# **Power MOSFET with Current Mirror FET**

## 24 V, 9.5 A, N-Channel, ESD Protected, 1:250 Current Mirror, SO-8 Leadless

N-Channel MOSFET with 1:250 current mirror device utilizing the latest ON Semiconductor technology to achieve low figure of merit while keeping a high accuracy in the linear region. This device takes advantage of the latest leadless QFN package to improve thermal transfer.

#### **Features**

- Current Sense MOSFET
- ±15% Current Mirror Accuracy
- ESD Protected on the Main and the Mirror MOSFET
- Low Gate Charge
- Pb-Free Package is Available\*

#### **Applications**

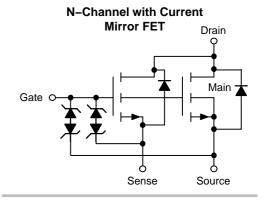
- DC-DC Converters
- Voltage Regulator Modules
- Small DC Motor Controls



### ON Semiconductor®

#### http://onsemi.com

V <sub>DSS</sub>	R <sub>DS(on)</sub> Typ	I <sub>D</sub> MAX
24 V	12 mΩ @ 4.5 V	9.5 A





## DIAGRAM



**MARKING** 

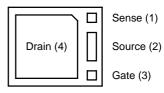
4501N = Device Code A = Assembly Location

PLLP4

CASE 508AA

Y = Year
WW = Work Week
Pb-Free Package

#### **PIN CONNECTIONS**



(Bottom View)

#### ORDERING INFORMATION

Device	Package	Shipping <sup>†</sup>
NILMS4501NR2	PLLP4	2500/Tape & Reel
NILMS4501NR2G	PLLP4 (Pb-Free)	2500/Tape & Reel

<sup>†</sup>For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specification Brochure, BRD8011/D.

<sup>\*</sup>For additional information on our Pb–Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

#### MAIN MOSFET MAXIMUM RATINGS (T<sub>A</sub> = 25°C unless otherwise noted)

Rating	Symbol	Value	Unit
Drain-to-Source Voltage	V <sub>DSS</sub>	24	V
Gate-to-Source Voltage	V <sub>GS</sub>	±10	V
Drain Current (Note 1) Continuous @ $T_A = 25^{\circ}C$ Continuous @ $T_A = 100^{\circ}C$ Pulsed ( $t_p \le 10$ s)	I <sub>D</sub> I <sub>D</sub> I <sub>DM</sub>	9.5 6.7 14	Adc Adc Apk
Total Power Dissipation @ $T_A = 25$ °C (Note 1) Total Power Dissipation @ $T_A = 25$ °C (Note 2)	P <sub>D</sub> P <sub>D</sub>	2.7 1.4	W
Thermal Resistance Junction–to–Ambient (Note 1) Junction–to–Ambient (Note 2) Junction–to–Ambient ( $t_p \le 10$ s) (Note 3)	R <sub>θ</sub> ja R <sub>θ</sub> ja R <sub>θ</sub> ja	55 110 25	°C/W
Operating Junction and Storage Temperature	$T_J, T_{STG}$	-55 to 175	°C
Single Pulse Drain–to–Source Avalanche ( $V_{DD}$ = 24 V, $V_{GS}$ = 10 V, $I_{L}$ = 9.5 A, L = 1.0 mH, $R_{G}$ = 25 $\Omega$ )	E <sub>AS</sub>	50	mJ
Electrostatic Discharge Capability Human Body Model Charged Device Model	ESD <sub>HBM</sub> CMD	4000 2000	V

Stresses exceeding Maximum Ratings may damage the device. Maximum Ratings are stress ratings only. Functional operation above the Recommended Operating Conditions is not implied. Extended exposure to stresses above the Recommended Operating Conditions may affect device reliability.

- Surface mounted on FR4 board using 1 in sq pad size (Cu area = 1.127 in sq [1 oz] including traces).
   Surface mounted on FR4 board using the minimum recommended pad size (Cu area = 0.0821 in sq).
- 3. Surface mounted on FR4 board using 1 in sq pad size (Cu area = 1.127 in sq [1 oz] including traces) and 200 LFM airflow.

