# **NPT Series N-Channel IGBT**

72 A, 1200 V

# HGTG27N120BN

The HGTG27N120BN is Non–Punch Through (NPT) IGBT design. This is a new member of the MOS gated high voltage switching IGBT family. IGBTs combine the best features of MOSFETs and bipolar transistors. This device has the high input impedance of a MOSFET and the low on–state conduction loss of a bipolar transistor.

The IGBT is ideal for many high voltage switching applications operating at moderate frequencies where low conduction losses are essential, such as: AC and DC motor controls, power supplies and drivers for solenoids, relays and contactors.

Formerly Developmental Type TA49280.

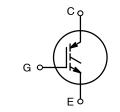
## Features

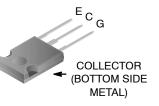
- 72 A, 1200 V,  $T_C = 25^{\circ}C$
- 1200 V Switching SOA Capability
- Typical Fall Time 140 ns at  $T_J = 150^{\circ}C$
- Short Circuit Rating
- Low Conduction Loss
- *Thermal Impedance* SPICE Model *Temperature Compensating* SABER<sup>™</sup> Model
- Avalanche Rated
- This is a Pb–Free Device



# **ON Semiconductor®**

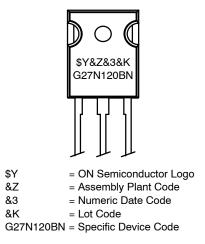
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TO-247-3LD SHORT LEAD CASE 340CK JEDEC STYLE

### MARKING DIAGRAM



#### **ORDERING INFORMATION**

See detailed ordering and shipping information on page 7 of this data sheet.

#### ABSOLUTE MAXIMUM RATINGS (T<sub>C</sub> = 25°C unless otherwise specified)

Parameter	Symbol	HGTG27N120BN	Unit	
Collector to Emitter Voltage	BV <sub>CES</sub>	1200	V	
Collector Current Continuous At $T_C = 25^{\circ}C$ At $T_C = 110^{\circ}C$	I <sub>C25</sub> I <sub>C110</sub>	72 34	A A	
Collector Current Pulsed (Note 1)	I <sub>CM</sub>	216	А	
Gate to Emitter Voltage Continuous	V <sub>GES</sub>	±20	V	
Gate to Emitter Voltage Pulsed	V <sub>GEM</sub>	±30	V	
Switching Safe Operating Area at $T_J = 150^{\circ}C$ (Figure 2)	SSOA	150 A at 1200 V		
Power Dissipation Total at $T_C = 25^{\circ}C$	PD	500	W	
Power Dissipation Derating $T_C > 25^{\circ}C$		4.0	W/°C	
Forward Voltage Avalanche Energy (Note 2)	E <sub>AV</sub>	135	W/°C	
Operating and Storage Junction Temperature Range	T <sub>J</sub> , T <sub>STG</sub>	–55 to 150	°C	
Maximum Lead Temperature for Soldering	TL	260	°C	
Short Circuit Withstand Time (Note 3) at $V_{GE}$ = 15 V	t <sub>SC</sub>	8	μs	
Short Circuit Withstand Time (Note 3) at V <sub>GE</sub> = 12 V	t <sub>SC</sub>	15	μs	

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.
Pulse width limited by maximum junction temperature.
I<sub>CE</sub> = 30 A, L = 400 mH, T<sub>J</sub> = 125°C.
V<sub>CE(PK)</sub> = 960 V, T<sub>J</sub> = 125°C, R<sub>G</sub> = 3 Ω.

