, the following procedure occurs, as shown in Figure 19.

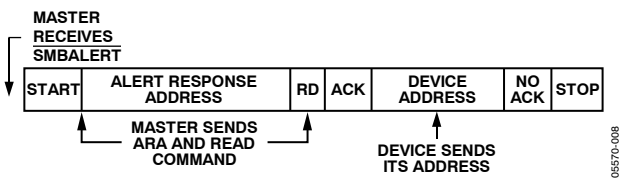


Figure 19. Use of SMBALERT

1. SMBALERT is pulled low.
2. Master initiates a read operation and sends the alert response address (ARA = 0001 100). This is a general call address that must not be used as a specific device address.
3. The device whose ALERT output is low responds to the alert response address, and the master reads its device address. The device address is seven bits, so an LSB of 1 is

added. The address of the device is now known and it can be interrogated in the usual way.

4. If more than one device's ALERT output is low, the one with the lowest device address will have priority, in accordance with normal SMBus arbitration.
5. Once the ADT7483A has responded to the alert response address, it will reset its ALERT output, provided that the error condition that caused the ALERT no longer exists. If the SMBALERT line remains low, the master will send the ARA again, and so on, until all devices whose ALERT outputs were low have responded.

**MASKING THE ALERT OUTPUT**

The ALERT output can be masked for local, Remote 1, Remote 2, or all three channels. This is done by setting the appropriate mask bits in either the Configuration 1 register (read address = 0x03, write address = 0x09) or in the consecutive ALERT register (address = 0x22)

To mask ALERTs due to local temperature, set Bit 5 of the consecutive ALERT register to 1. Default = 0.

To mask ALERTs due to Remote 1 temperature, set Bit 1 of the Configuration 1 register to 1. Default = 0.

To mask ALERTs due to Remote 2 temperature, set Bit 0 of the Configuration 1 register to 1. Default = 0.

To mask ALERTs due to any channel, set Bit 7 of the Configuration 1 register to 1. Default = 0.

**LOW POWER STANDBY MODE**

The ADT7483A can be put into low power standby mode by setting Bit 6 (Mon/STBY bit) of the Configuration 1 register (read address = 0x03, write address = 0x09) to 1. When Bit 6 is 0, the ADT7483A operates normally. When Bit 6 is 1, the ADC is inhibited, and any conversion in progress is terminated without writing the result to the corresponding value register.

The SMBus is still enabled. Power consumption in the standby mode is reduced to less than 5 μA.

When the device is in standby mode, it is still possible to initiate a one-shot conversion of both channels by writing to the one-shot register (Address 0x0F), after which the device will return to standby. It does not matter what is written to the one-shot register, all data written to it is ignored.

It is also possible to write new values to the limit register while in standby mode. If the values stored in the temperature value

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registers are now outside the new limits, an ALERT is generated, even though the ADT7483A is still in standby.

## SENSOR FAULT DETECTION

The ADT7483A has internal sensor fault detection circuitry located at its D+ input. This circuit can detect situations where a remote diode is not connected, or is incorrectly connected, to the ADT7483A. A simple voltage comparator trips if the voltage at D+ exceeds  $V_{DD} - 1\text{ V}$  (typical), signifying an open circuit between D+ and D-. The output of this comparator is checked when a conversion is initiated. Bit 2 (D1 OPEN flag) of the Status Register 1 (Address 0x02) is set if a fault is detected on the Remote 1 channel. Bit 2 (D2 OPEN flag) of the Status Register 2 (Address 0x23) is set if a fault is detected on the Remote 2 channel. If the ALERT pin is enabled, setting this flag will cause ALERT to assert low.

If a remote sensor is not used with the ADT7483A, then the D+ and D- inputs of the ADT7483A need to be tied together to prevent the OPEN flag from being continuously set.