#### MAIN MOSFET ELECTRICAL CHARACTERISTICS ( $T_J = 25$ °C unless otherwise noted)

Characteristic	Symbol	Min	Тур	Max	Unit
OFF CHARACTERISTICS		•			•
Drain-to-Source Breakdown Voltage (V <sub>GS</sub> = 0 V, I <sub>D</sub> = 250 μA) Temperature Coefficient (Positive)	V <sub>(BR)DSS</sub>	24	29 23	_ _	V mV/°C
Zero Gate Voltage Drain Current $(V_{DS} = 24 \text{ V}, V_{GS} = 0 \text{ V})$ $(V_{DS} = 24 \text{ V}, V_{GS} = 0 \text{ V}, T_{J} = 125^{\circ}\text{C})$ $(V_{DS} = 24 \text{ V}, V_{GS} = 0 \text{ V}, T_{J} = 175^{\circ}\text{C})$	I <sub>DSS</sub>	- - -	0.05 1.0 30	1.0 100 100	μΑ
Gate-Body Leakage Current $(V_{GS} = 3.0 \text{ V}, V_{DS} = 0 \text{ V})$ $(V_{GS} = 9.0 \text{ V}, V_{DS} = 0 \text{ V})$	less		40 1.3	100 10	nA μA
ON CHARACTERISTICS					
Gate Threshold Voltage $(V_{DS} = V_{GS}, I_D = 250 \mu A)$ Threshold Temperature Coefficient (Negative)	V <sub>GS(th)</sub>	1.1	1.60 -5.0	2.0	V mV/°C
Static Drain-to-Source On-Resistance (Note 4) $ \begin{array}{c} (V_{GS}=10 \text{ V}, I_D=6.0 \text{ A}, T_J @ 25^{\circ}\text{C}) \\ (V_{GS}=10 \text{ V}, I_D=6.0 \text{ A}, T_J @ 125^{\circ}\text{C}) \\ (V_{GS}=10 \text{ V}, I_D=6.0 \text{ A}, T_J @ 175^{\circ}\text{C}) \end{array} $	R <sub>DS(on)</sub>	- - -	9.0 12 14	13 17 20	mΩ
Static Drain-to-Source On-Resistance (Note 4) ( $V_{GS} = 4.5 \text{ V}$ , $I_D = 6.0 \text{ A}$ , $T_J @ 25^{\circ}\text{C}$ ) ( $V_{GS} = 4.5 \text{ V}$ , $I_D = 6.0 \text{ A}$ , $T_J @ 125^{\circ}\text{C}$ ) ( $V_{GS} = 4.5 \text{ V}$ , $I_D = 6.0 \text{ A}$ , $T_J @ 175^{\circ}\text{C}$ )	R <sub>DS(on)</sub>	- - -	12 16 18	16 20 24	mΩ
Main/Mirror MOSFET Current Ratio $(V_{GS} = 4.5 \text{ V}, I_D = 1.0 \text{ A})$ $(V_{GS} = 4.5 \text{ V}, I_D = 1.0 \text{ A}, T_A = 175^{\circ}\text{C})$	I <sub>RAT</sub>	212	250 268	287 -	_
Forward Transconductance (Note 4) $(V_{DS} = 6.0 \text{ V}, I_{D} = 6.0 \text{ A})$	9FS	15	23	-	Mhos
4. Pulse Test: Pulse Width = 300 μs, Duty Cycle = 2%.					