### **ELECTRICAL CHARACTERISTICS** ( $T_C = 25^{\circ}C$ unless otherwise specified)

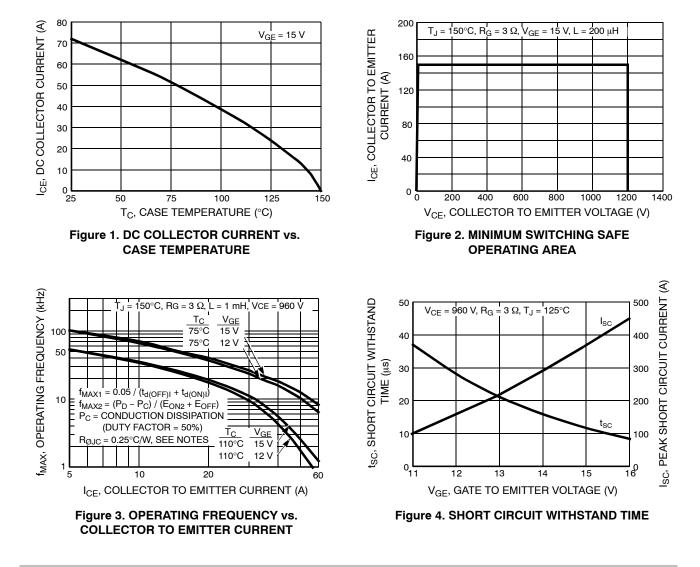
Parameter	Symbol	Test Condition		Min	Тур	Max	Unit
Collector to Emitter Breakdown Voltage	BV <sub>CES</sub>	$I_{C} = 250 \ \mu A, \ V_{GE} = 0 \ V$		1200	-	-	V
Emitter to Collector Breakdown Voltage	BV <sub>ECS</sub>	I <sub>C</sub> = 10 mA, V <sub>GE</sub> = 0 V		15	-	-	V
Collector to Emitter Leakage Current	I <sub>CES</sub>	V <sub>CE</sub> = 1200 V	$T_C = 25^{\circ}C$	-	-	250	μA
			T <sub>C</sub> = 125°C	-	300	-	μA
			T <sub>C</sub> = 150°C	-	-	4	mA
Collector to Emitter Saturation Voltage $V_{CE(SAT)}$ $I_C = 27$	I <sub>C</sub> = 27 A, V <sub>GE</sub> = 15 V	T <sub>C</sub> = 25°C	-	2.45	2.7	V	
			T <sub>C</sub> = 150°C	-	3.8	4.2	V
Gate to Emitter Threshold Voltage	V <sub>GE(TH)</sub>	I <sub>C</sub> = 250 μA, V <sub>CE</sub> = V <sub>GE</sub>		6	6.6	-	V
Gate to Emitter Leakage Current	I <sub>GES</sub>	V <sub>GE</sub> = ±20 V		-	-	±250	nA
Switching SOA	SSOA	$ \begin{array}{l} T_J = 150^{\circ}C, \ R_G = 3 \ \Omega, \ V_{GE} = 15 \ V, \\ L = 200 \ \mu H, \ V_{CE(PK)} = \ 1200 \ V \end{array} $		150	_	-	A
Gate to Emitter Plateau Voltage	V <sub>GEP</sub>	$I_{C} = I_{C110}, V_{CE} = 0.5 \text{ BV}_{CES}$		-	9.2	-	V
On-State Gate Charge	Q <sub>G(ON)</sub>	I <sub>C</sub> = 27 A, V <sub>CE</sub> = 600 V	V <sub>GE</sub> = 15 V	-	270	325	nC
			V <sub>GE</sub> = 20 V	-	350	420	nC
Current Turn-On Delay Time	t <sub>d(ON)</sub> I	IGBT and Diode at $T_J = 25^{\circ}C$ , $I_{CE} = 27 A$ , $V_{CE} = 960 V$ $V_{GE} = 15 V$ , $R_G = 3 \Omega$ , L = 1 mH, Test Circuit (Figure 18)		_	24	30	ns
Current Rise Time	t <sub>rl</sub>			-	20	25	ns
Current Turn-Off Delay Time	t <sub>d(OFF)</sub> I			-	195	240	ns
Current Fall Time	t <sub>fl</sub>			-	80	120	ns
Turn-On Energy (Note 5)	E <sub>ON1</sub>			-	2.2	-	mJ
Turn-On Energy (Note 5)	E <sub>ON2</sub>			-	2.7	3.3	mJ
Turn–Off Energy (Note 4)	E <sub>OFF</sub>			_	2.3	2.8	mJ