Most temperature sensing diodes have an operating temperature range of  $-55^{\circ}\text{C}$  to  $+150^{\circ}\text{C}$ . Above  $150^{\circ}\text{C}$ , they lose their semiconductor characteristics and approximate conductors instead. This results in a diode short, setting the OPEN flag. The remote diode in this case no longer gives an accurate temperature measurement. A read of the temperature result register will give the last good temperature measurement. The user should be aware that, while the diode fault is triggered, the temperature measurement on the remote channels may not be accurate.

## INTERRUPT SYSTEM

The ADT7483A has two interrupt outputs, ALERT and THERM. Both outputs have different functions and behavior. ALERT is maskable and responds to violations of software programmed temperature limits or an open-circuit fault on the remote diode. THERM is intended as a fail-safe interrupt output that cannot be masked.

If the Remote 1, Remote 2, or local temperature exceeds the programmed high temperature limits, or equals or exceeds the low temperature limits, the ALERT output is asserted low. An open-circuit fault on the remote diode also causes ALERT to assert. ALERT is reset when serviced by a master reading its device address, provided the error condition has gone away and the status register has been reset.

Similarly, the THERM output asserts low if the Remote 1, Remote 2, or local temperature exceeds the programmed THERM limits. The THERM temperature limits should normally be equal to or greater than the high temperature limits. THERM is automatically reset when the temperature falls back within the (THERM - Hysteresis) limit. The local and remote THERM limits are set by default to  $85^{\circ}\text{C}$ . An hysteresis value can be programmed, in which case, THERM resets when the temperature falls to the limit value minus the hysteresis value. This applies to both local and remote measurement channels. The power-on hysteresis default value is  $10^{\circ}\text{C}$ , but this may be reprogrammed to any value after power-up.

The hysteresis loop on the THERM outputs is useful when THERM is used for on/off control of a fan. The user's system can be set up so that when THERM asserts, a fan can be switched on to cool the system. When THERM goes high again, the fan can be switched off. Programming an hysteresis value protects from fan jitter, wherein the temperature hovers around the THERM limit and the fan is constantly being switched.

Table 17. THERM Hysteresis

THERM Hysteresis	Binary Representation
$0^{\circ}\text{C}$	0 000 0000
$1^{\circ}\text{C}$	0 000 0001
$10^{\circ}\text{C}$	0 000 1010

Figure 20 shows how the THERM and ALERT outputs operate. The ALERT output can be used as an SMBALERT to signal to the host via the SMBus that the temperature has risen. If the temperature continues to increase, the THERM output can be used to turn on a fan to cool the system. This method ensures that there is a fail-safe mechanism to cool the system, without the need for host intervention.

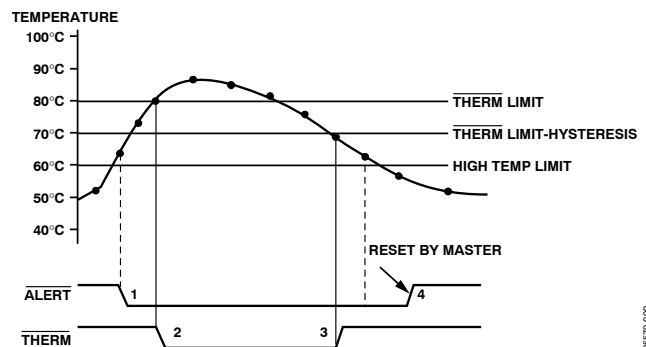


Figure 20. Operation of the ALERT and THERM Outputs

- If the measured temperature exceeds the high temperature limit, the  $\overline{\text{ALERT}}$  output asserts low.
- If the temperature continues to increase and exceeds the THERM limit, the THERM output asserts low. This can be used to throttle the CPU clock or switch on a fan.
- The  $\overline{\text{THERM}}$  output deasserts (goes high) when the temperature falls to  $\overline{\text{THERM}}$  limit minus hysteresis. In Figure 20, the default hysteresis value of 10°C is shown.
- The  $\overline{\text{ALERT}}$  output deasserts only when the temperature has fallen below the high temperature limit, and the master has read the device address and cleared the status register.