## $\textbf{MAIN MOSFET ELECTRICAL CHARACTERISTICS (continued)} \ (T_J = 25^{\circ}C \ unless \ otherwise \ noted)$

	Characteristic	Symbol	Min	Тур	Max	Unit
DYNAMIC CHARACTERISTI	CS (Note 6)	•	•			
Input Capacitance		C <sub>iss</sub>	_	1380	1500	pF
Output Capacitance	$(V_{DS} = 6.0 \text{ V}, V_{GS} = 0 \text{ V}, f = 1.0 \text{ MHz})$	C <sub>oss</sub>	_	870	1000	
Transfer Capacitance		C <sub>rss</sub>	_	275	350	
SWITCHING CHARACTERIS	TICS (Note 6)					
Turn-On Delay Time		t <sub>d(on)</sub>	_	12	14	ns
Rise Time	$(V_{DD} = 6.0 \text{ V}, I_D = 2.0 \text{ A},$	t <sub>r</sub>	_	15	18	
Turn-Off Delay Time	$V_{GS} = 4.5 \text{ V}, R_{G} = 2.5 \Omega$	t <sub>d(off)</sub>	_	17	20	
Fall Time		t <sub>f</sub>	_	6.0	8.0	
Turn-On Delay Time		t <sub>d(on)</sub>	_	8.5	11	ns
Rise Time	$(V_{DD} = 6.0 \text{ V}, I_D = 2.0 \text{ A},$	t <sub>r</sub>	_	15	20	
Turn-Off Delay Time	$V_{GS} = 10 \text{ V}, R_G = 2.5 \Omega$	t <sub>d(off)</sub>	_	22.5	27	
Fall Time		t <sub>f</sub>	_	6.5	9.0	
Gate Charge		Q <sub>T</sub>	_	11	14	nC
	0/ 60// 20/ / 45/0	Q <sub>G(th)</sub>	_	1.7	2.5	
	$(V_{DS} = 6.0 \text{ V}, I_{D} = 2.0 \text{ A}, V_{GS} = 4.5 \text{ V})$	Q <sub>gs</sub>	_	3.5	4.5	
		Q <sub>gd</sub>	_	3.6	4.3	
Gate Charge		Q <sub>T</sub>	_	23.5	25	nC
	()/ -60)/   - 20   1   - 10   1	Q <sub>G(th)</sub>	_	4.4	5.5	
	$(V_{DS} = 6.0 \text{ V}, I_D = 2.0 \text{ A}, V_{GS} = 10 \text{ V})$	Q <sub>gs</sub>	_	5.6	10	
		Q <sub>gd</sub>	-	2.5	7.0	
SOURCE-DRAIN DIODE CH	ARACTERISTICS					
Forward On–Voltage (Notes 5 & 6)	$(I_S = 6.0 \text{ A}, V_{GS} = 0 \text{ V})$ $(I_S = 6.0 \text{ A}, V_{GS} = 0 \text{ V}, T_J = 175^{\circ}\text{C})$	V <sub>SD</sub>	_ _	0.80 0.57	1.1	V
Reverse Recovery Time		t <sub>rr</sub>	_	42	55	ns
(Note 6)		ta	_	19.5	25	
	$(I_S = 3.0 \text{ A}, V_{GS} = 0 \text{ V}, dI_S/dt = 100 \text{ A/}\mu\text{s})$	t <sub>b</sub>	_	22.5	30	
Reverse Recovery Stored Charge (Note 6)		Q <sub>RR</sub>	-	0.042	0.06	μС

<sup>5.</sup> Pulse Test: Pulse Width ≤ 300 μs, Duty Cycle ≤ 2%.
6. Switching characteristics are independent of operating junction temperatures.