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Current Turn-On Delay Time	t <sub>d(ON)I</sub>	IGBT and Diode at $T_J = 150^{\circ}C$ ,	-	22	28	ns
Current Rise Time	t <sub>rl</sub>	$I_{CE} = 27 \text{ A},$ $V_{CE} = 960 \text{ V},$ $V_{GE} = 15 \text{ V},$ $R_G = 3 \Omega,$ L = 1  mH, Test Circuit (Figure 18)	-	20	25	ns
Current Turn-Off Delay Time	t <sub>d(OFF)</sub> I		-	220	280	ns
Current Fall Time	t <sub>fl</sub>		-	140	200	ns
Turn-On Energy (Note 5)	E <sub>ON1</sub>		-	2.7	-	mJ
Turn-On Energy (Note 5)	E <sub>ON2</sub>		-	5.1	6.5	mJ
Turn-Off Energy (Note 4)	E <sub>OFF</sub>		-	3.4	4.2	mJ
Thermal Resistance Junction To Case	$R_{\theta JC}$		-	-	0.25	°C/W

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

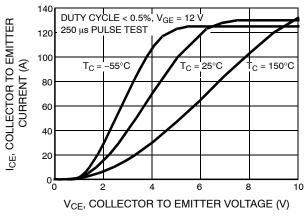
4. Turn-Off Energy Loss (E<sub>OFF</sub>) is defined as the integral of the instantaneous power loss starting at the trailing edge of the input pulse and ending at the point where the collector current equals zero (I<sub>CE</sub> = 0 A). All devices were tested per JEDEC Standard No. 24–1 Method for Measurement of Power Device Turn-Off Switching Loss. This test method produces the true total Turn-Off Energy Loss.

5. Values for two Turn-On loss conditions are shown for the convenience of the circuit designer. E<sub>ON1</sub> is the turn-on loss of the IGBT only. E<sub>ON2</sub> is the turn-on loss when a typical diode is used in the test circuit and the diode is at the same T<sub>J</sub> as the IGBT. The diode type is specified in Figure 18.

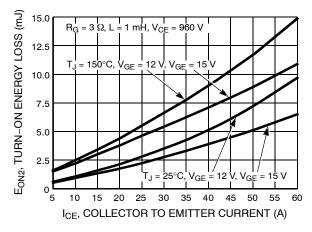


TYPICAL PERFORMANCE CURVES (unless otherwise specified)

### TYPICAL PERFORMANCE CURVES (unless otherwise specified) (continued)









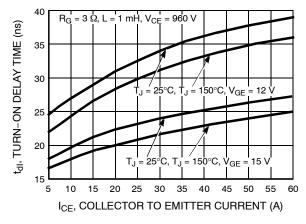


Figure 9. TURN-ON DELAY TIME vs. COLLECTOR TO EMITTER CURRENT

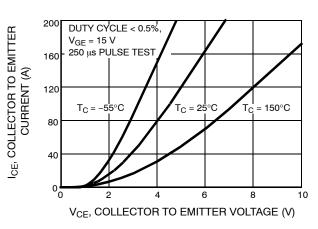


Figure 6. COLLECTOR TO EMITTER ON-STATE VOLTAGE

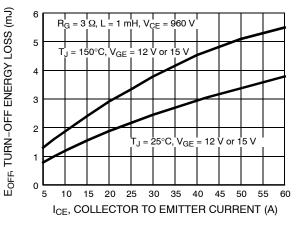
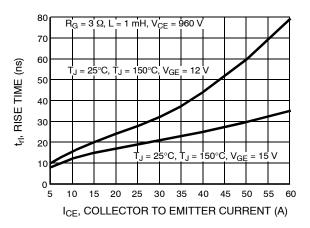
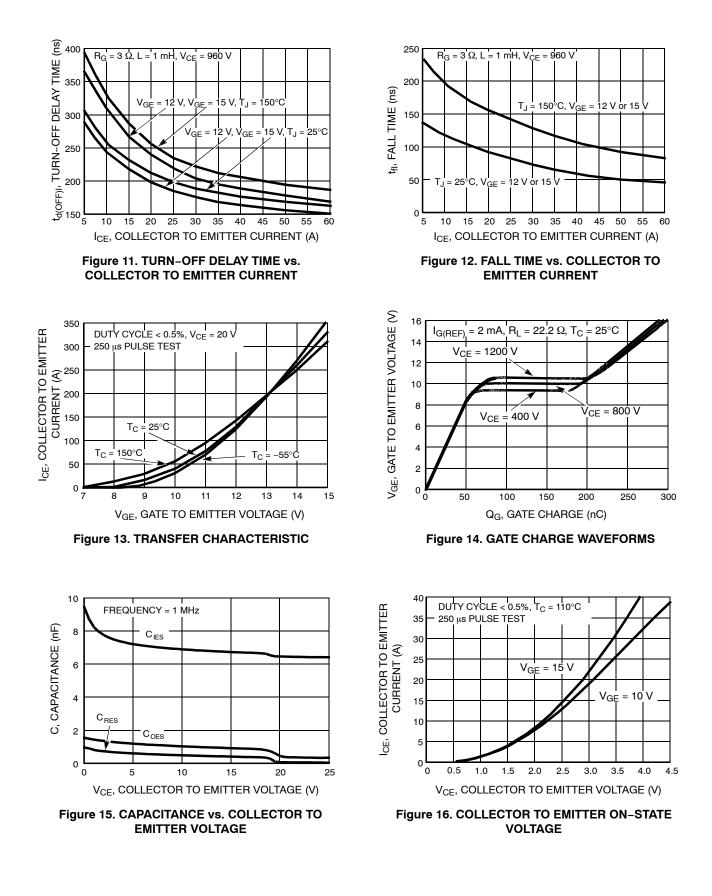