Pin 13 on the ADT7483A can be configured as either an  $\overline{\text{ALERT}}$  output or as an additional THERM output. THERM2 will assert low when the temperature exceeds the programmed local and/or remote high temperature limits. It is reset in the same manner as THERM, and it is not maskable. The programmed hysteresis value also applies to THERM2. Figure 21 shows how THERM and THERM2 might operate together to implement two methods of cooling the system. In this example, the THERM2 limits are set lower than the THERM limits. The THERM2 output can be used to turn on a fan. If the temperature continues to rise and exceeds the THERM limits, the  $\overline{\text{THERM}}$  output can provide additional cooling by throttling the CPU.

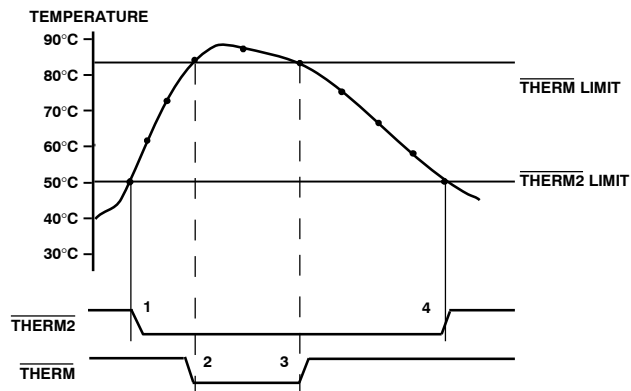


Figure 21. Operation of the  $\overline{\text{THERM}}$  and  $\overline{\text{THERM2}}$  Interrupts

- When the THERM2 limit is exceeded, the  $\overline{\text{THERM2}}$  signal asserts low.
- If the temperature continues to increase and exceeds the THERM limit, the  $\overline{\text{THERM}}$  output asserts low.
- The  $\overline{\text{THERM}}$  output deasserts (goes high) when the temperature falls to THERM limit minus hysteresis. In Figure 21, there is no hysteresis value shown.
- As the system cools further, and the temperature falls below the THERM2 limit, the THERM2 signal resets. Again, no hysteresis value is shown for THERM2.

The temperature measurement can be either the local or the remote temperature measurement.

## APPLICATIONS

### NOISE FILTERING

For temperature sensors operating in noisy environments, previous practice was to place a capacitor across the D+ and D– pins to help combat the effects of noise. However, large capacitances affect the accuracy of the temperature measurement, leading to a recommended maximum capacitor value of 1,000 pF.

### FACTORS AFFECTING DIODE ACCURACY

#### Remote Sensing Diode

The ADT7483A is designed to work with substrate transistors built into processors or with discrete transistors. Substrate transistors will generally be PNP types with the collector connected to the substrate. Discrete types can be either a PNP or NPN transistor connected as a diode (base shorted to collector). If an NPN transistor is used, the collector and base are connected to D+ and the emitter to D–. If a PNP transistor is used, the collector and base are connected to D– and the emitter to D+.

To reduce the error due to variations in both substrate and discrete transistors, the following factors should be taken into consideration:

- The ideality factor,  $n_f$ , of the transistor is a measure of the deviation of the thermal diode from ideal behavior. The ADT7483A is trimmed for an  $n_f$  value of 1.008. Use the following equation to calculate the error introduced at a temperature,  $T$  (°C) when using a transistor whose  $n_f$  does not equal 1.008. Consult the processor data sheet for the  $n_f$  values.

$$\Delta T = (n_f - 1.008) / 1.008 \times (273.15 \text{ Kelvin} + T)$$

To factor this in, write the  $\Delta T$  value to the offset register. It is then automatically added to, or subtracted from, the temperature measurement by the ADT7483A.