#### TYPICAL ELECTRICAL CHARACTERISTICS

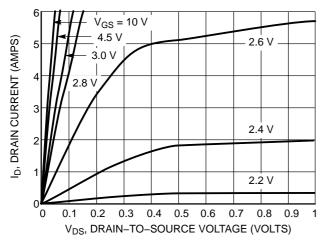


Figure 1. On-Region Characteristics

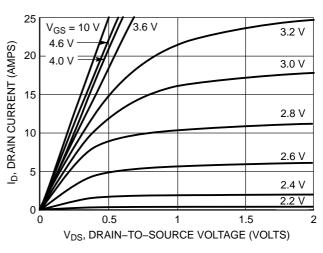


Figure 2. On-Region Characteristics

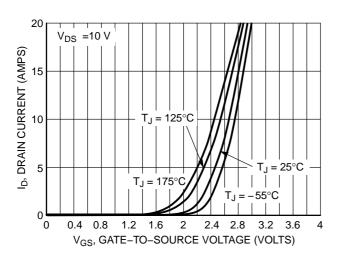


Figure 3. Transfer Characteristics

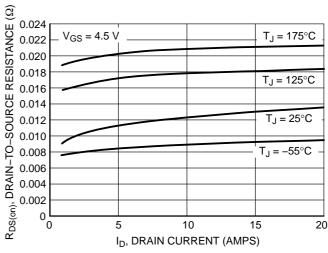


Figure 4. On–Resistance versus Drain Current and Temperature

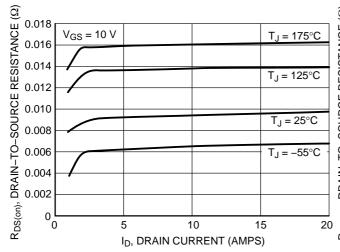


Figure 5. On–Resistance versus Drain Current and Temperature

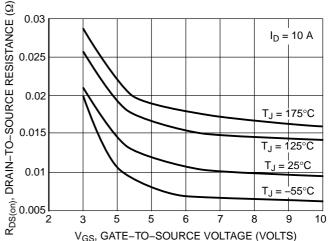
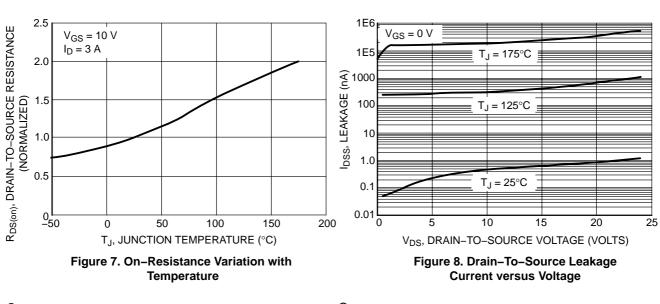


Figure 6. On–Resistance versus Gate Voltage and Temperature

#### TYPICAL ELECTRICAL CHARACTERISTICS



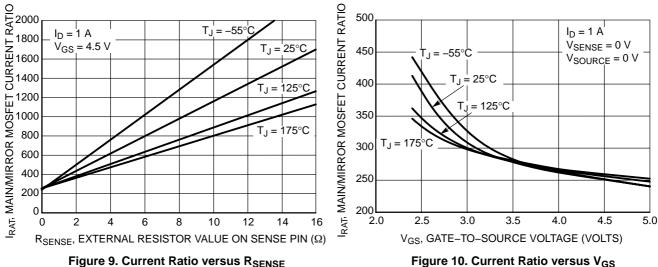


Figure 9. Current Ratio versus R<sub>SENSE</sub>

 $I_D = 1 A$ 

V<sub>SENSE</sub> = 0 V

 $V_{GS} = 4.5 \text{ V}$ 

 $T_1 = -55^{\circ}C$ 

25°C

125°C

0

= 175°C 0 -0.01 -0.005

RAT, MAIN/MIRROR MOSFET CURRENT RATIO 2000 –55°Ċ 25°C  $I_D = 1 A'$ 1800 V<sub>SOURCE</sub> = 0 V 125°C 175°C 1600  $V_{GS} = 4.5 V$ 1400 1200 1000 800 600 400 200 0 0.01 0.015 0.02 0.025 0.03 -0.005 -0.01 0.005 0.01 0.015 0.02 V<sub>SOURCE</sub>, VOLTAGE DROP FROM SOURCE PIN TO GROUND (V) V<sub>SENSE</sub>, VOLTAGE DROP FROM SENSE PIN TO GROUND (V)

Figure 11. I<sub>RATIO</sub> versus V<sub>SOURCE</sub>

0.005

Figure 12. Current Ratio versus V<sub>SENSE</sub>

#### **POWER MOSFET SWITCHING**

Switching behavior is most easily modeled and predicted by recognizing that the power MOSFET is charge controlled. The lengths of various switching intervals ( $\Delta t$ ) are determined by how fast the FET input capacitance can be charged by current from the generator.

The published capacitance data is difficult to use for calculating rise and fall because drain–gate capacitance varies greatly with applied voltage. Accordingly, gate charge data is used. In most cases, a satisfactory estimate of average input current ( $I_{G(AV)}$ ) can be made from a rudimentary analysis of the drive circuit so that

$$t = Q/I_{G(AV)}$$

During the rise and fall time interval when switching a resistive load,  $V_{GS}$  remains virtually constant at a level known as the plateau voltage,  $V_{SGP}$ . Therefore, rise and fall times may be approximated by the following:

$$t_r = Q_2 x R_G/(V_{GG} - V_{GSP})$$
  
$$t_f = Q_2 x R_G/V_{GSP}$$

where

 $V_{GG}$  = the gate drive voltage, which varies from zero to  $V_{GG}$   $R_G$  = the gate drive resistance

and Q2 and VGSP are read from the gate charge curve.