Figure 8. TURN-OFF ENERGY LOSS vs. COLLECTOR TO EMITTER CURRENT





## TYPICAL PERFORMANCE CURVES (unless otherwise specified) (continued)



TYPICAL PERFORMANCE CURVES (unless otherwise specified) (continued)

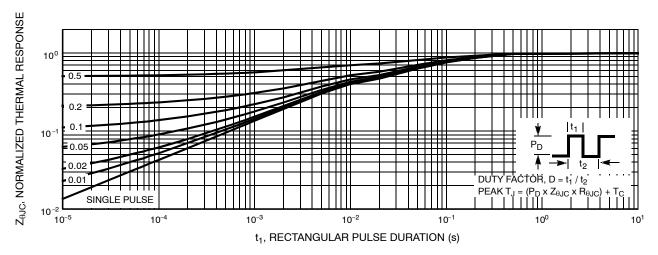


Figure 17. NORMALIZED TRANSIENT THERMAL RESPONSE, JUNCTION TO CASE

**TEST CIRCUIT AND WAVEFORMS** 

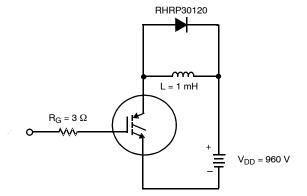


Figure 18. INDUCTIVE SWITCHING TEST CIRCUIT

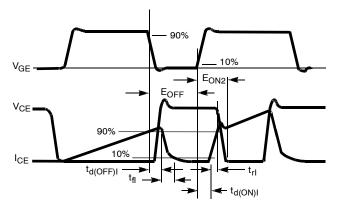


Figure 19. SWITCHING TEST WAVEFORMS

### HANDLING PRECAUTIONS FOR IGBTs

Insulated Gate Bipolar Transistors are susceptible to gateinsulation damage by the electrostatic discharge of energy through the devices. When handling these devices, care should be exercised to assure that the static charge built in the handler's body capacitance is not discharged through the device. With proper handling and application procedures, however, IGBTs are currently being extensively used in production by numerous equipment manufacturers in military, industrial and consumer applications, with virtually no damage problems due to electrostatic discharge. IGBTs can be handled safely if the following basic precautions are taken:

- 1. Prior to assembly into a circuit, all leads should be kept shorted together either by the use of metal shorting springs or by the insertion into conductive material such as "ECCOSORBD<sup>™</sup> LD26" or equivalent.
- 2. When devices are removed by hand from their carriers, the hand being used should be grounded by any suitable means for example, with a metallic wristband.
- 3. Tips of soldering irons should be grounded.
- 4. Devices should never be inserted into or removed from circuits with power on.
- 5. Gate Voltage Rating Never exceed the gate–voltage rating of  $V_{GEM}$ . Exceeding the rated  $V_{GE}$  can result in permanent damage to the oxide layer in the gate region.
- 6. Gate Termination The gates of these devices are essentially capacitors. Circuits that leave the gate open– circuited or floating should be avoided. These conditions can result in turn–on of the device due to voltage buildup on the input capacitor due to leakage currents or pickup.