- Some CPU manufacturers specify the high and low current levels of the substrate transistors. The high current level of the ADT7483A,  $I_{\text{HIGH}}$ , is 200  $\mu\text{A}$ , and the low level current,  $I_{\text{LOW}}$ , is 12  $\mu\text{A}$ . If the ADT7483A current levels do not match the current levels specified by the CPU manufacturer, it may be necessary to remove an offset. Refer to the CPU data sheet to determine whether this offset needs to be removed and how to calculate it. This offset is programmed to the offset register. It is important to note that if more than one offset must be considered, program the algebraic sum of these offsets to the offset register.

If a discrete transistor is used with the ADT7483A, the best accuracy is obtained by choosing devices according to the following criteria:

- Base-emitter voltage greater than 0.25 V at 6  $\mu\text{A}$ , at the highest operating temperature.
- Base-emitter voltage less than 0.95 V at 100  $\mu\text{A}$ , at the lowest operating temperature.
- Base resistance less than 100  $\Omega$ .
- Small variation in  $h_{\text{FE}}$  (50 to 150) that indicates tight control of  $V_{\text{BE}}$  characteristics.

Transistors such as 2N3904, 2N3906, or equivalents in SOT-23 packages, are suitable devices to use.

### THERMAL INERTIA AND SELF-HEATING

Accuracy depends on the temperature of the remote sensing diode and/or the local temperature sensor being at the same temperature as that being measured. A number of factors can affect this. Ideally, the sensor should be in good thermal contact with the part of the system being measured. If it is not, the thermal inertia caused by the sensor's mass causes a lag in the response of the sensor to a temperature change. In the case of the remote sensor, this should not be a problem, since it will either be a substrate transistor in the processor or a small package device, such as SOT-23, placed in close proximity to it.

The on-chip sensor, however, is often remote from the processor and only monitors the general ambient temperature around the package. In practice, the ADT7483A package will be in electrical, and hence thermal, contact with a PCB and may also be in a forced airflow. How accurately the temperature of the board and/or the forced airflow reflects the temperature to be measured will also affect the accuracy. Self-heating, due to the power dissipated in the ADT7483A or the remote sensor, causes the chip temperature of the device or remote sensor to rise above ambient. However, the current forced through the remote sensor is so small that self-heating is negligible. In the case of the ADT7483A, the worst-case condition occurs when the device is converting at 64 conversions per second while sinking the maximum current of 1 mA at the ALERT and THERM output. In this case, the total power dissipation in the device is about 4.5 mW. The thermal resistance,  $\theta_{\text{JA}}$ , of the QSOP-16 package is about 150°C/W.

## LAYOUT CONSIDERATIONS

Digital boards can be electrically noisy environments, and the ADT7483A measures very small voltages from the remote sensor, so care must be taken to minimize noise induced at the sensor inputs. Follow these precautions:

- Place the ADT7483A as close as possible to the remote sensing diode. Provided that the worst noise sources such as clock generators, data/address buses, and CRTs are avoided, this distance can be 4 inches to 8 inches.
- Route the D+ and D- tracks close together, in parallel, with grounded guard tracks on each side. To minimize inductance and reduce noise pickup, a 5 mil track width and spacing is recommended. Provide a ground plane under the tracks, if possible.



Figure 22. Typical Arrangement of Signal Tracks

- Minimize the number of copper/solder joints that can cause thermocouple effects. Where copper/solder joints are used, make sure that they are in both the D+ and D- path and at the same temperature.

Thermocouple effects should not be a major problem as 1°C corresponds to approximately 200 mV, and thermocouple voltages are about 3 mV/°C of temperature difference. Unless there are two thermocouples with a large temperature differential between them, thermocouple voltages should be much less than 200 mV.