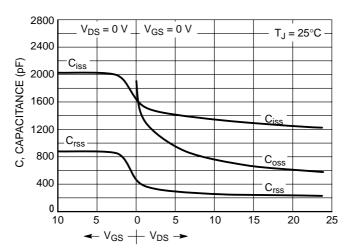
During the turn-on and turn-off delay times, gate current is not constant. The simplest calculation uses appropriate values from the capacitance curves in a standard equation for voltage change in an RC network. The equations are:

$$\begin{split} t_{d(on)} &= R_G \; C_{iss} \; In \; [V_{GG}/(V_{GG} - V_{GSP})] \\ t_{d(off)} &= R_G \; C_{iss} \; In \; (V_{GG}/V_{GSP}) \end{split}$$

The capacitance ( $C_{iss}$ ) is read from the capacitance curve at a voltage corresponding to the off–state condition when calculating  $t_{d(on)}$  and is read at a voltage corresponding to the on–state when calculating  $t_{d(off)}$ .

At high switching speeds, parasitic circuit elements complicate the analysis. The inductance of the MOSFET source lead, inside the package and in the circuit wiring which is common to both the drain and gate current paths, produces a voltage at the source which reduces the gate drive current. The voltage is determined by Ldi/dt, but since di/dt is a function of drain current, the mathematical solution is complex. The MOSFET output capacitance also complicates the mathematics. And finally, MOSFETs have finite internal gate resistance which effectively adds to the resistance of the driving source, but the internal resistance is difficult to measure and, consequently, is not specified.

The resistive switching time variation versus gate resistance shows how typical switching performance is affected by the parasitic circuit elements. If the parasitics were not present, the slope of the curves would maintain a value of unity regardless of the switching speed. The circuit used to obtain the data is constructed to minimize common inductance in the drain and gate circuit loops and is believed readily achievable with board mounted components. Most power electronic loads are inductive; the data in the figure is taken with a resistive load, which approximates an optimally snubbed inductive load. Power MOSFETs may be safely operated into an inductive load; however, snubbing reduces switching losses.



GATE-TO-SOURCE OR DRAIN-TO-SOURCE VOLTAGE (VOLTS)

Figure 13. Capacitance Variation

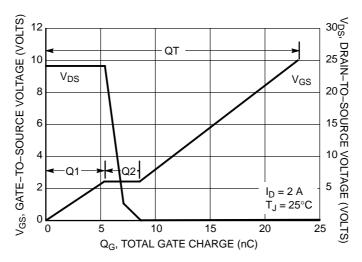


Figure 14. Gate-To-Source and Drain-To-Source Voltage versus Total Charge

#### DRAIN-TO-SOURCE DIODE CHARACTERISTICS

The switching characteristics of a MOSFET body diode are very important in systems using it as a freewheeling or commutating diode. Of particular interest are the reverse recovery characteristics which play a major role in determining switching losses, radiated noise, EMI and RFI.

System switching losses are largely due to the nature of the body diode itself. The body diode is a minority carrier device, therefore it has a finite reverse recovery time,  $t_{rr}$ , due to the storage of minority carrier charge,  $Q_{RR}$ , as shown in the typical reverse recovery wave form of Figure 15. It is this stored charge that, when cleared from the diode, passes through a potential and defines an energy loss. Obviously, repeatedly forcing the diode through reverse recovery further increases switching losses. Therefore, one would like a diode with short  $t_{rr}$  and low  $Q_{RR}$  specifications to minimize these losses.

The abruptness of diode reverse recovery effects the amount of radiated noise, voltage spikes, and current ringing. The mechanisms at work are finite irremovable circuit parasitic inductances and capacitances acted upon by

high di/dts. The diode's negative di/dt during  $t_a$  is directly controlled by the device clearing the stored charge. However, the positive di/dt during  $t_b$  is an uncontrollable diode characteristic and is usually the culprit that induces current ringing. Therefore, when comparing diodes, the ratio of  $t_b/t_a$  serves as a good indicator of recovery abruptness and thus gives a comparative estimate of probable noise generated. A ratio of 1 is considered ideal and values less than 0.5 are considered snappy.

Compared to ON Semiconductor standard cell density low voltage MOSFETs, high cell density MOSFET diodes are faster (shorter t<sub>IT</sub>), have less stored charge and a softer reverse recovery characteristic. The softness advantage of the high cell density diode means they can be forced through reverse recovery at a higher di/dt than a standard cell MOSFET diode without increasing the current ringing or the noise generated. In addition, power dissipation incurred from switching the diode will be less due to the shorter recovery time and lower switching losses.