7. *Gate Protection* - These devices do not have an internal monolithic Zener diode from gate to emitter. If gate protection is required an external Zener is recommended.

### **OPERATING FREQUENCY INFORMATION**

Operating frequency information for a typical device (Figure 3) is presented as a guide for estimating device performance for a specific application. Other typical frequency vs collector current ( $I_{CE}$ ) plots are possible using the information shown for a typical unit in Figures 5, 6, 7, 8, 9 and 11. The operating frequency plot (Figure 3) of a typical device shows  $f_{MAX1}$  or  $f_{MAX2}$ ; whichever is smaller at each point. The information is based on measurements of a typical device and is bounded by the maximum rated junction temperature.

 $f_{MAX1}$  is defined by  $f_{MAX1} = 0.05 / (t_{d(OFF)I} + t_{d(ON)I})$ . Deadtime (the denominator) has been arbitrarily held to 10% of the on–state time for a 50% duty factor. Other definitions are possible.  $t_{d(OFF)I}$  and  $t_{d(ON)I}$  are defined in Figure 19. Device turn–off delay can establish an additional frequency limiting condition for an application other than  $T_{JM}$ .  $t_{d(OFF)I}$  is important when controlling output ripple under a lightly loaded condition.

 $f_{MAX2}$  is defined by  $f_{MAX2} = (P_D - P_C) / (E_{OFF} + E_{ON2})$ . The allowable dissipation (P\_D) is defined by  $P_D = (T_{JM} - T_C) / R_{\theta JC}$ . The sum of device switching and conduction losses must not exceed P\_D. A 50% duty factor was used (Figure 3) and the conduction losses (P\_C) are approximated by  $P_C = (V_{CE} \ x \ I_{CE}) / 2$ .

 $E_{ON2}$  and  $E_{OFF}$  are defined in the switching waveforms shown in Figure 19.  $E_{ON2}$  is the integral of the instantaneous power loss ( $I_{CE} \times V_{CE}$ ) during turn–on and  $E_{OFF}$  is the integral of the instantaneous power loss ( $I_{CE} \times V_{CE}$ ) during turn–off. All tail losses are included in the calculation for  $E_{OFF}$ ; i.e., the collector current equals zero ( $I_{CE} = 0$ ).

#### **ORDERING INFORMATION**

Part Number	Package	Brand	Shipping	
HGTG27N120BN	TO-247	G27N120BN	450 Units / Tube	

NOTE: When ordering, use the entire part number.

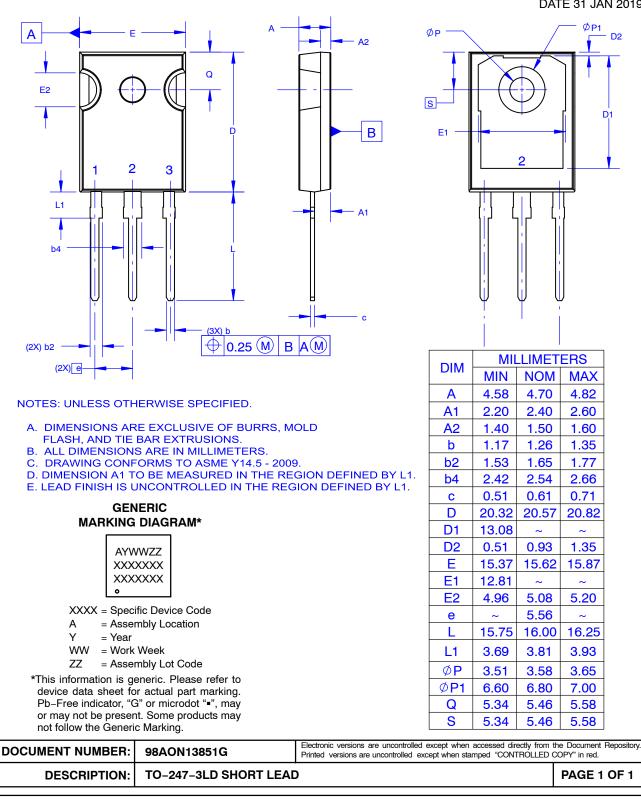
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TO-247-3LD SHORT LEAD CASE 340CK **ISSUE A** 

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