- Place a 0.1 μF bypass capacitor close to the V<sub>DD</sub> pin. In extremely noisy environments, place an input filter capacitor across D+ and D- close to the ADT7483A. This capacitance can effect the temperature measurement, so care must be taken to ensure that any capacitance seen at D+ and D- is a maximum of 1,000 pF. This maximum value includes the filter capacitance, plus any cable or stray capacitance between the pins and the sensor diode.
- If the distance to the remote sensor is more than 8 inches, the use of twisted pair cable is recommended. A total of 6 feet to 12 feet is needed.
- For very long distances (up to 100 feet), use shielded twisted pair, such as Belden No. 8451 microphone cable. Connect the twisted pair to D+ and D-, and the shield to GND close to the ADT7483A. Leave the remote end of the shield unconnected to avoid ground loops.

Because the measurement technique uses switched current sources, excessive cable or filter capacitance can affect the measurement. When using long cables, the filter capacitance can be reduced or removed.

## APPLICATION CIRCUIT

Figure 23 shows a typical application circuit for the ADT7483A, using discrete sensor transistors. The pull-ups on SCLK, SDATA, and ALERT are required only if they are not already provided elsewhere in the system.

The SCLK and SDATA pins of the ADT7483A can be interfaced directly to the SMBus of an I/O controller, such as the Intel® 820 chipset.

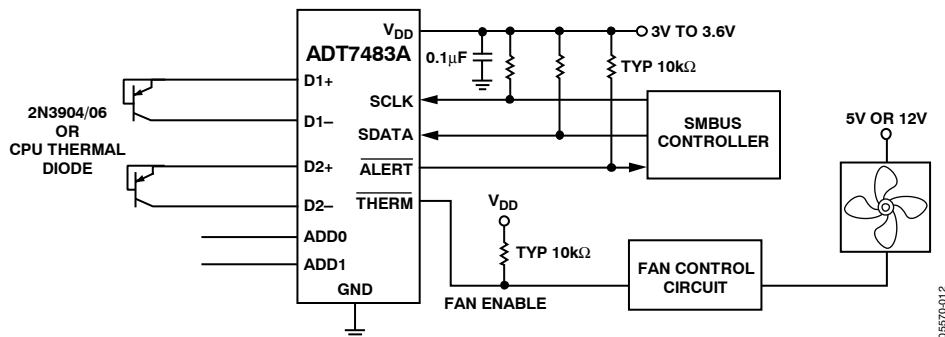
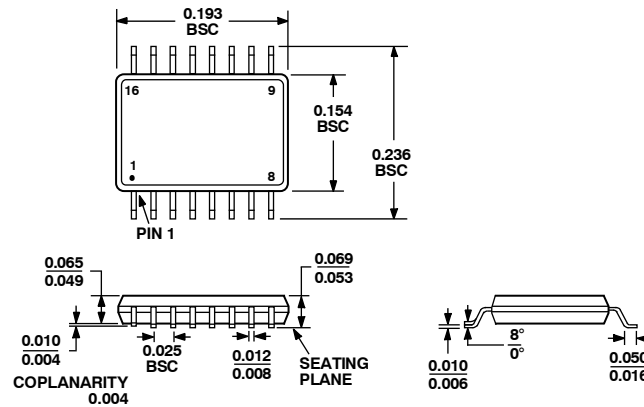


Figure 23. Typical Application Circuit

## OUTLINE DIMENSIONS



COMPLIANT TO JEDEC STANDARDS MO-137-AB


Figure 24. 16-Lead Shrink Small Outline Package [QSOP]  
(RQ-16)

Dimensions shown in inches

## ORDERING GUIDE

Model	Operating Temperature Range	Package Description	Package Option
ADT7483AARQZ <sup>1</sup>	-40°C to +125°C	16-Lead QSOP Package	RQ-16
ADT7483AARQZ-RL <sup>1</sup>	-40°C to +125°C	16-Lead QSOP Package	RQ-16
ADT7483AARQZ-R7 <sup>1</sup>	-40°C to +125°C	16-Lead QSOP Package	RQ-16
EVAL-ADT7483EB	-40°C to +125°C	Evaluation Board	

<sup>1</sup>Z = Pb-free part.

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