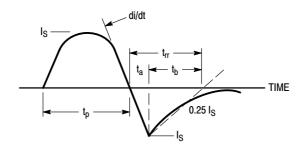


Figure 15. Diode Reverse Recovery Waveform

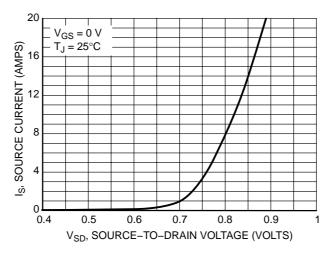


Figure 16. Diode Forward Voltage versus Current

#### SAFE OPERATING AREA

The Forward Biased Safe Operating Area curves define the maximum simultaneous drain–to–source voltage and drain current that a transistor can handle safely when it is forward biased. Curves are based upon maximum peak junction temperature and a case temperature (T<sub>C</sub>) of 25°C. Peak repetitive pulsed power limits are determined by using the thermal response data in conjunction with the procedures discussed in AN569, "Transient Thermal Resistance – General Data and Its Use."

Switching between the off-state and the on-state may traverse any load line provided neither rated peak current ( $I_{DM}$ ) nor rated voltage ( $V_{DSS}$ ) is exceeded, and that the transition time ( $t_{t_1}$ ,  $t_f$ ) does not exceed 10  $\mu$ s. In addition the

total power averaged over a complete switching cycle must not exceed  $(T_{J(MAX)} - T_C)/(R_{\theta JC})$ .

A power MOSFET designated E–FET can be safely used in switching circuits with unclamped inductive loads. For reliable operation, the stored energy from circuit inductance dissipated in the transistor while in avalanche must be less than the rated limit and must be adjusted for operating conditions differing from those specified. Although industry practice is to rate in terms of energy, avalanche energy capability is not a constant. The energy rating decreases non–linearly with an increase of peak current in avalanche and peak junction temperature.

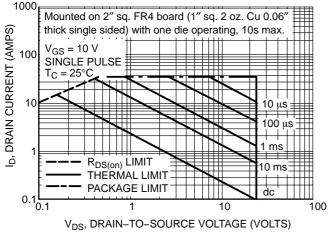


Figure 17. Maximum Rated Forward Biased Safe Operating Area

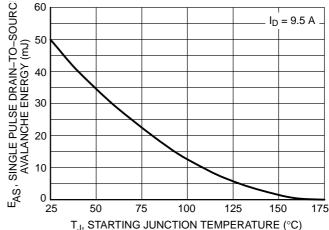


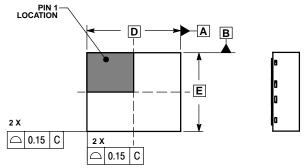
Figure 18. Maximum Avalanche Energy versus Starting Junction Temperature

**DATE 21 JUL 2004** 

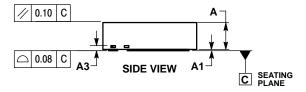
#### PLLP4 CASE 508AA-01 **ISSUE O**

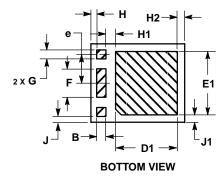






**TOP VIEW** 





- noties:

  1. DIMENSIONS AND TOLERANCING PER ASME Y14.5M, 1994.

  2. DIMENSIONS IN MILLIMETERS.

  3. TOLERANCES: ±0.10 MM.

	MILLIMETERS		
DIM	MIN	MAX	
Α	1.750	1.950	
A1	0.000	0.050	
A3	0.254	REF	
В	0.500	0.700	
D	6.200	BSC	
D1	3.979	4.179	
E	5.200	BSC	
E1	4.087	4.287	
е	1.905	BSC	
F	1.860	1.880	
G	0.500	0.700	
Н	0.379 REF 0.635 REF 0.507 REF 0.404 REF 0.507 REF		
H1			
H2			
J			
J1			

#### **GENERIC MARKING DIAGRAM\***



= Specific Device Code XXXXX = Assembly Location Α Υ

= Year WW = Work Week

\*This information is generic. Please refer to device data sheet for actual part marking. Pb–Free indicator, "G" or microdot " •", may or may not be present.

DOCUMENT NUMBER:	98AON19199D	Ele
STATUS:	ON SEMICONDUCTOR STANDARD	aco ver
NEW STANDARD:		"Co
DESCRIPTION:	PLLP4, 6.2X5.2X1.85 MM PITCH	

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<b>98AON1</b>	9199	)D

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